

NCS21911, NCV21911, NCS21912, NCV21912, NCS21914, NCV21914

Precision Operational Amplifier, 25 μ V Offset, Zero-Drift, 36 V Supply, 2 MHz

The NCS2191x family of high precision op amps feature low input offset voltage and near-zero drift over time and temperature. These op amps operate over a wide supply range from 4 V to 36 V with low quiescent current. The rail-to-rail output swings within 10 mV of the rails. The family includes the single channel NCS(V)21911, the dual channel NCS(V)21912, and the quad channel NCS(V)21914 in a variety of packages. All versions are specified for operation from -40°C to $+125^{\circ}\text{C}$. Automotive qualified options are available under the NCV prefix.

Features

- Input Offset Voltage: 25 μ V max
- Zero-Drift Offset Voltage: 0.085 μ V/ $^{\circ}\text{C}$ max
- Voltage Noise Density: 22 nV/ $\sqrt{\text{Hz}}$ typical
- Unity Gain Bandwidth: 2 MHz typical
- Supply Voltage: 4 V to 36 V
- Quiescent Current: 570 μ A max
- Rail-to-Rail Output
- NCV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC-Q100 Qualified and PPAP Capable
- These Devices are Pb-free, Halogen free/BFR free and are RoHS compliant

Typical Applications

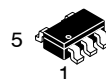
- Temperature Measurements
- Transducer Applications
- Electronic Scales
- Medical Instrumentation
- Current Sensing
- Automotive



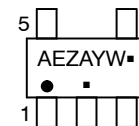
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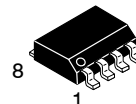
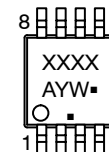
MARKING DIAGRAMS



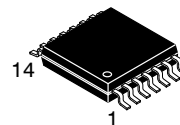
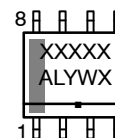
TSOP-5
CASE 483



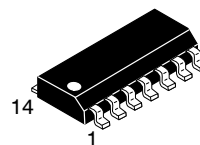
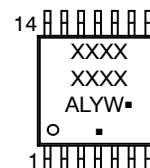
Micro8
CASE 846A-02



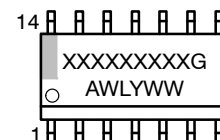
SOIC-8 NB
CASE 751-07



TSSOP-14 WB
CASE 948G



SOIC-14 NB
CASE 751A-03



XXXXX = Specific Device Code
A = Assembly Location
L or WL = Wafer Lot
Y = Year
W = Work Week
▪ = Pb-Free Package

(Note: Microdot may be in either location)

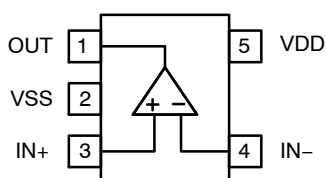
ORDERING INFORMATION

See detailed ordering and shipping information on page 2 of this data sheet.

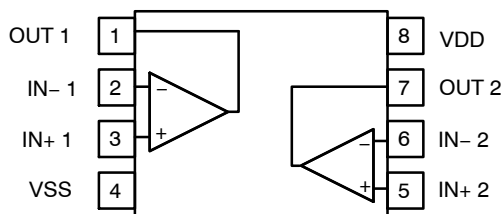
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PIN CONNECTIONS

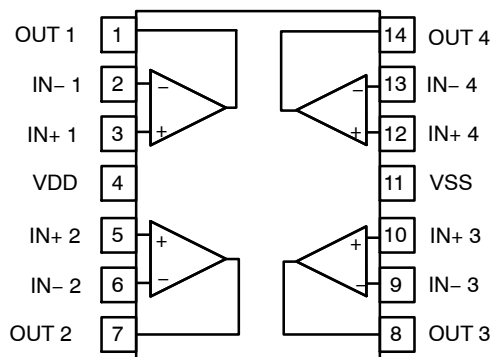
Single Channel Configuration
NCS21911



Dual Channel Configuration
NCS21912



Quad Channel Configuration
NCS21914



ORDERING INFORMATION

Channels	Device	Package	Shipping †
Single	NCS21911SN2T1G	SOT23-5 / TSOP-5	3000 / Tape & Reel
Dual	NCS21912DR2G (In Development*)	SOIC-8	2500 / Tape & Reel
	NCS21912DMR2G (In Development*)	MICRO-8	4000 / Tape & Reel
Quad	NCS21914DR2G (In Development*)	SOIC-14	2500 / Tape & Reel
	NCS21914DBR2G (In Development*)	TSSOP-14	2500 / Tape & Reel
Automotive Qualified			
Channels	Device	Package	Shipping †
Single	NCV21911SN2T1G	SOT23-5 / TSOP-5	3000 / Tape & Reel
Dual	NCV21912DR2G (In Development*)	SOIC-8	2500 / Tape & Reel
	NCV21912DMR2G (In Development*)	MICRO-8	4000 / Tape & Reel
Quad	NCV21914DR2G (In Development*)	SOIC-14	2500 / Tape & Reel
	NCV21914DBR2G (In Development*)	TSSOP-14	2500 / Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

*Contact local sales office for more information.

ABSOLUTE MAXIMUM RATINGS

Parameter	Rating	Unit
Supply Voltage (VDD – VSS)	40	V

INPUT AND OUTPUT PINS

Input Voltage (Note 1)	VSS – 0.3 to VDD + 0.3	V
Differential Input Voltage (Note 2)	±17	V
Input Current (Notes 1 and 2)	±10	mA
Output Short Circuit Current (Note 3)	Continuous	mA

TEMPERATURE

Operating Temperature	–40 to +125	°C
Storage Temperature	–65 to +150	°C
Junction Temperature	+150	°C

ESD RATINGS (Note 4)

Human Body Model (HBM)	3000	V
Charged Device Model (CDM)	2000	V

OTHER RATINGS

Latch-up Current (Note 5)	100	mA
MSL	Level 1	

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.3 V beyond the supply rails should be current limited to 10 mA or less.
2. The inputs are diode connected with a total input protection of 1.65 kΩ, increasing the absolute maximum differential voltage to ±17 V_{DC}. If the applied differential voltage is expected to exceed this rating, external resistors should be added in series with the inputs to limit the input current to ±10 mA.
3. Short-circuit to V_{DD} or V_{SS}. Short circuits to either rail can cause an increase in the junction temperature. The total power dissipation must be limited to prevent the junction temperature from exceeding the 150°C limit.
4. This device series incorporates ESD protection and is tested by the following methods:
ESD Human Body Model tested per JEDEC standard JS-001-2017 (AEC-Q100-002)
ESD Charged Device Model tested per JEDEC standard JS-002-2014 (AEC-Q100-011)
5. Latch-up Current tested per JEDEC standard JESD78E (AEC-Q100-004).

THERMAL INFORMATION (Note 6)

Rating	Symbol	Package	Value	Unit
Thermal Resistance, Junction to Ambient	θ_{JA}	TSOP-5 / SOT23-5	170	°C/W
		Micro8/MSOP8	TBD	
		SOIC-8	TBD	
		SOIC-14	TBD	
		TSSOP-14	TBD	

6. As mounted on an 80x80x1.5 mm FR4 PCB with 2S2P, 2 oz copper, and a 200 mm² heat spreader area. Following JEDEC JESD51-7 guidelines.

OPERATING CONDITIONS

Parameter	Symbol	Range	Unit
Supply Voltage (V _{DD} – V _{SS})	V _S	4 to 36	V
Specified Operating Temperature Range	T _A	–40 to 125	°C
Input Common Mode Voltage Range	V _{CM}	V _{SS} to V _{DD} –1.5	V
Differential Voltage (Note 7)	V _{DIFF}	±17	V

7. The inputs are diode connected with a total input protection of 1.65 kΩ, increasing the absolute maximum differential voltage to ±17 V_{DC}. If the applied differential voltage is expected to exceed this rating, external resistors should be added in series with the inputs to limit the input current to ±10 mA.

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ELECTRICAL CHARACTERISTICS $V_S = 4\text{ V to }36\text{ V}$

At $T_A = +25^\circ\text{C}$, $R_L = 10\text{ k}\Omega$ connected to midsupply, $V_{CM} = V_{OUT} = \text{midsupply}$, unless otherwise noted.

Boldface limits apply over the specified temperature range, $T_A = -40^\circ\text{C to }125^\circ\text{C}$, guaranteed by characterization and/or design.

Parameter	Symbol	Conditions	Min	Typ	Max	Unit
INPUT CHARACTERISTICS						
Offset Voltage	V_{OS}			± 1	± 25	μV
Offset Voltage Drift vs Temp	$\Delta V_{OS}/\Delta T$			± 0.02	± 0.085	$\mu\text{V}/^\circ\text{C}$
Input Bias Current	I_{IB}			± 100	± 500	pA
					± 3500	pA
Input Offset Current	I_{OS}			± 200	± 500	pA
					± 3500	pA
Common Mode Rejection Ratio	CMRR	$V_{SS} \leq V_{CM} \leq V_{DD} - 1.5\text{ V}$	$V_S = 36\text{ V}$	140	150	dB
				130		
			$V_S = 12\text{ V}$ (Note 8)	130	150	
				120		
			$V_S = 8\text{ V}$ (Note 8)	130	140	
				120		
			$V_S = 4\text{ V}$	120	130	
				110		
Input Capacitance	C_{IN}	Common Mode		3		pF
EMI Rejection Ratio	EMIRR	$f = 5\text{ GHz}$		100		dB
		$f = 400\text{ MHz}$		80		

OUTPUT CHARACTERISTICS

Open Loop Voltage Gain	A_{VOL}	$V_{SS} + 0.5\text{ V} < V_O < V_{DD} - 0.5\text{ V}$	130	150		dB
			125	135		
Open Loop Output Impedance	Z_{OUT_OL}	No Load		See Figure 23		Ω
Output Voltage High, Referenced to Rail	V_{OH}	No Load		5	10	mV
		$R_L = 10\text{ k}\Omega$		100	210	
				140	250	
Output Voltage Low, Referenced to Rail	V_{OL}	No Load		5	10	mV
		$R_L = 10\text{ k}\Omega$		100	210	
				140	250	
Short Circuit Current	I_{SC}	Sinking Current		18		mA
		Sourcing Current		16		
Capacitive Load Drive	C_L			1		nF

DYNAMIC PERFORMANCE

Gain Bandwidth Product	GBW	$C_L = 100\text{ pF}$		2		MHz
Gain Margin	A_M	$C_L = 100\text{ pF}$		13		dB
Phase Margin	ϕ_M	$C_L = 100\text{ pF}$		55		$^\circ$
Slew Rate	SR	$G = +1$		1.6		$\text{V}/\mu\text{s}$
Settling Time	t_S	$V_S = 36\text{ V}$	0.1%	20		μs
			0.01%	45		μs
Overload Recovery Time	t_{OR}	$V_S = \pm 18\text{ V}$, $A_V = -10$, $V_{IN} = \pm 2.5\text{ V}$		1		μs

8. Guaranteed by characterization and/or design.

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ELECTRICAL CHARACTERISTICS $V_S = 4\text{ V to }36\text{ V}$

At $T_A = +25^\circ\text{C}$, $R_L = 10\text{ k}\Omega$ connected to midsupply, $V_{CM} = V_{OUT} = \text{midsupply}$, unless otherwise noted.

Boldface limits apply over the specified temperature range, $T_A = -40^\circ\text{C}$ to 125°C , guaranteed by characterization and/or design.

Parameter	Symbol	Conditions	Min	Typ	Max	Unit
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NOISE PERFORMANCE

Total Harmonic Distortion + Noise	THD+N	$f_{IN} = 1\text{ kHz}$, $A_V = 1$, $V_{OUT} = 1\text{ V}_{rms}$		0.0003		%
Voltage Noise Density	e_N	$f = 1\text{ kHz}$		22		$\text{nV}/\sqrt{\text{Hz}}$
Current Noise Density	i_N	$f = 1\text{ kHz}$		100		$\text{fA}/\sqrt{\text{Hz}}$
Voltage Noise, Peak-to-Peak	e_{pp}	$f = 0.1\text{ Hz to }10\text{ Hz}$		400		nV_{pp}
Voltage Noise, RMS	e_{rms}	$f = 0.1\text{ Hz to }10\text{ Hz}$		70		nV_{rms}

POWER SUPPLY

Power Supply Rejection Ratio	PSRR	$V_S = 4\text{ V to }36\text{ V}$		0.02	0.3	$\mu\text{V}/\text{V}$
			130	154		dB
Quiescent Current	I_Q	Per channel		475	570	μA
					570	

GRAPHS

Typical performance at $T_A = 25^\circ\text{C}$, unless otherwise noted.

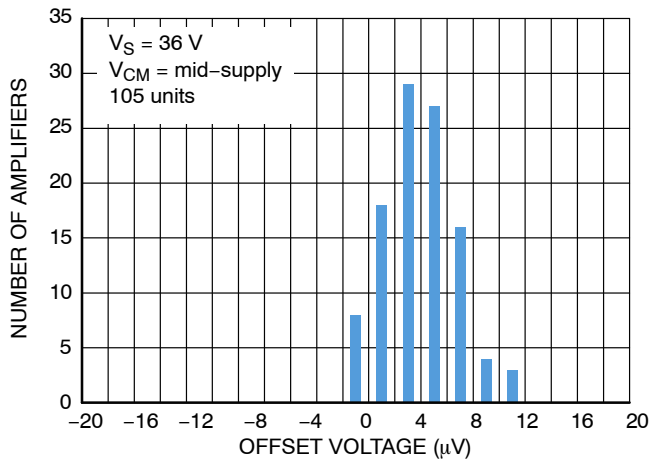


Figure 1. Offset Voltage Distribution

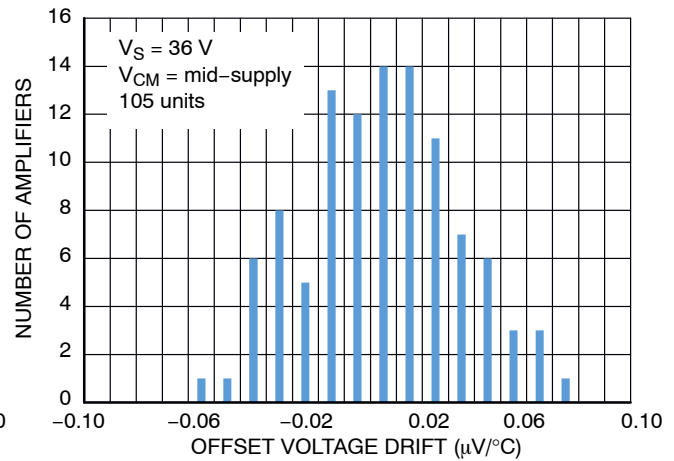


Figure 2. Offset Voltage Drift Distribution

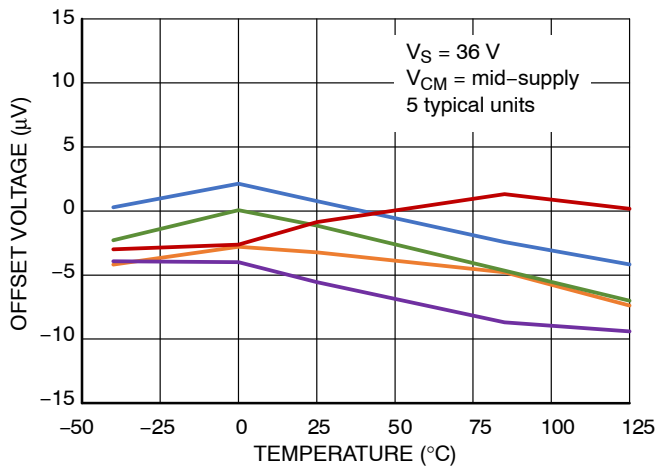


Figure 3. Offset Voltage vs. Temperature

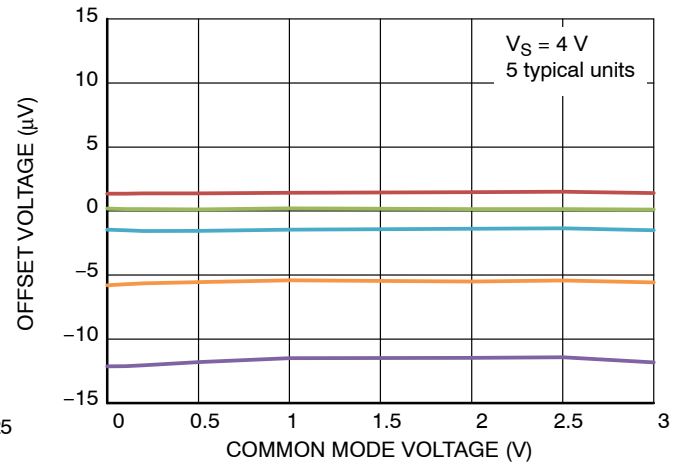


Figure 4. Offset Voltage vs. Common Mode Voltage

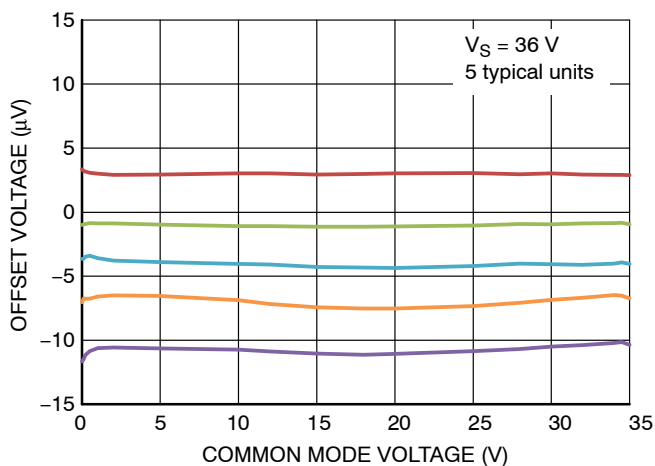


Figure 5. Offset Voltage vs. Common Mode Voltage

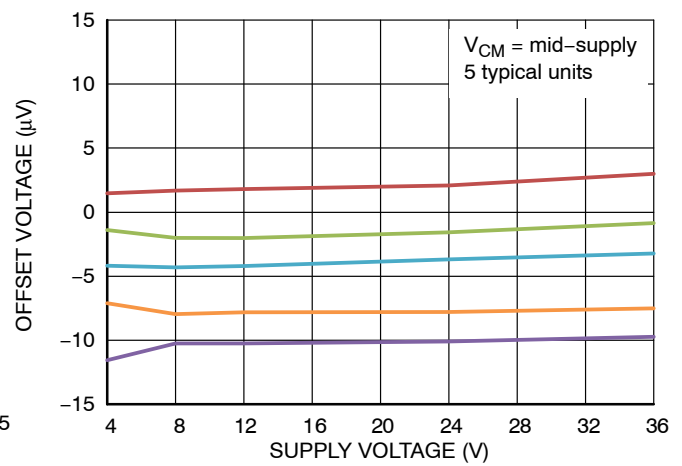


Figure 6. Offset Voltage vs. Power Supply

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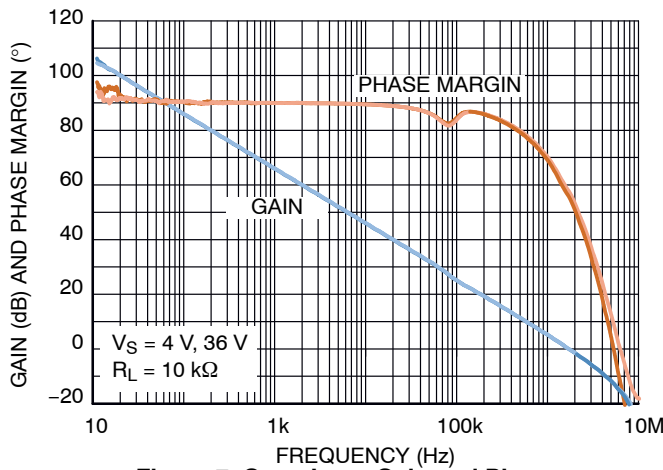


Figure 7. Open Loop Gain and Phase vs. Frequency

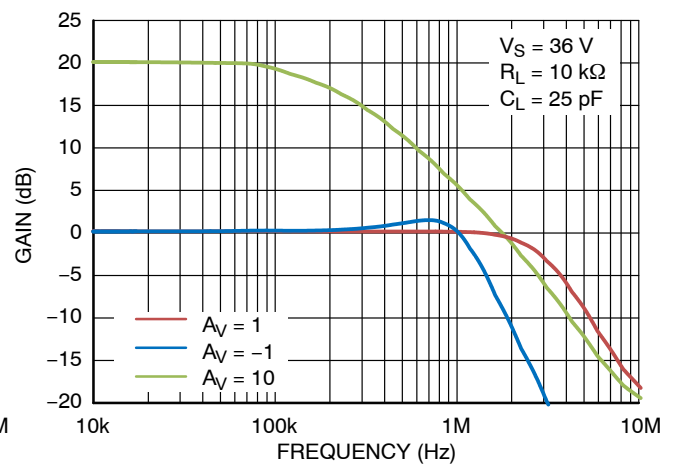


Figure 8. Closed Loop Gain vs. Frequency

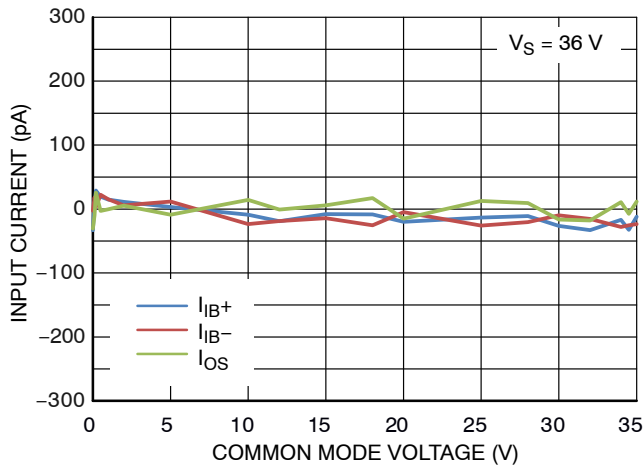


Figure 9. Input Current vs. Common Mode Voltage

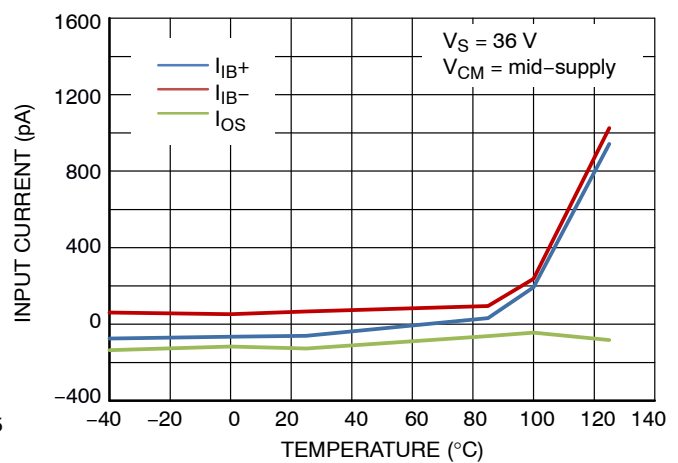


Figure 10. Input Current vs. Temperature

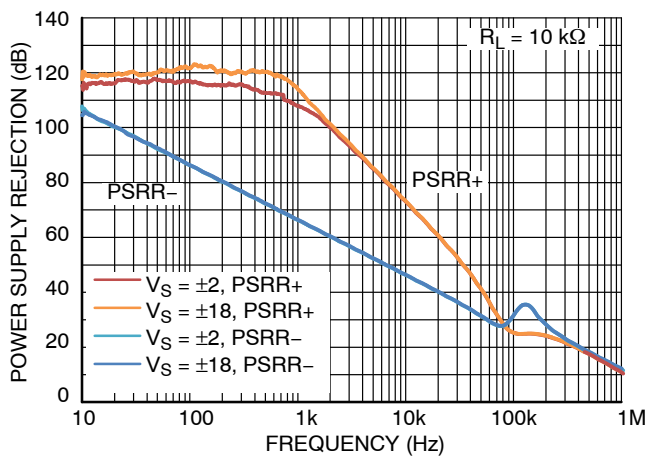


Figure 11. PSRR vs. Frequency

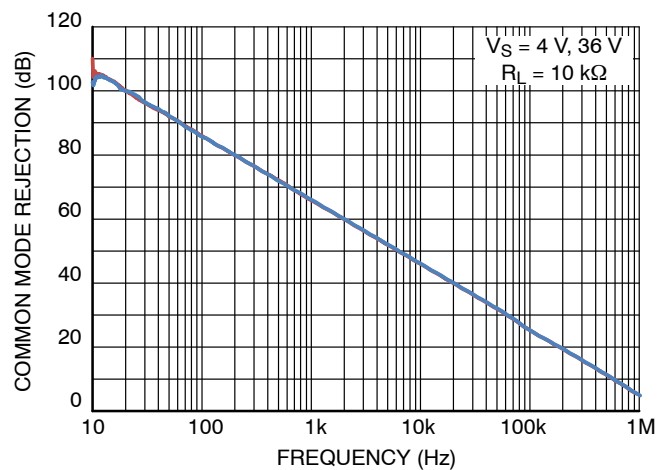


Figure 12. CMRR vs. Frequency

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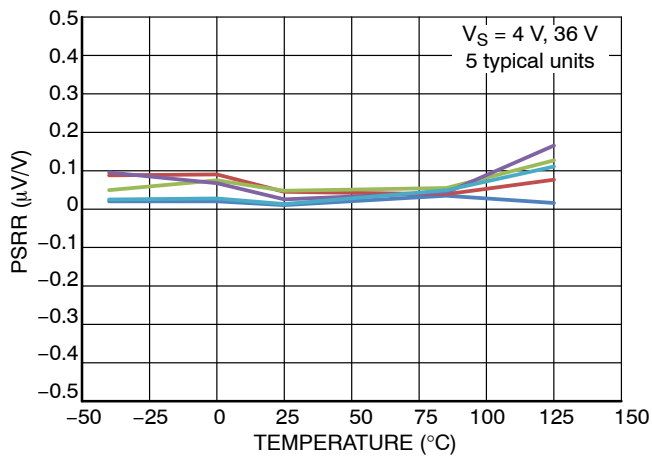


Figure 13. PSRR vs. Temperature

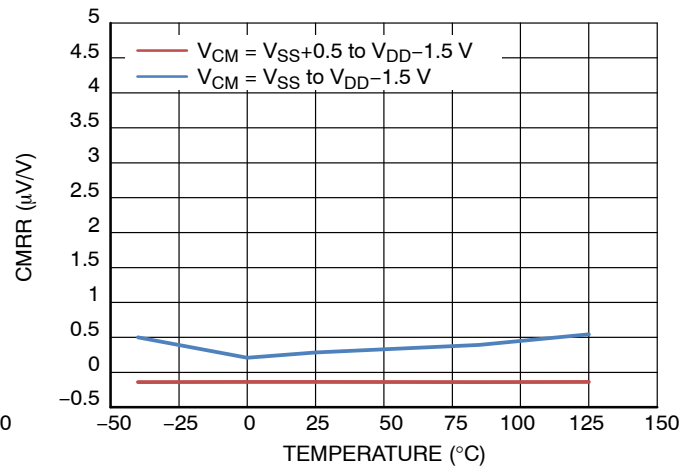


Figure 14. CMRR vs. Temperature at $V_S = 4\text{ V}$

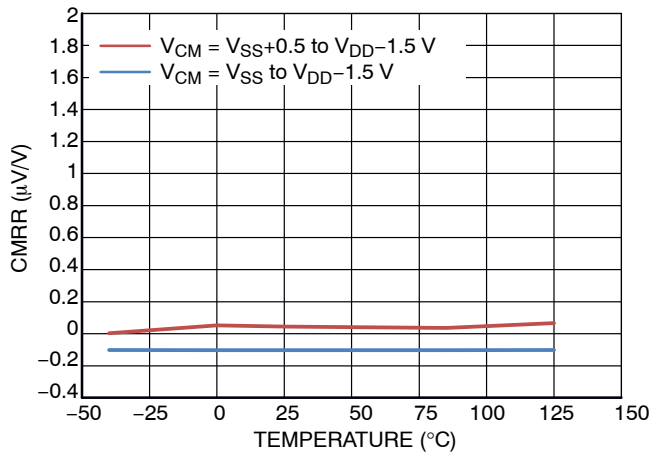


Figure 15. CMRR vs. Temperature at $V_S = 36\text{ V}$

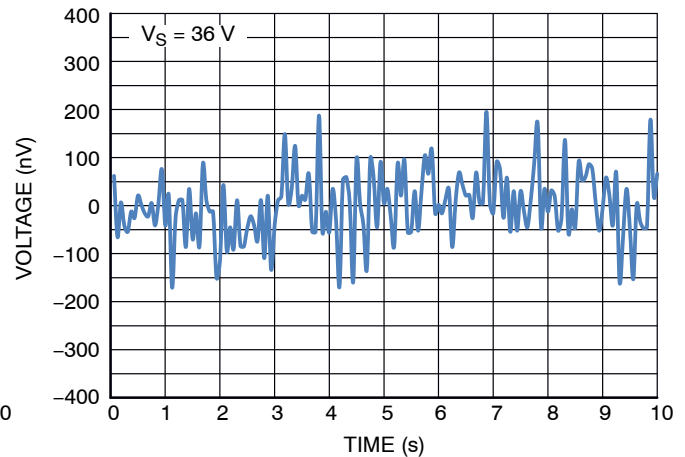


Figure 16. 0.1 Hz to 10 Hz Noise

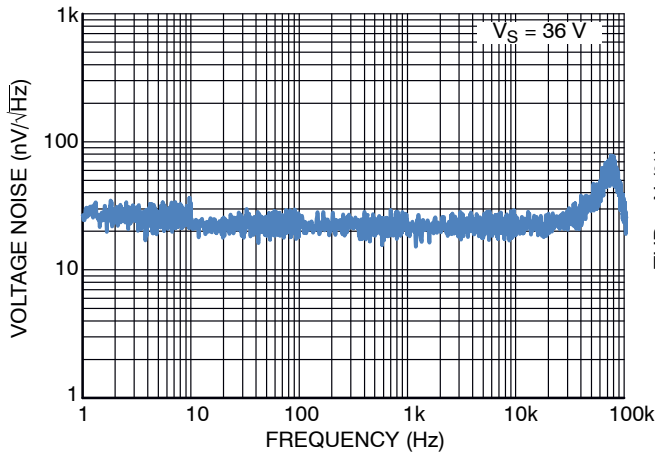


Figure 17. Voltage Noise Density vs. Frequency

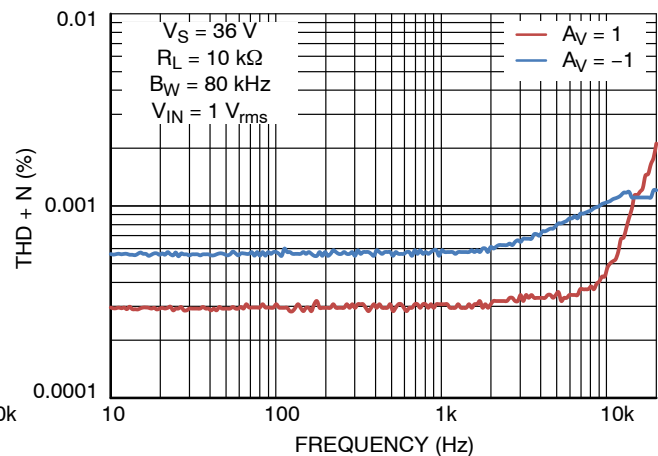
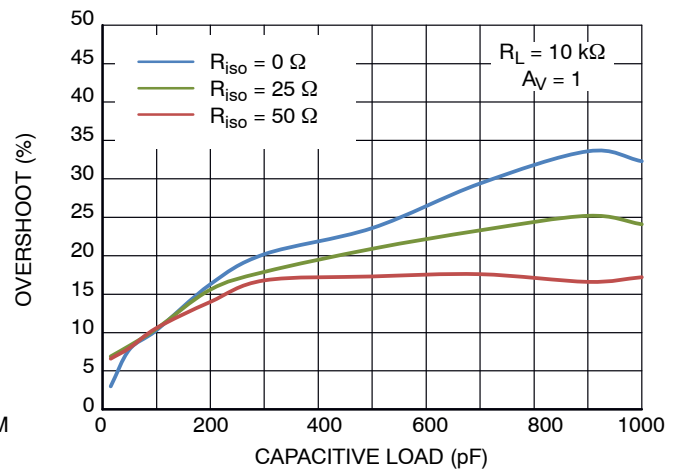
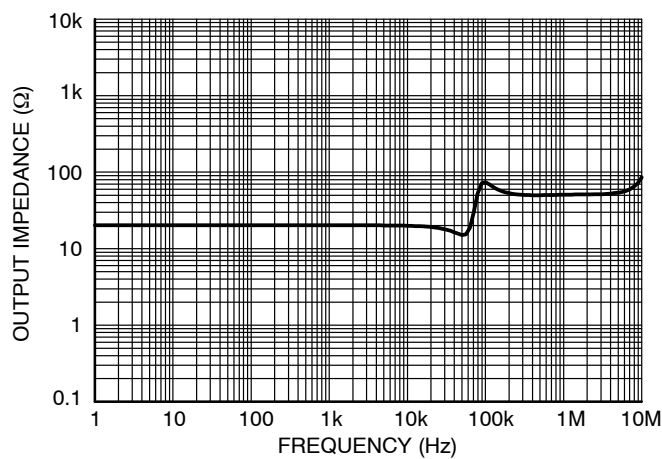
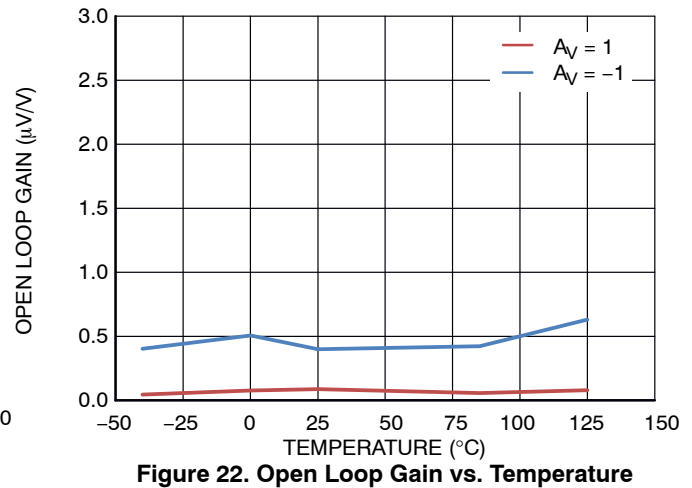
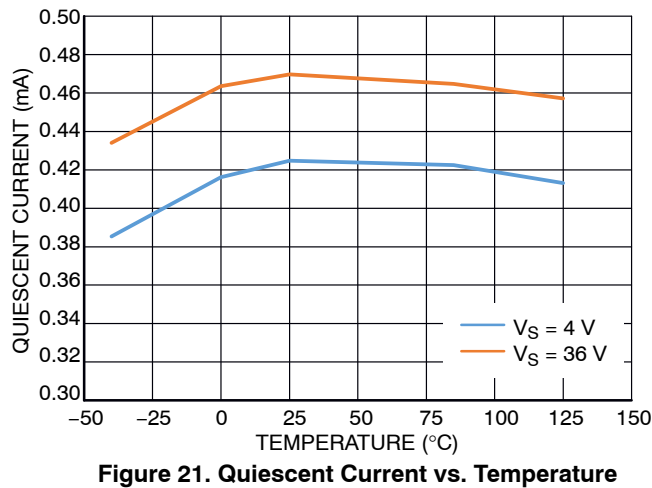
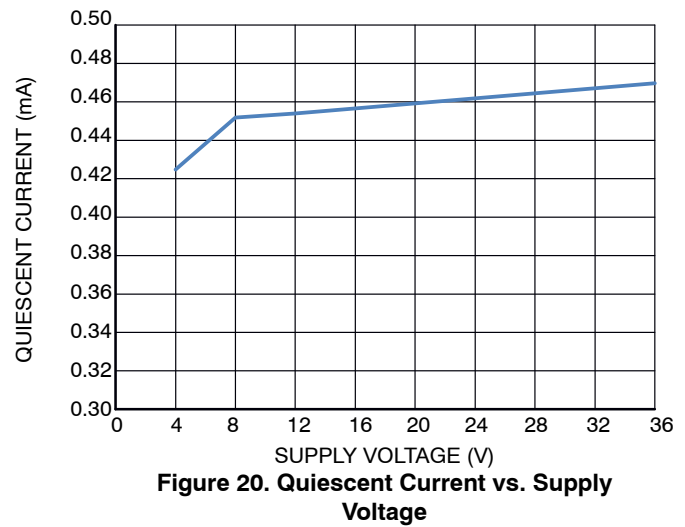
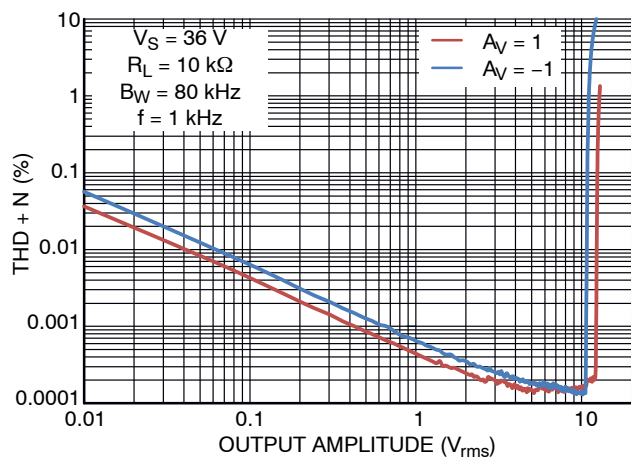


Figure 18. THD+N vs. Frequency

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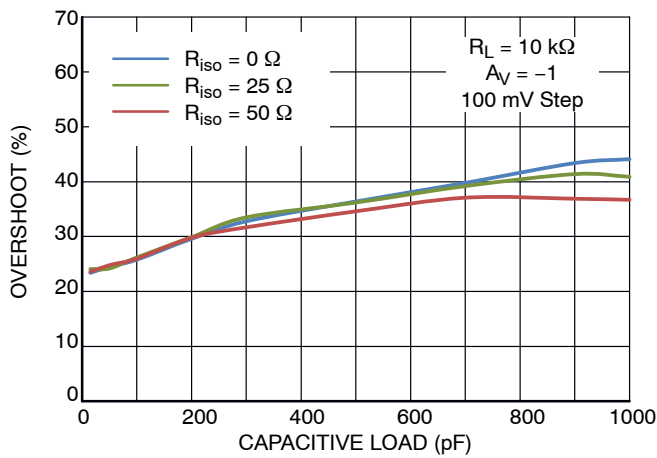


Figure 25. Small Signal Overshoot vs. Capacitive Load (100 mV Output Step)

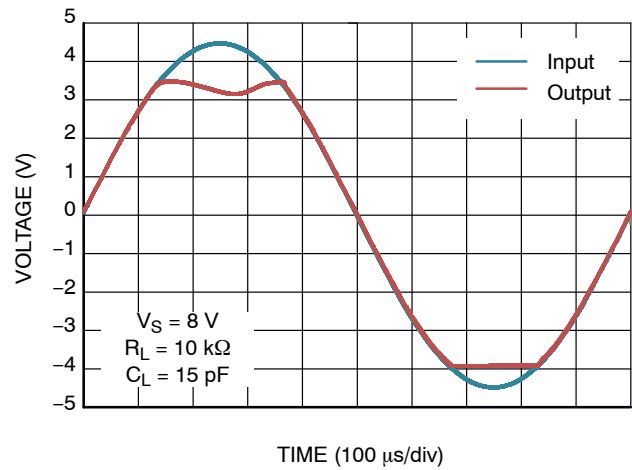


Figure 26. No Phase Reversal

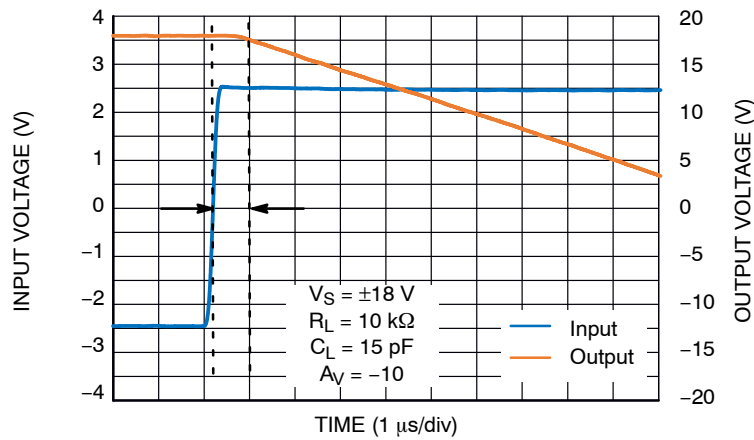


Figure 27. Positive Overload Recovery

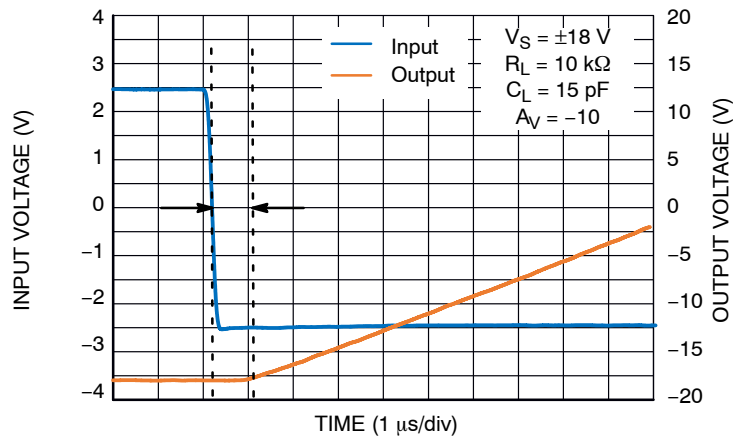


Figure 28. Negative Overload Recovery

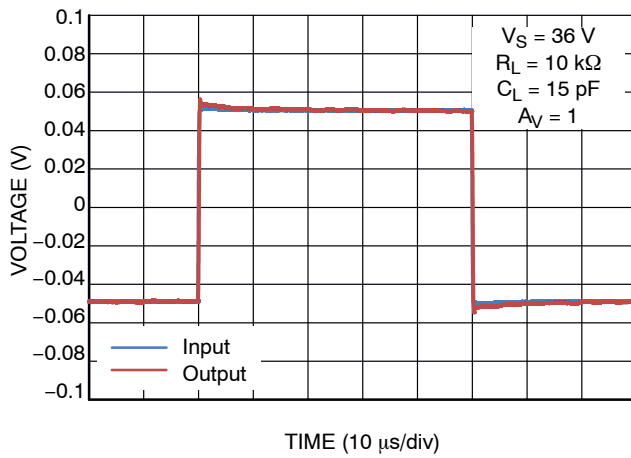


Figure 29. Non-Inverting Small Signal Step Response

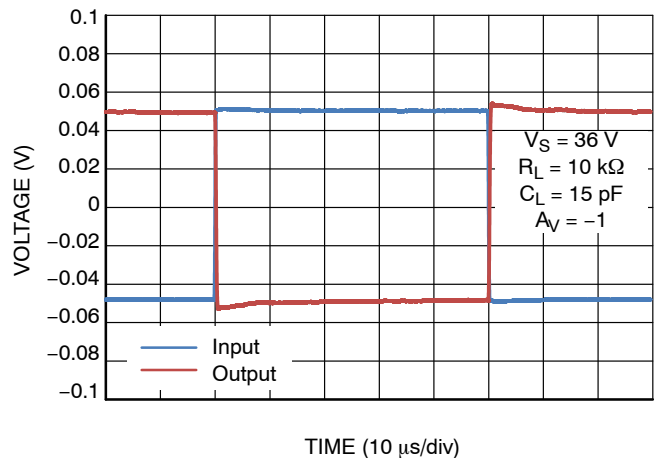


Figure 30. Inverting Small Signal Step Response

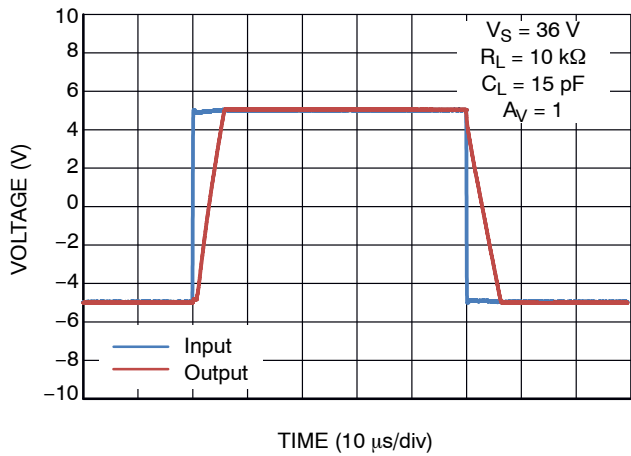


Figure 31. Non-Inverting Large Signal Step Response

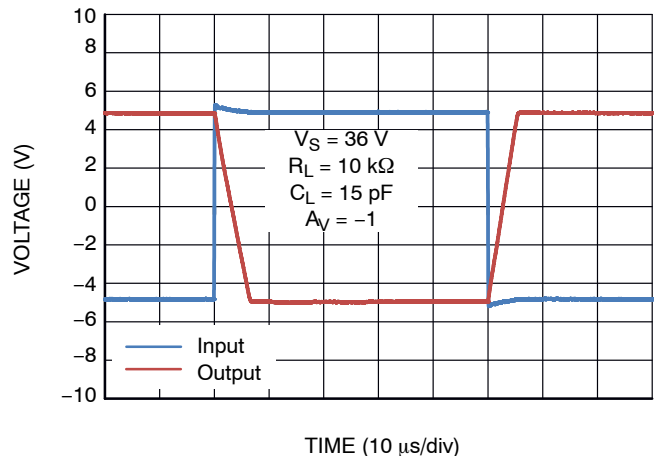


Figure 32. Inverting Large Signal Step Response

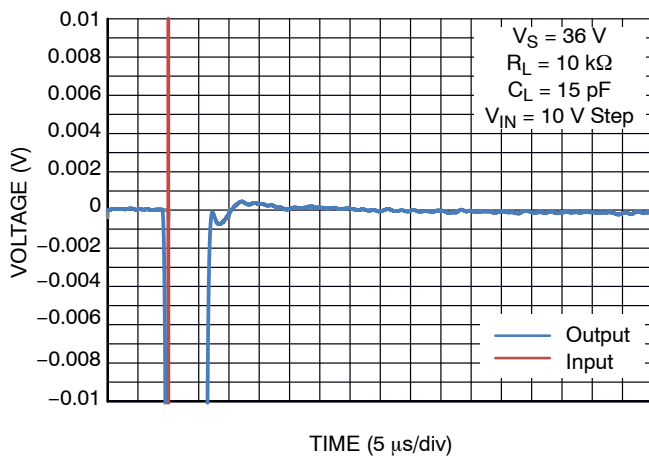


Figure 33. Large Signal Settling Time, Low-to-High

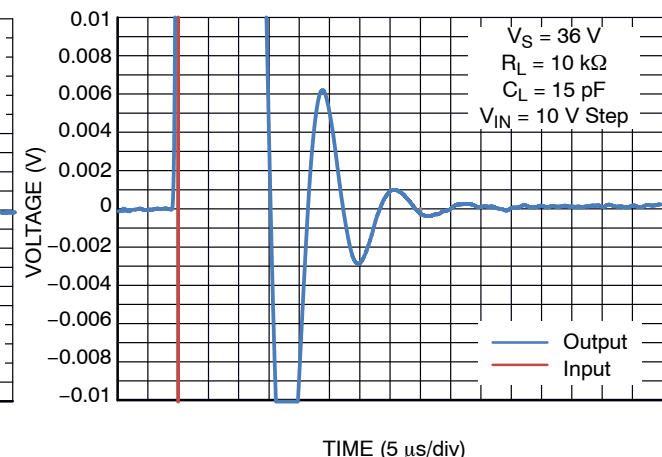


Figure 34. Large Signal Settling Time, High-to-Low

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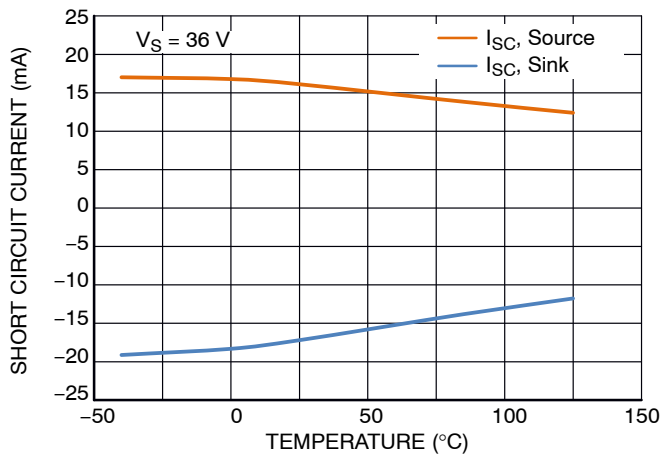


Figure 35. Short Circuit Current vs. Temperature

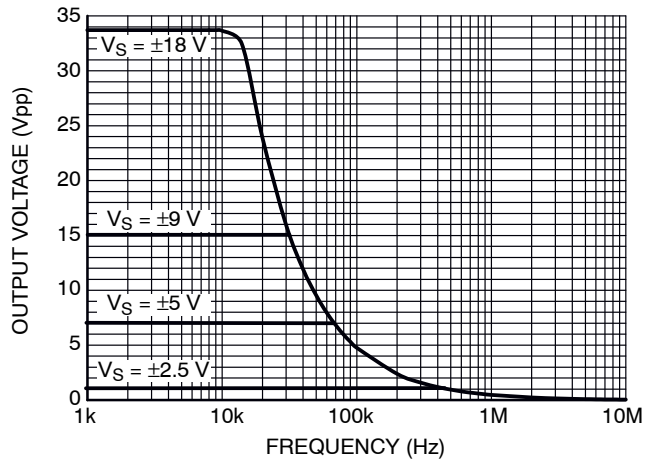


Figure 36. Maximum Output Voltage vs. Frequency ($A_V = 1$ for $V_S = \pm 2.5 \text{ V}$, $\pm 5 \text{ V}$, $\pm 9 \text{ V}$; $A_V = 2$ for $V_S = \pm 18 \text{ V}$)

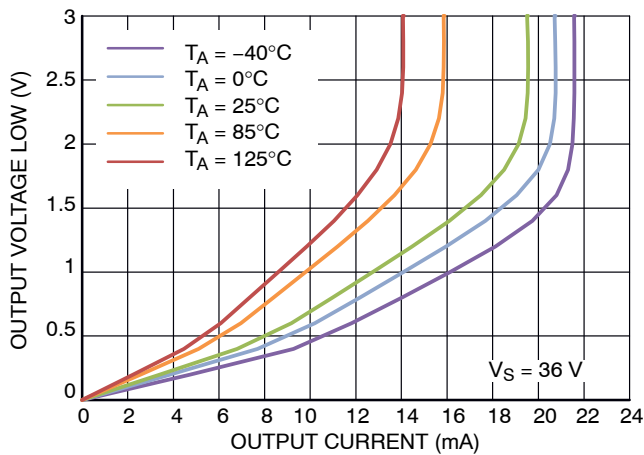


Figure 37. Output Voltage Low vs. Output Current

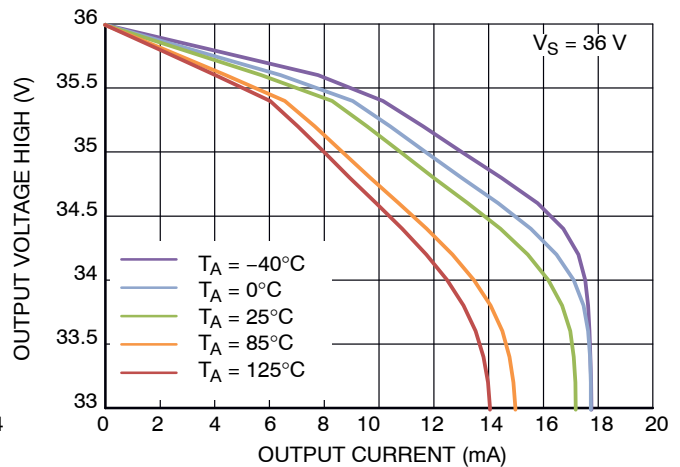


Figure 38. Output Voltage High vs. Output Current

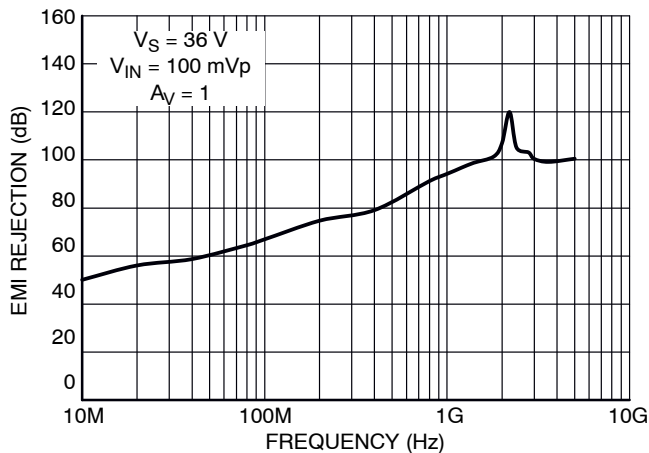


Figure 39. EMIRR IN+ vs. Frequency

APPLICATION INFORMATION

Overview

The NCS21911, NCS21912, and NCS21914 precision op amps provide low offset voltage and zero drift over temperature. With a maximum offset voltage of 25 μV and input common mode voltage range that includes ground, the NCS21911 series is well-suited for applications where precision is required, such as low side current sensing and interfacing with sensors.

The NCS21911 series of amplifiers uses a chopper-stabilized architecture, which provides the advantage of minimizing offset voltage drift over temperature and time. The simplified block diagram is shown in Figure 40. Unlike the classical chopper architecture, the chopper stabilized architecture has two signal paths.

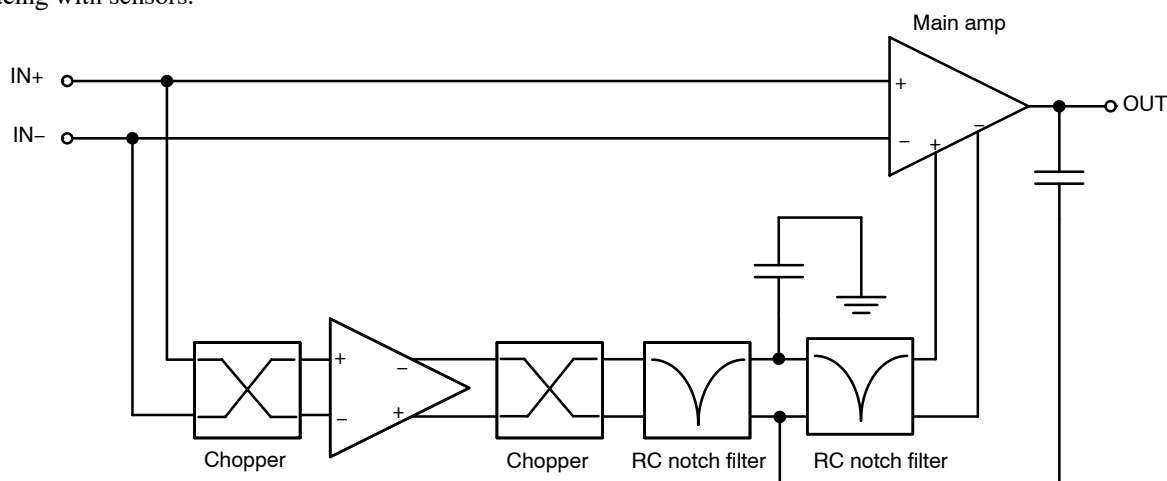


Figure 40. Simplified NCS21911 Block Diagram

In Figure 40, the lower signal path is where the chopper samples the input offset voltage, which is then used to correct the offset at the output. The offset correction occurs at a frequency of 250 kHz. The chopper-stabilized architecture is optimized for best performance at frequencies up to the related Nyquist frequency (1/2 of the offset correction frequency). As the signal frequency exceeds the Nyquist frequency, 125 kHz, aliasing may occur at the output. This is an inherent limitation of all chopper and chopper-stabilized architectures. Nevertheless, the NCS21911 series op amps have minimal aliasing up to 200 kHz and are less susceptible to aliasing effects when compared to competitor parts from other manufacturers. ON Semiconductor's patented approach utilizes two cascaded, symmetrical, RC notch filters tuned to the chopper frequency and its fifth harmonic to reduce aliasing effects.

The chopper-stabilized architecture also benefits from the feed-forward path, which is shown as the upper signal path of the block diagram in Figure 40. This is the high speed signal path that extends the gain bandwidth up to 2 MHz. Not

only does this help retain high frequency components of the input signal, but it also improves the loop gain at low frequencies. This is especially useful for low-side current sensing and sensor interface applications where the signal is low frequency and the differential voltage is relatively small.

Application Circuits*Low-Side Current Sensing*

Low-side current sensing is used to monitor the current through a load. This method can be used to detect over-current conditions and is often used in feedback control, as shown in Figure 41. A sense resistor is placed in series with the load to ground. Typically, the value of the sense resistor is less than 100 m Ω to reduce power loss across the resistor. The op amp amplifies the voltage drop across the sense resistor with a gain set by external resistors R1, R2, R3, and R4 (where R1 = R2, R3 = R4). Precision resistors are required for high accuracy, and the gain is set to utilize the full scale of the ADC for the highest resolution.

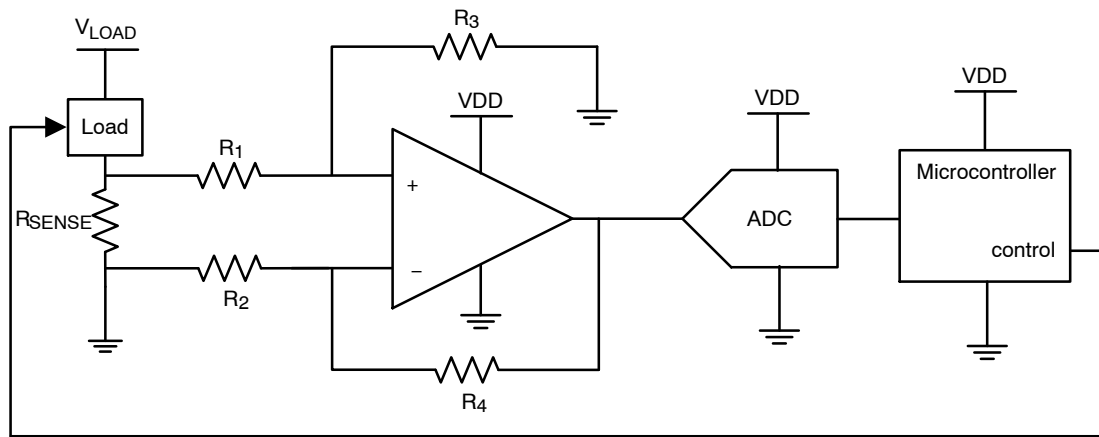


Figure 41. Low-Side Current Sensing

Differential Amplifier for Bridged Circuits

Sensors to measure strain, pressure, and temperature are often configured in a Wheatstone bridge circuit as shown in Figure 42. In the measurement, the voltage change that is

produced is relatively small and needs to be amplified before going into an ADC. Precision amplifiers are recommended in these types of applications due to their high gain, low noise, and low offset voltage.

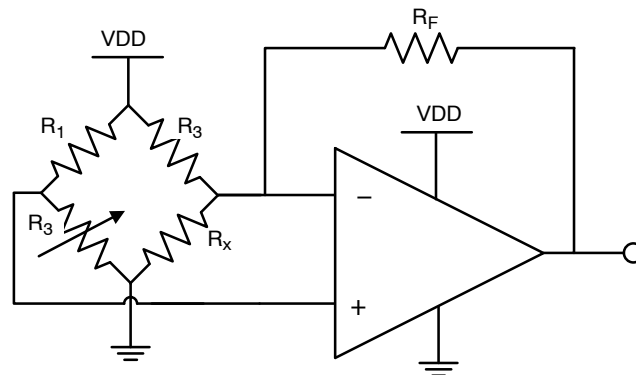


Figure 42. Wheatstone Bridge Circuit Amplification

EMI Susceptibility and Input Filtering

Op amps have varying amounts of EMI susceptibility. Semiconductor junctions can pick up and rectify EMI signals, creating an EMI-induced voltage offset at the output, adding another component to the total error. Input pins are the most sensitive to EMI. The NCS2191x integrates low-pass filters to decrease its sensitivity to EMI. Figure 39 shows the EMIRR performance.

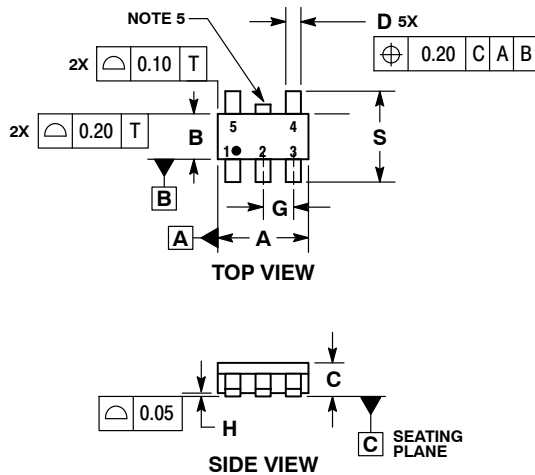
General Layout Guidelines

To ensure optimum device performance, it is important to follow good PCB design practices. Place 0.1 μ F decoupling

capacitors as close as possible to the supply pins. Keep traces short, utilize a ground plane, choose surface-mount components, and place components as close as possible to the device pins. These techniques will reduce susceptibility to electromagnetic interference (EMI). Thermoelectric effects can create an additional temperature dependent offset voltage at the input pins. To reduce these effects, use metals with low thermoelectric coefficients and prevent temperature gradients from heat sources or cooling fans.

PACKAGE DIMENSIONS

TSOP-5
CASE 483
ISSUE M

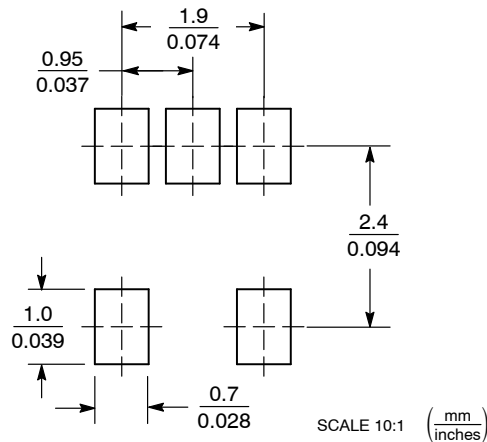


NOTES:

1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: MILLIMETERS.
3. MAXIMUM LEAD THICKNESS INCLUDES LEAD FINISH THICKNESS. MINIMUM LEAD THICKNESS IS THE MINIMUM THICKNESS OF BASE MATERIAL.
4. DIMENSIONS A AND B DO NOT INCLUDE MOLD FLASH, PROTRUSIONS, OR GATE BURRS. MOLD FLASH, PROTRUSIONS, OR GATE BURRS SHALL NOT EXCEED 0.15 PER SIDE. DIMENSION A.
5. OPTIONAL CONSTRUCTION: AN ADDITIONAL TRIMMED LEAD IS ALLOWED IN THIS LOCATION. TRIMMED LEAD NOT TO EXTEND MORE THAN 0.2 FROM BODY.

DIM	MILLIMETERS	
	MIN	MAX
A	2.85	3.15
B	1.35	1.65
C	0.90	1.10
D	0.25	0.50
G	0.95 BSC	
H	0.01	0.10
J	0.10	0.26
K	0.20	0.60
M	0°	10°
S	2.50	3.00

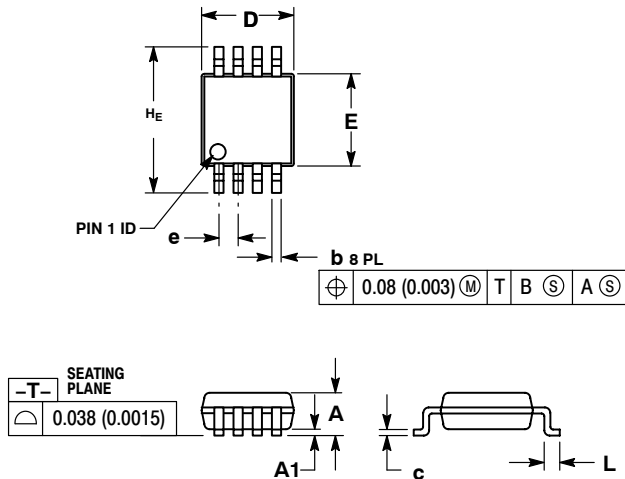
SOLDERING FOOTPRINT*



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PACKAGE DIMENSIONS

Micro8 CASE 846A-02 ISSUE J

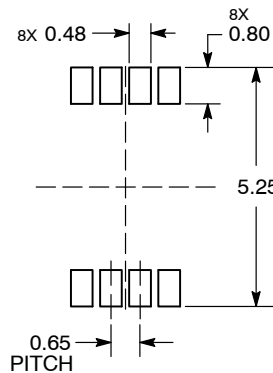


NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSION A DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.15 (0.006) PER SIDE.
4. DIMENSION B DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25 (0.010) PER SIDE.
5. 846A-01 OBSOLETE, NEW STANDARD 846A-02.

DIM	MILLIMETERS			INCHES		
	MIN	NOM	MAX	MIN	NOM	MAX
A	---	---	1.10	---	---	0.043
A1	0.05	0.08	0.15	0.002	0.003	0.006
b	0.25	0.33	0.40	0.010	0.013	0.016
c	0.13	0.18	0.23	0.005	0.007	0.009
D	2.90	3.00	3.10	0.114	0.118	0.122
E	2.90	3.00	3.10	0.114	0.118	0.122
e	0.65 BSC			0.026 BSC		
L	0.40	0.55	0.70	0.016	0.021	0.028
HE	4.75	4.90	5.05	0.187	0.193	0.199

RECOMMENDED SOLDERING FOOTPRINT*

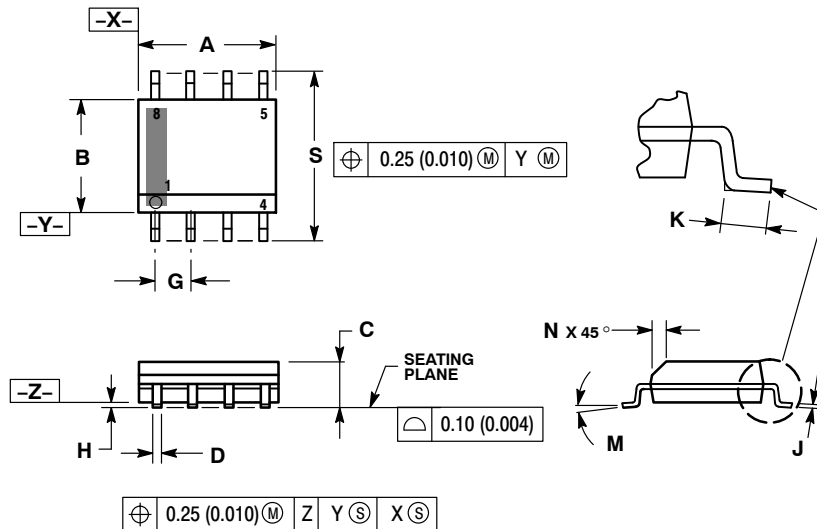


DIMENSION: MILLIMETERS

*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

PACKAGE DIMENSIONS

SOIC-8 NB
CASE 751-07
ISSUE AK

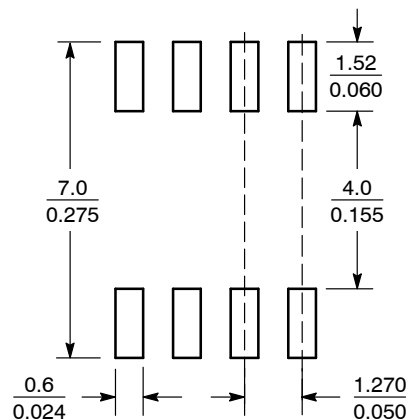


NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSION A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.
6. 751-01 THRU 751-06 ARE OBSOLETE. NEW STANDARD IS 751-07.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	4.80	5.00	0.189	0.197
B	3.80	4.00	0.150	0.157
C	1.35	1.75	0.053	0.069
D	0.33	0.51	0.013	0.020
G	1.27 BSC		0.050 BSC	
H	0.10	0.25	0.004	0.010
J	0.19	0.25	0.007	0.010
K	0.40	1.27	0.016	0.050
M	0°	8°	0°	8°
N	0.25	0.50	0.010	0.020
S	5.80	6.20	0.228	0.244

SOLDERING FOOTPRINT*

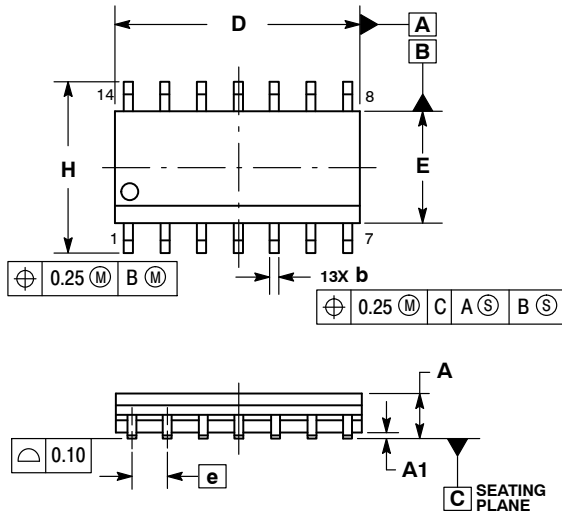


SCALE 6:1 (mm/inches)

*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

PACKAGE DIMENSIONS

SOIC-14 NB
CASE 751A-03
ISSUE L

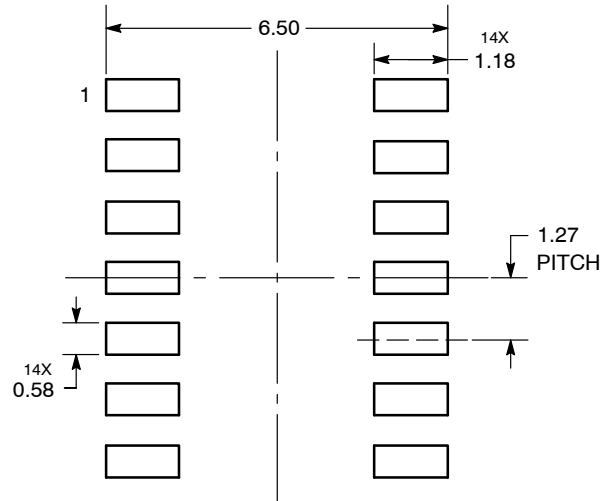


NOTES:

1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: MILLIMETERS.
3. DIMENSION b DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE PROTRUSION SHALL BE 0.13 TOTAL IN EXCESS OF AT MAXIMUM MATERIAL CONDITION.
4. DIMENSIONS D AND E DO NOT INCLUDE MOLD PROTRUSIONS.
5. MAXIMUM MOLD PROTRUSION 0.15 PER SIDE.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	1.35	1.75	0.054	0.068
A1	0.10	0.25	0.004	0.010
A3	0.19	0.25	0.008	0.010
b	0.35	0.49	0.014	0.019
D	8.55	8.75	0.337	0.344
E	3.80	4.00	0.150	0.157
e	1.27 BSC		0.050 BSC	
H	5.80	6.20	0.228	0.244
h	0.25	0.50	0.010	0.019
L	0.40	1.25	0.016	0.049
M	0°	7°	0°	7°

SOLDERING FOOTPRINT*

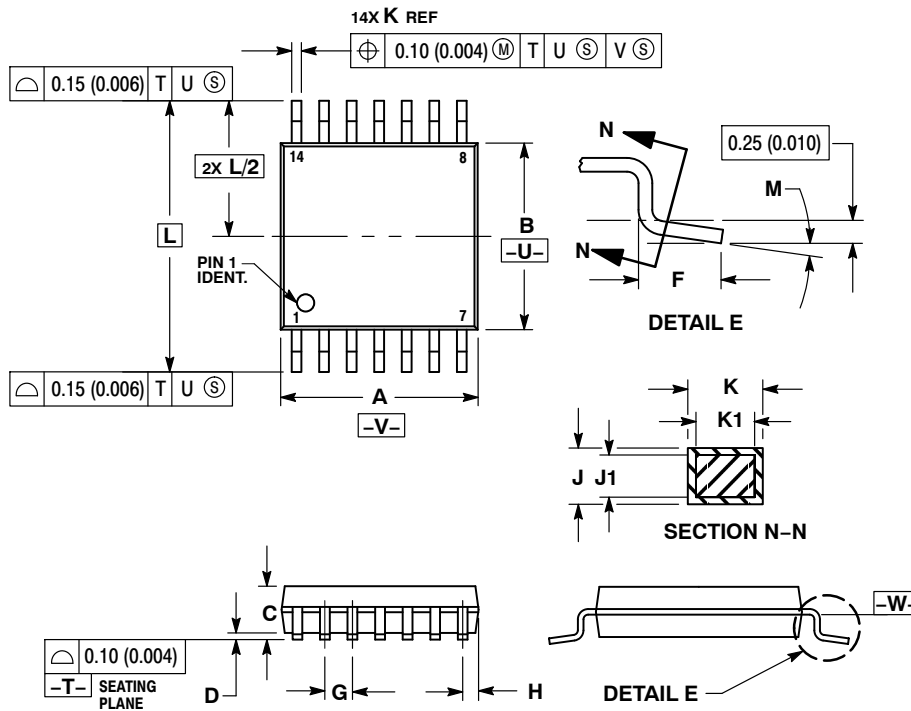


DIMENSIONS: MILLIMETERS

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PACKAGE DIMENSIONS

TSSOP-14 WB
CASE 948G
ISSUE C

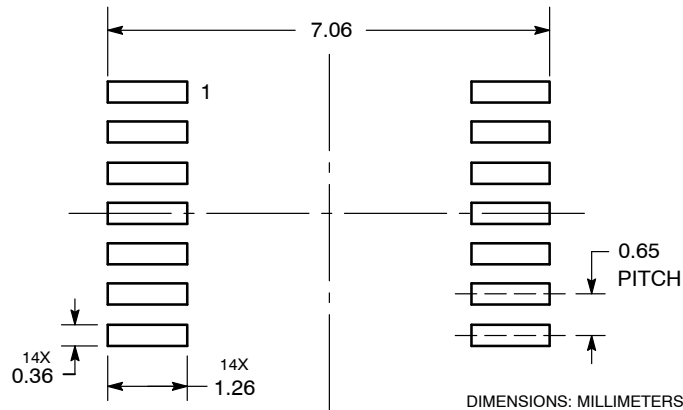



NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSION A DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH OR GATE BURRS SHALL NOT EXCEED 0.15 (0.006) PER SIDE.
4. DIMENSION B DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25 (0.010) PER SIDE.
5. DIMENSION K DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 (0.003) TOTAL IN EXCESS OF THE K DIMENSION AT MAXIMUM MATERIAL CONDITION.
6. TERMINAL NUMBERS ARE SHOWN FOR REFERENCE ONLY.
7. DIMENSION A AND B ARE TO BE DETERMINED AT DATUM PLANE -W-.

	MILLIMETERS		INCHES	
DIM	MIN	MAX	MIN	MAX
A	4.90	5.10	0.193	0.200
B	4.30	4.50	0.169	0.177
C	---	1.20	---	0.047
D	0.05	0.15	0.002	0.006
F	0.50	0.75	0.020	0.030
G	0.65 BSC		0.026 BSC	
H	0.50	0.60	0.020	0.024
J	0.09	0.20	0.004	0.008
J1	0.09	0.16	0.004	0.006
K	0.19	0.30	0.007	0.012
K1	0.19	0.25	0.007	0.010
L	6.40 BSC		0.252 BSC	
M	0° 8°		0° 8°	

SOLDERING FOOTPRINT



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