

Design of Compact UWB Multilayered Microstrip Filter with Wide Stopband

N. Azadi-Tinat, H. Oraizi

Abstract—Design of compact UWB multilayered microstrip filter with E-shape resonator is presented, which provides wide stopband up to 20 GHz and arbitrary impedance matching. The design procedure is developed based on the method of least squares and theory of N-coupled transmission lines. The dimensions of designed filter are about 11 mm × 11 mm and the three E-shape resonators are placed among four dielectric layers. The average insertion loss in the passband is less than 1 dB and in the stopband is about 30 dB up to 20 GHz. Its group delay in the UWB region is about 0.5 ns. The performance of the optimized filter design perfectly agrees with the microwave simulation softwares.

Keywords—Ultra-wideband, method of least square, multilayer microstrip filter, n-coupled transmission lines.

I. INTRODUCTION

IN 2002, the USA Federal Communications Commission (FCC) allocated the frequency band 3.1-10.6 GHz to the UWB commercial applications [1]. The UWB technology has since been applied to various communication and radar systems and many efforts have been made in the development and realization of various UWB configurations and structures [2], [3].

Microstrip filters due to low weight and cost and ease of manufacture are suitable choices to use in many microwave circuits. Many planar microstrip filters use coupled lines to achieve a desired coupling, such as combline, interdigital and hairpin filters [4], [5]. But the planar structure of these types of filters in the wideband and UWB applications make the edges of microstrip lines too close and in some cases they may not be realizable by the available technology, such as photolithography. In such cases, multilayered structures of microstrip filters are preferable, because they provide additional degrees of freedom for the filter design.

In this paper, a four layered E-shape microstrip filter for UWB applications is introduced. Fig. 1 (a) shows the configuration of proposed filter and for a better view, the heights of layers are scaled. Fig. 1 (b) shows the side view of filter and the location of vias. Fig. 1 (c) shows each layer separately and shows that layers 2 and 3 have been connected with one via passing through the ground plane. Such a structure is actually the advantage of this design, which provides for the precise location of resonators above each

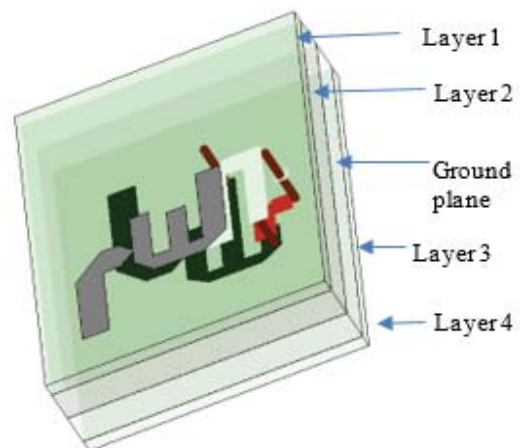
other for miniaturization. The E-shaped structure of resonators provides the required coupling to obtain the UWB bandwidth. Fig. 2 shows the effect of height on the two coupled lines. The ratio of h_1/h_2 is chosen less than one in the follow-up design to achieve better couplings. The short and open stubs in the middle of resonators could produce transmission zeroes to suppress the spurious responses and obtain a very wide stopband up to 20 GHz. Use of the short circuit stubs for the lower degree filters to provide transmission zeroes in the lower stopband is necessary, but it has the disadvantage of greater lengths of stubs compared to open circuit stubs (Fig. 3)

The design procedure is based on the method of least squares and the theory of N-coupled transmission lines which determines the optimum filter geometrical dimensions.

II. DESIGN PROCEDURE

The four-layer-three-E-shape resonator microstrip filter configuration is depicted in Fig. 1. Its equivalent circuit is shown in Fig. 4, which is composed of straight strips, bends, T-junctions and coupled lines. For each section the transmission matrix is developed [6] and used to obtain overall transmission matrix

$$[T] = [T_{s2}][T_{s1}][T_{m1}][T_{c1}][T_{m2}][T_{c2}][T_{m3}][T_{r1}][T_{r2}] \quad (1)$$



(a)



(b)

N. Azadi-Tinat is with the Electrical Engineering Department, Shahrood University of Technology, Islamic Republic of Iran (e-mail: azadi@shahroodut.ac.ir).

H. Oraizi is with the Electrical Engineering Department, University of Science and Technology, Islamic Republic of Iran (e-mail: h_orazi@iust.ac.ir).

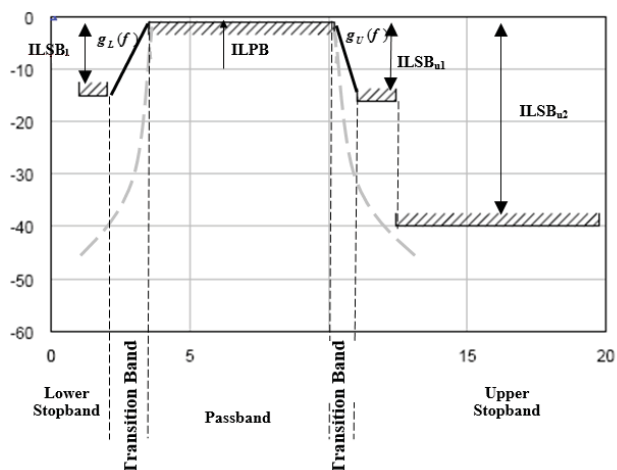


Fig. 5 Specified frequency response of the bandpass filter

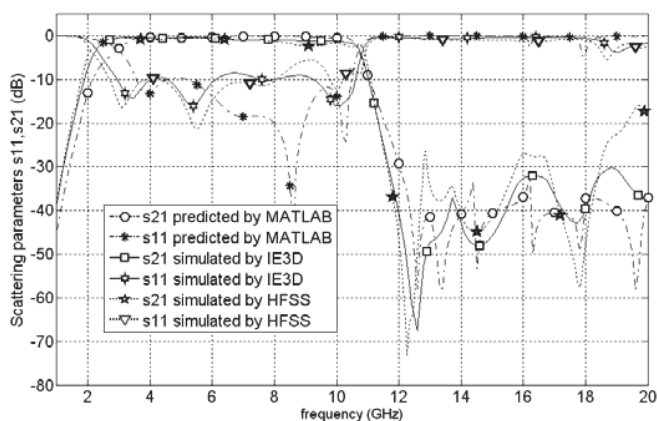


Fig. 6 Comparison of the frequency responses of proposed multilayered microstrip filter as obtained by the simulation softwares (IE3D & HFSS) and predicted by MATLAB

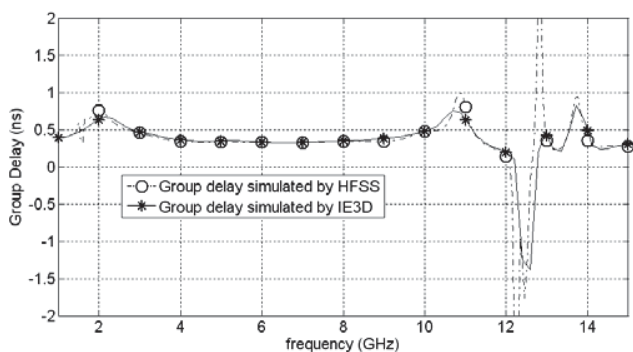


Fig. 7 Group delayed of proposed multilayered microstrip filter.

The specified frequency band is divided into K discrete frequencies, $ILSB$ and $ILPB$ are the desired insertion losses in the stop and pass bands, and $g_{TL}(f_k)$ and $g_{TU}(f_k)$ are the linear functions in the lower and upper transition bands, respectively which join the stop and pass bands, and w_{ti} are the weighting functions which enhance the effect of different bands relative to each other. The minimization of the error function gives the physical dimensions of the filter.

A computer program was written in MATLAB [7] for the optimum design of multilayered E-shaped microstrip filter based on the minimization of error function in (6). Its input parameters are: filter bandwidth, the design specifications on the insertion losses in the passband, stopbands and transition bands, number of discrete frequencies in the bandwidth and other appropriate parameters required in the error function.

The initial values of the geometrical dimensions of the filter, such as the widths, lengths and spacings of the strip line sections, are selected as follows: The width of all the strip lines are taken equal to 1 mm. They may also be selected equal to the total height of substrates. The displacement or the offset between coupled lines is taken equal to zero and remain zero in the design procedure. The initial lengths of open and short stubs may be taken equal to a quarter and half wavelengths at the center frequency, respectively.

III. DESIGN EXAMPLE

Consider four substrate layers with three E-shaped resonators. The layers one and four are made of substrates RO4003 with height of 8 mil and the layers two and three are made of RO5880 with height of 20 mil. The aforementioned related parameters and design data are given in Table I. The frequency responses of the filter (as s_{11} and s_{21}) obtained by the computer simulation results and full-wave software such as IE3D and HFSS [8], [9] are shown in Fig. 7. Observe that the suppression of spurious harmonics is about 30 dB. The group delay of filter is shown in Fig. 7, which is quite constant in the specified bandwidth 3.1-10.6 GHz and is less than 0.5 ns. For the proof of concept, a prototype model is fabricated and its photograph is shown in Fig. 8.

TABLE I
PARAMETERS OF DESIGNED MULTILAYERED MICROSTRIP FILTER

<i>parameters of substrates</i>	
substrate 1:	RO 4003, 8mil
substrate 2:	RO 5880, 20mil
substrate 3:	RO 5880, 20mil
substrate 4:	RO 4003, 8mil
Widths of input and output 50 Ω feedline 2 mm	
Dimension of resonators (all mm) (Fig .4)	
Input :	$s_2 = 2.95, s_1 = 1.06, w_s = 1.4$
Output :	$t_2 = 2.13, t_1 = 2.18, w_t = 1.6$
<i>Other dimensions:</i>	
$a_1 = 1.27, w_{a1} = 1.3, d_1 = 3.52, w_{d1} = 2.56$	
$l_1 = 11.92, w_{l1} = 1.98$	
$a_2 = 1.2, w_{a2} = 1.5, d_2 = 5.4, w_{d2} = 1.4$	
$l_2 = 1.9, w_{l2} = 1.6$	
$a_3 = 1.27, w_{a3} = 1, d_3 = 7.9, w_{d3} = 1.6$	

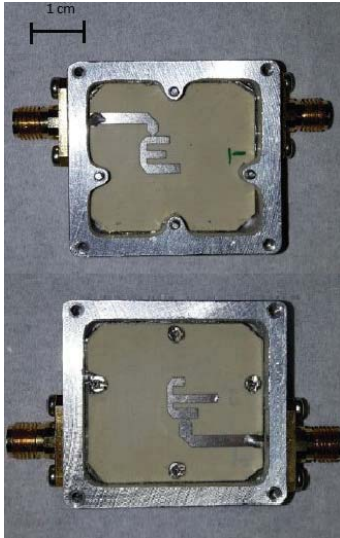


Fig. 8 The photograph of fabricated multilayered microstrip filter

IV. CONCLUSION

A compact multilayered E-shaped resonator microstrip bandpass filter with wide stopband is presented together with an optimum design procedure based on the theory of N-coupled transmission lines and method of least squares. The resonators are well-placed above each other in four layers providing a compact the size for the filter. The open and short stubs in the E-shaped resonators provide transmission zeroes in the upper and lower stopbands to create a wide stopband up to 20 GHz. The full-wave simulation and prototype fabrication of proposed filter has verified the results of proposed design method.

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Nima Azadi-Tinat was born in Mashhad, Iran, on March 14, 1980. He received B.S. degree from the Amirkabir University of Technology, Tehran and M.S and PhD degrees from the Iran University of Science and Technology (IUST), Tehran, Iran, all in electrical engineering, in 2003, 2005

and 2011 respectively. In 2011, he joined the Department of Electrical Engineering Shahrood University of Technology, Shahrood, Iran, where he is an assistant Professor of communications engineering.. His major research interests include passive microwave devices and antennas.

Prof. Homayoon Oraizi received B.E.E. degree from American University of Beirut, Lebanon, in 1967, M.Sc. and Ph.D. degrees in electrical engineering from Syracuse University, Syracuse, NY, in 1969 and 1973, respectively. From 1973 to 1974, he was at Toosi University of Technology, Tehran, Iran. From 1974 to 1985, he was with the Communications Division, Iran Electronics Industries, Shiraz, Iran. In 1985, he joined the Department of Electrical Engineering, Iran University of Science and Technology, Tehran, Iran, where he is a Full Professor of electromagnetic engineering. He has authored and translated several textbooks in Farsi. He has authored or coauthored over 300 papers in international journals and conferences. In 2006, he was elected an exemplary nation wide university professor in Iran. He is an Invited Professor of the Electrical Engineering Group, Academy of Sciences of Iran, and is listed as an elite engineer by the Iranology Foundation. Dr. Oraizi is a Fellow of Electromagnetic Academy, a fellow of Japan Society for the Promotion of Science and a life senior member of IEEE. He is listed in Who's Who in the World.