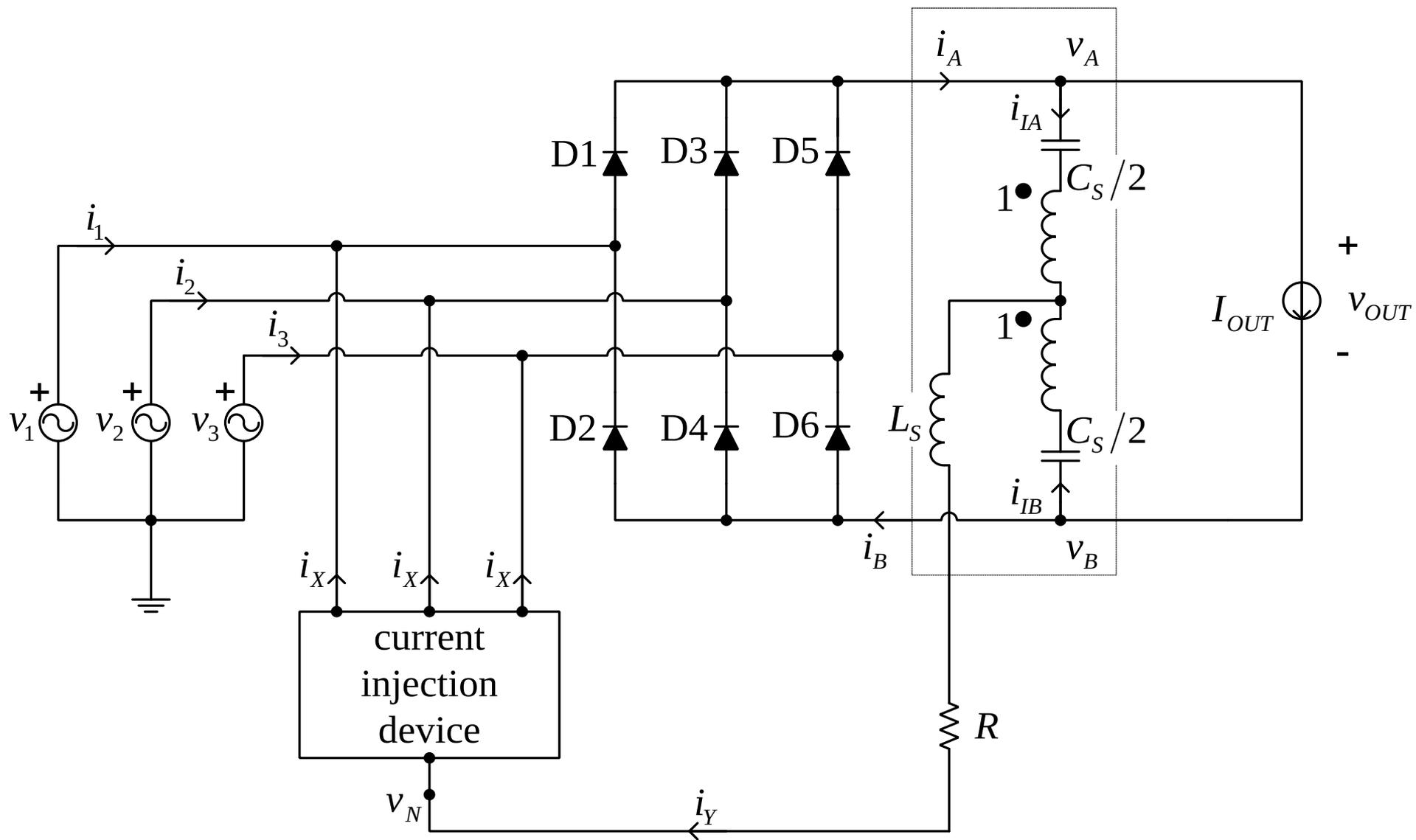


Computer-Aided Analysis of Three-Phase Rectifiers

Predrag Pejović
University of Belgrade

a problem:



Tek

Ω

Trig'd

M Pos: 4.800ms

CH4

Coupling

DC

Band Limit

On

20MHz

Volts/Div

Coarse

Probe

100X

Voltage

Invert

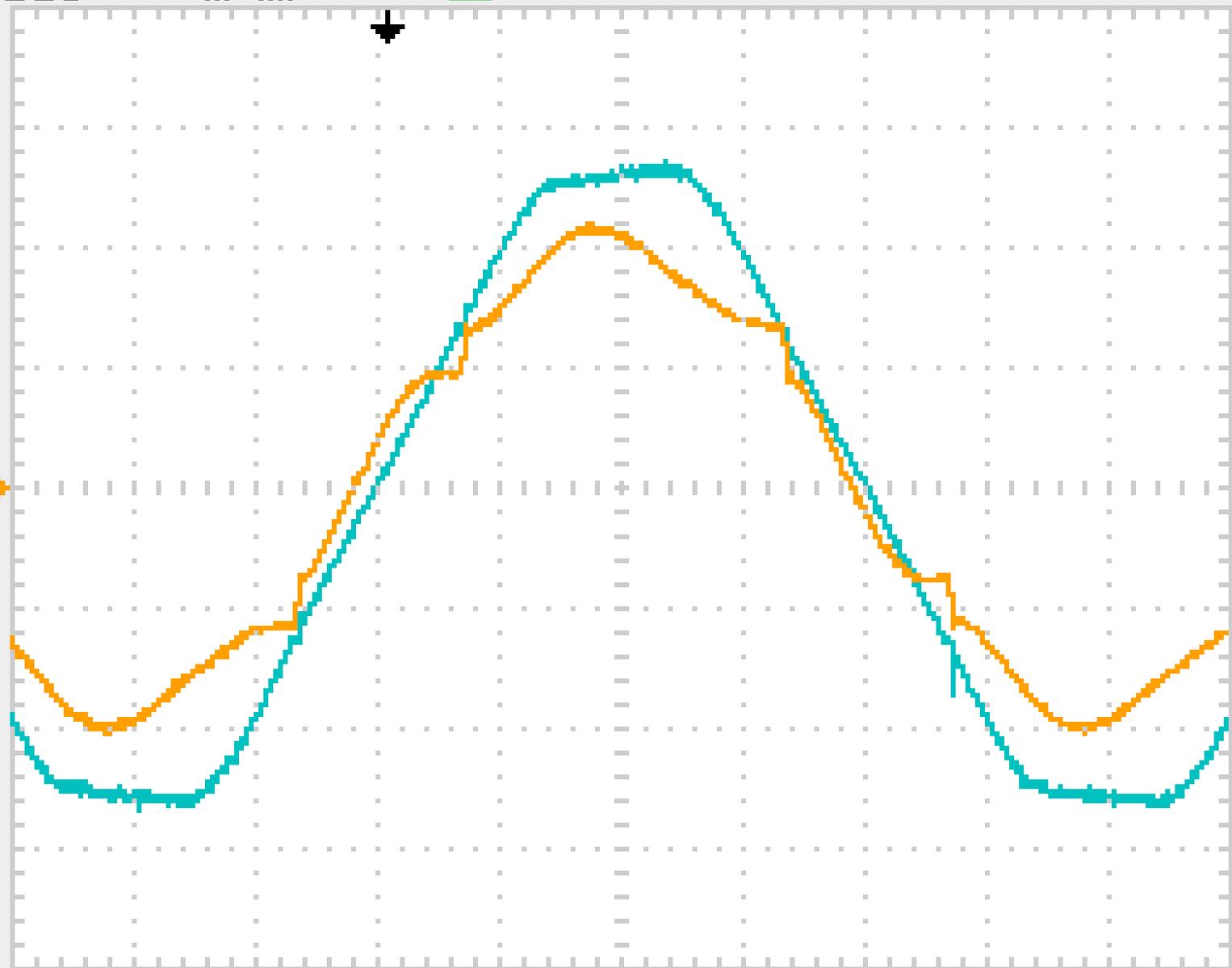
Off

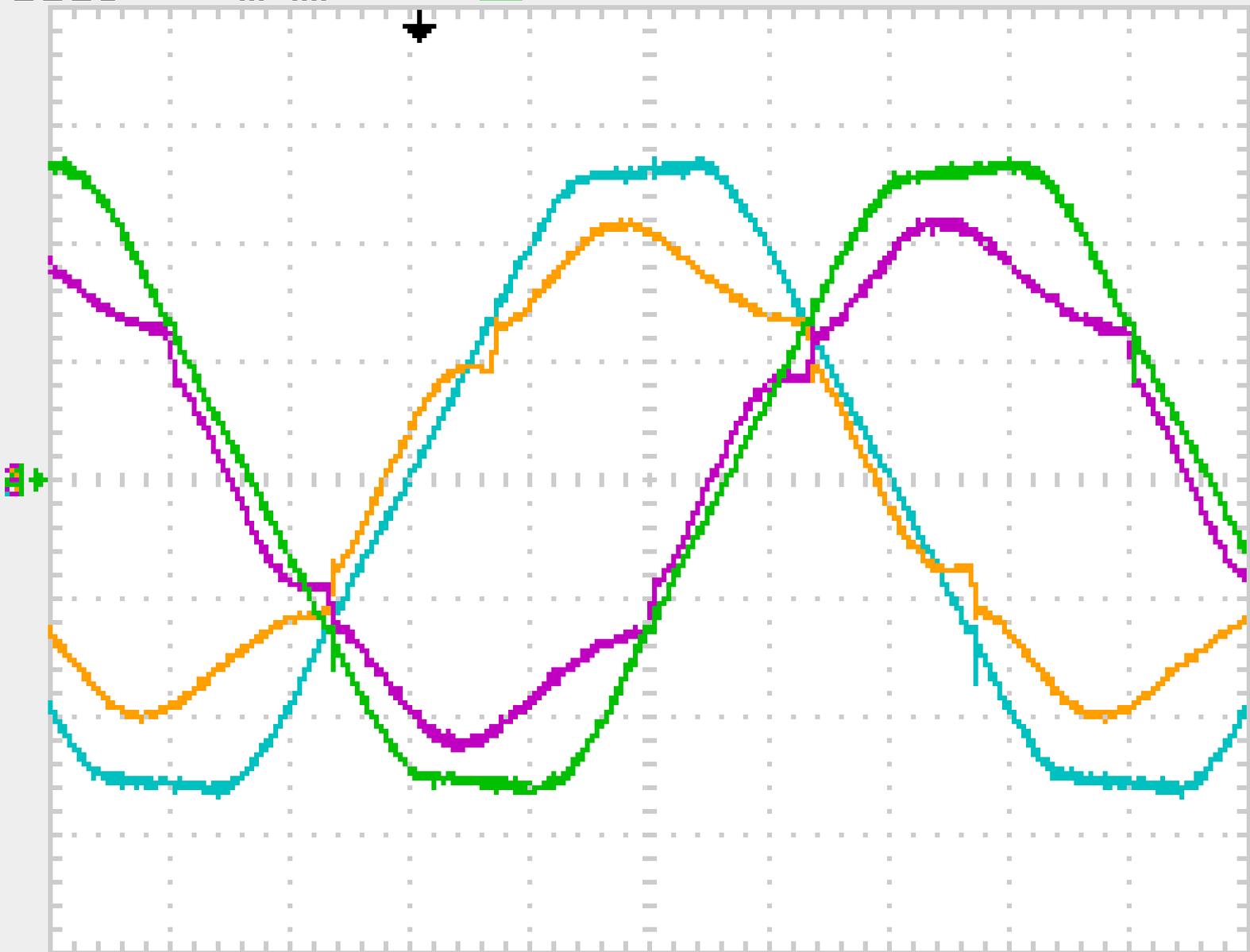
1+

CH1 5.00AB_w CH2 50.0VB_w M 2.50ms Ext / 200mV

12-Jun-08 20:17

49.9707Hz





Coupling
 DC

Band Limit
 On
20MHz

Volts/Div
 Coarse

Probe
100X
Voltage

Invert
 Off

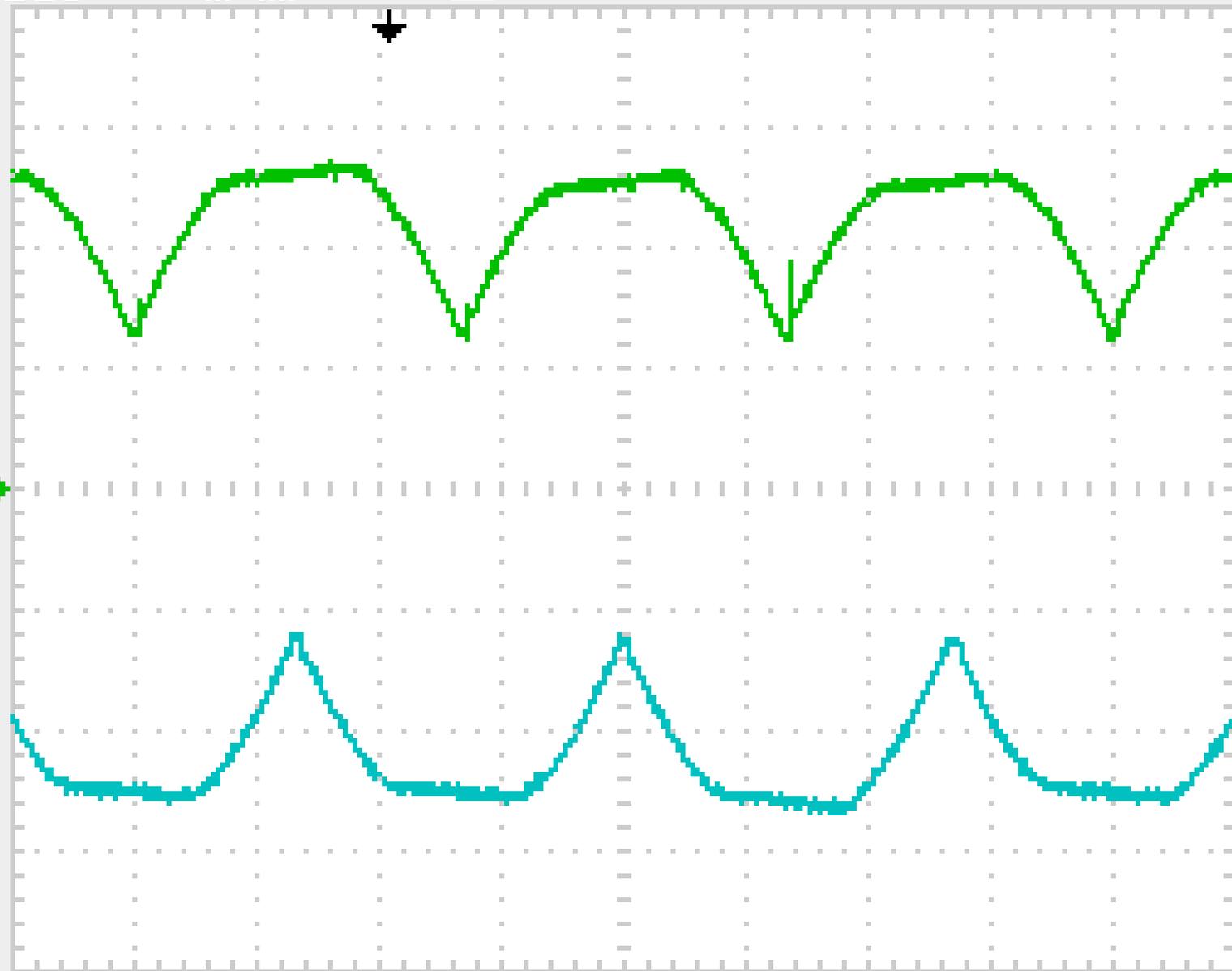
Tek

Ω

Trig'd

M Pos: 4.800ms

CH1



Coupling

DC

Band Limit

On

20MHz

Volts/Div

Coarse

Probe

100A/V

Current

Invert

Off

CH2 50.0V

M 2.50ms

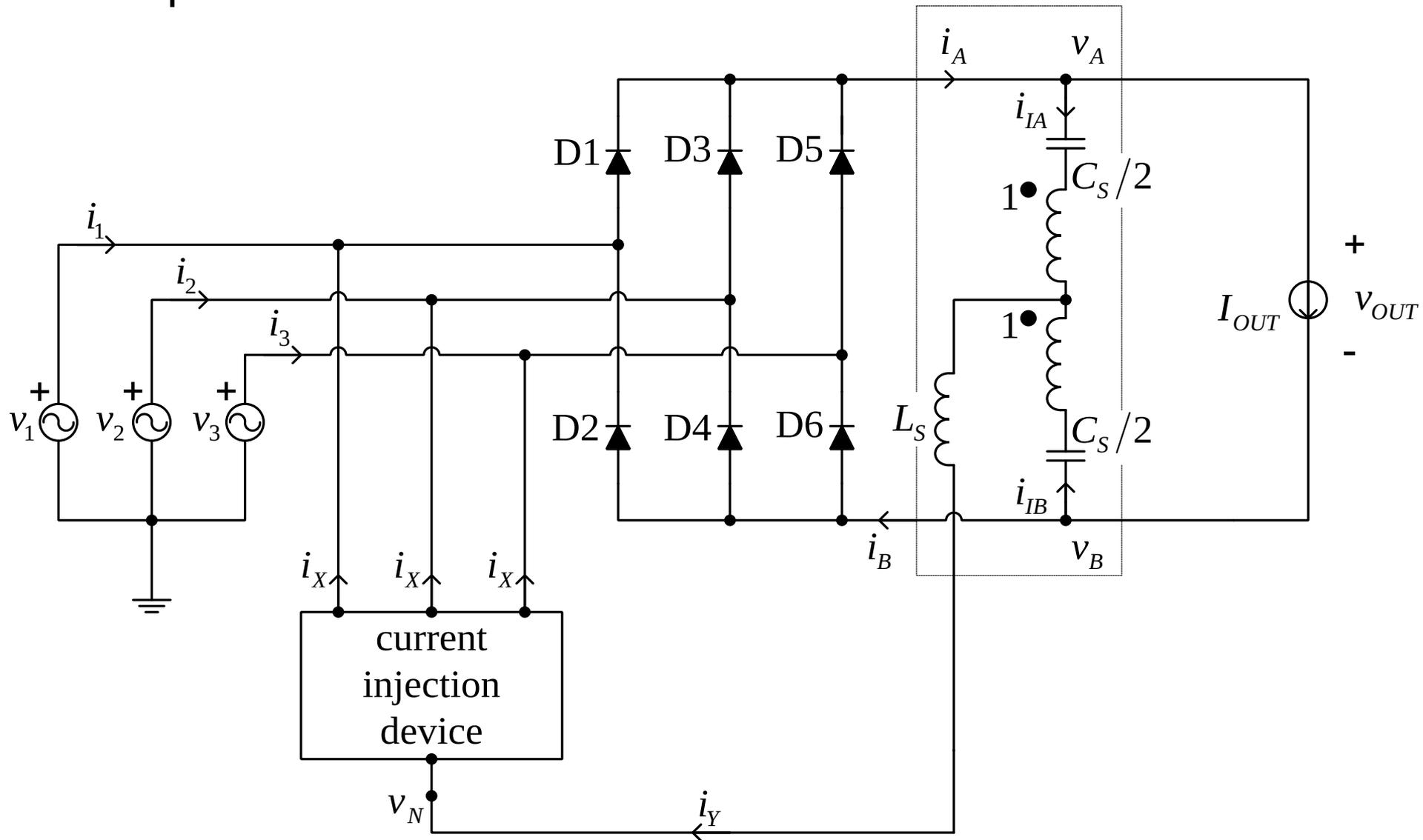
Ext / 200mV

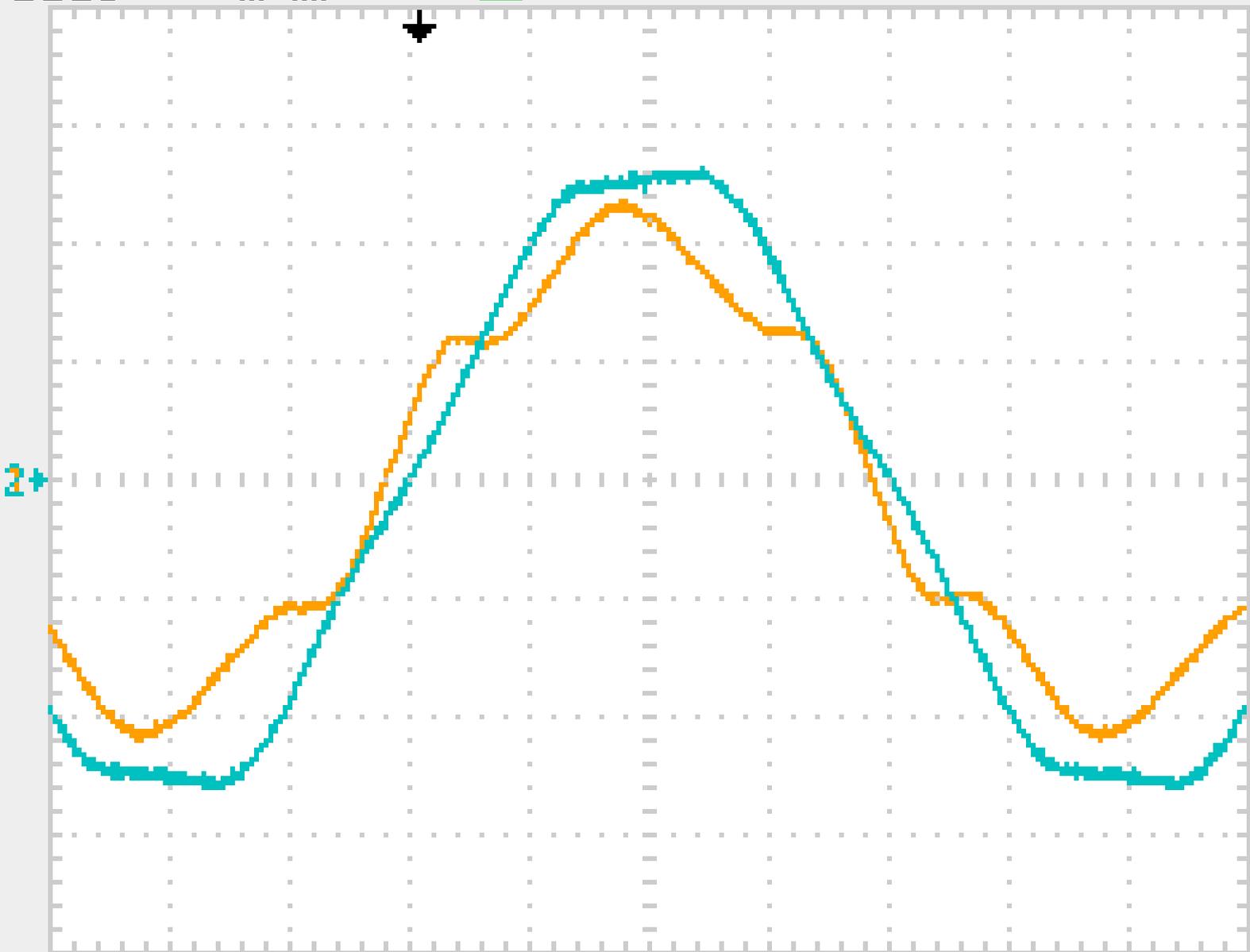
CH4 50.0V

12-Jun-08 20:27

49.9974Hz

real problem:





Coupling

DC

BW Limit

On

20MHz

Volts/Div

Coarse

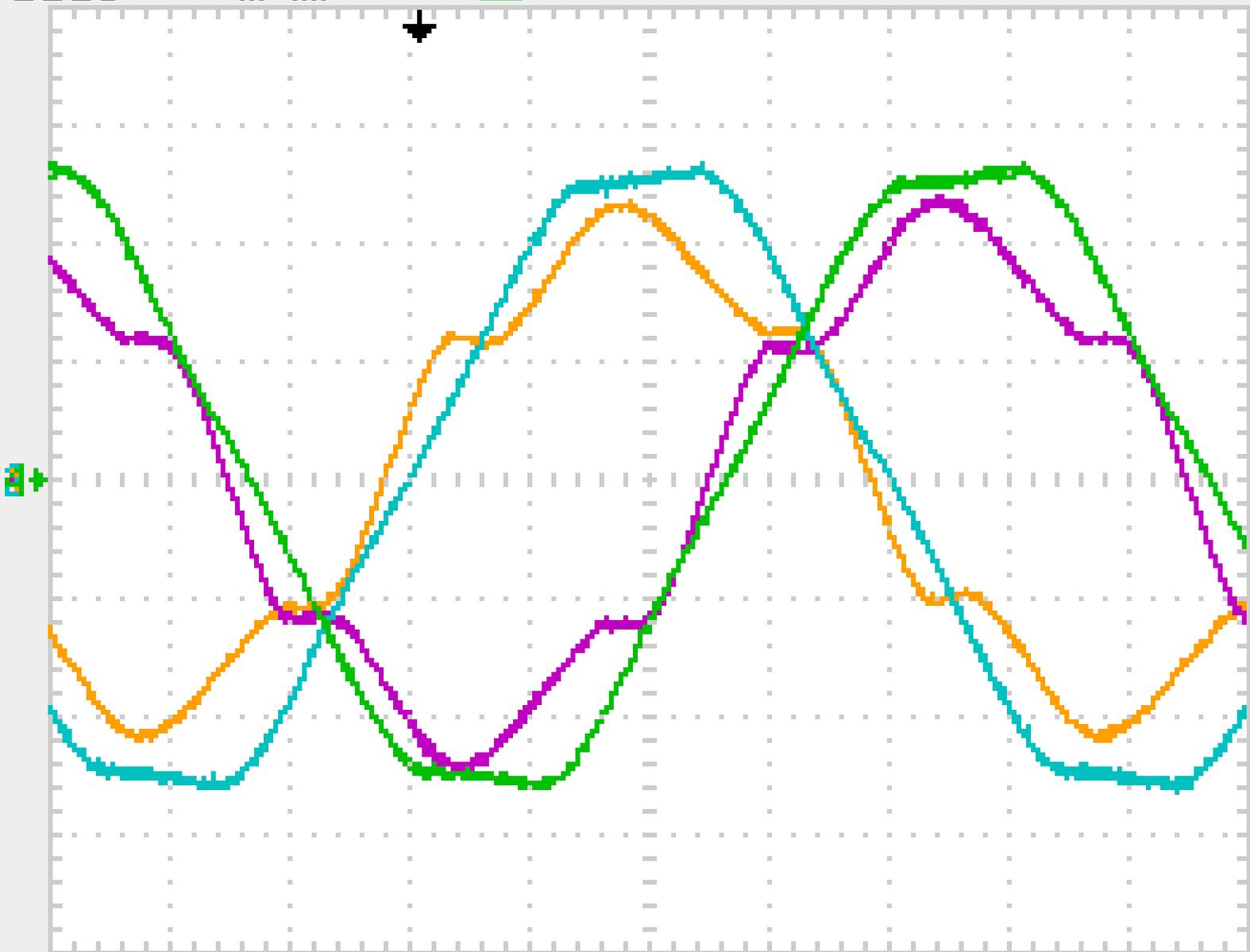
Probe

100A/V

Current

Invert

Off



Coupling

DC

Band Limit

On

20MHz

Volts/Div

Coarse

Probe

100X

Voltage

Invert

Off

CH1 5.00AB_{ij} CH2 50.0VB_{ij} M 2.50ms Ext / 200mV

CH3 5.00AB_{ij} CH4 50.0VB_{ij} 12-Jun-08 20:21

49.9893Hz

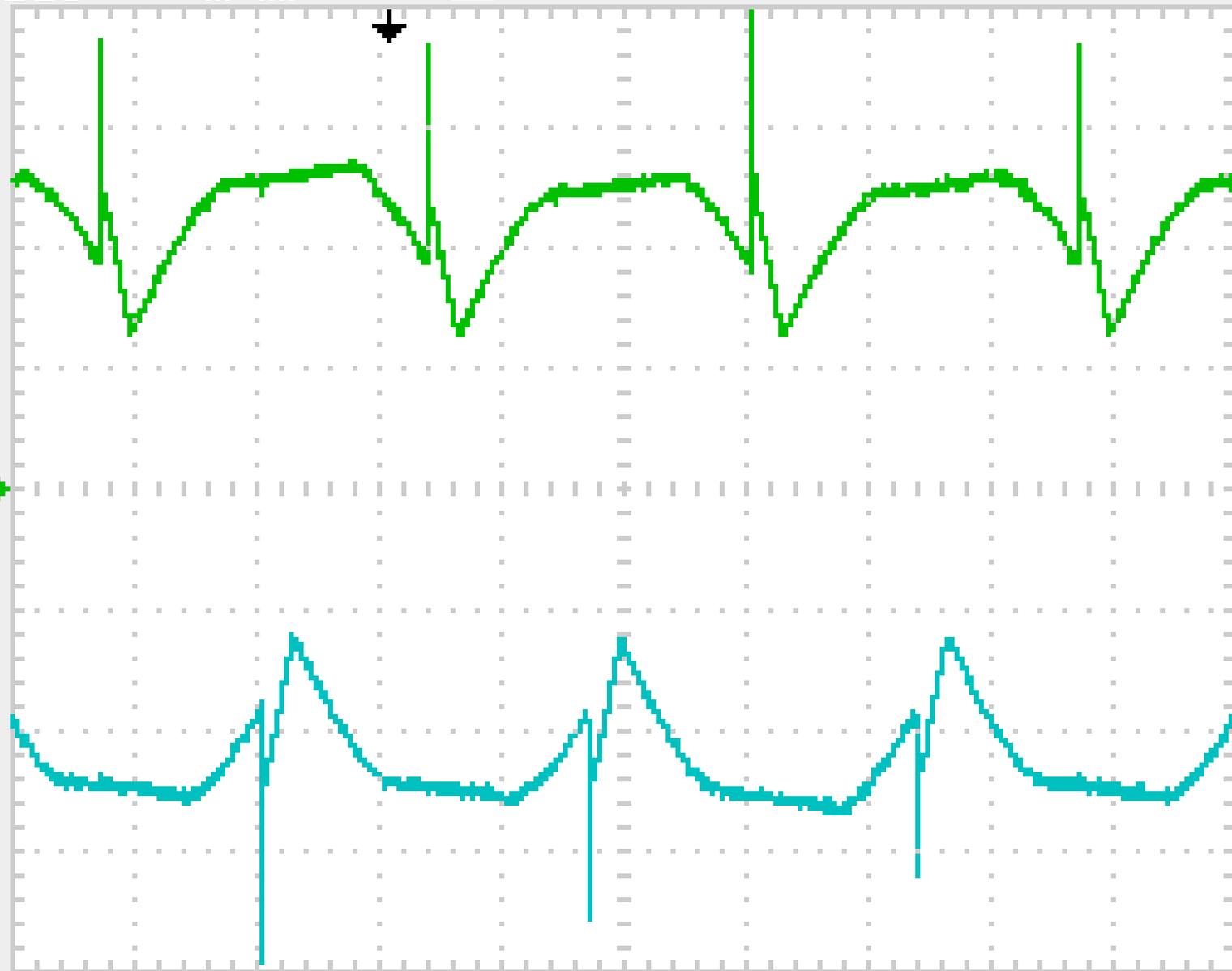
Tek

Ω

Trig'd

M Pos: 4.800ms

CH1



Coupling

DC

Band Limit

On

20MHz

Volts/Div

Coarse

Probe

100A/V

Current

Invert

Off

CH2 50.0VB_{off}

M 2.50ms

Ext \sqrt 200mV

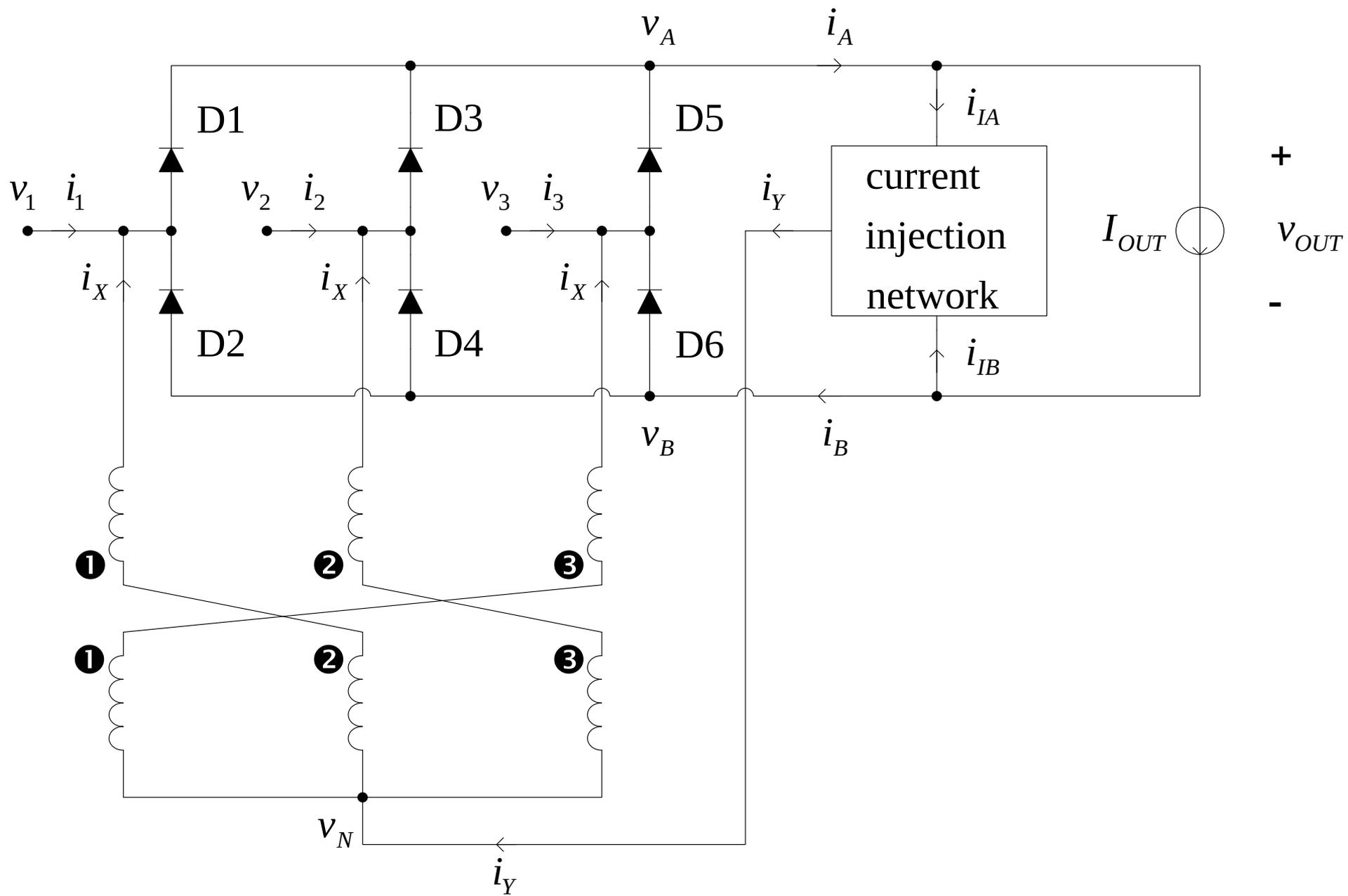
CH4 50.0VB_{off}

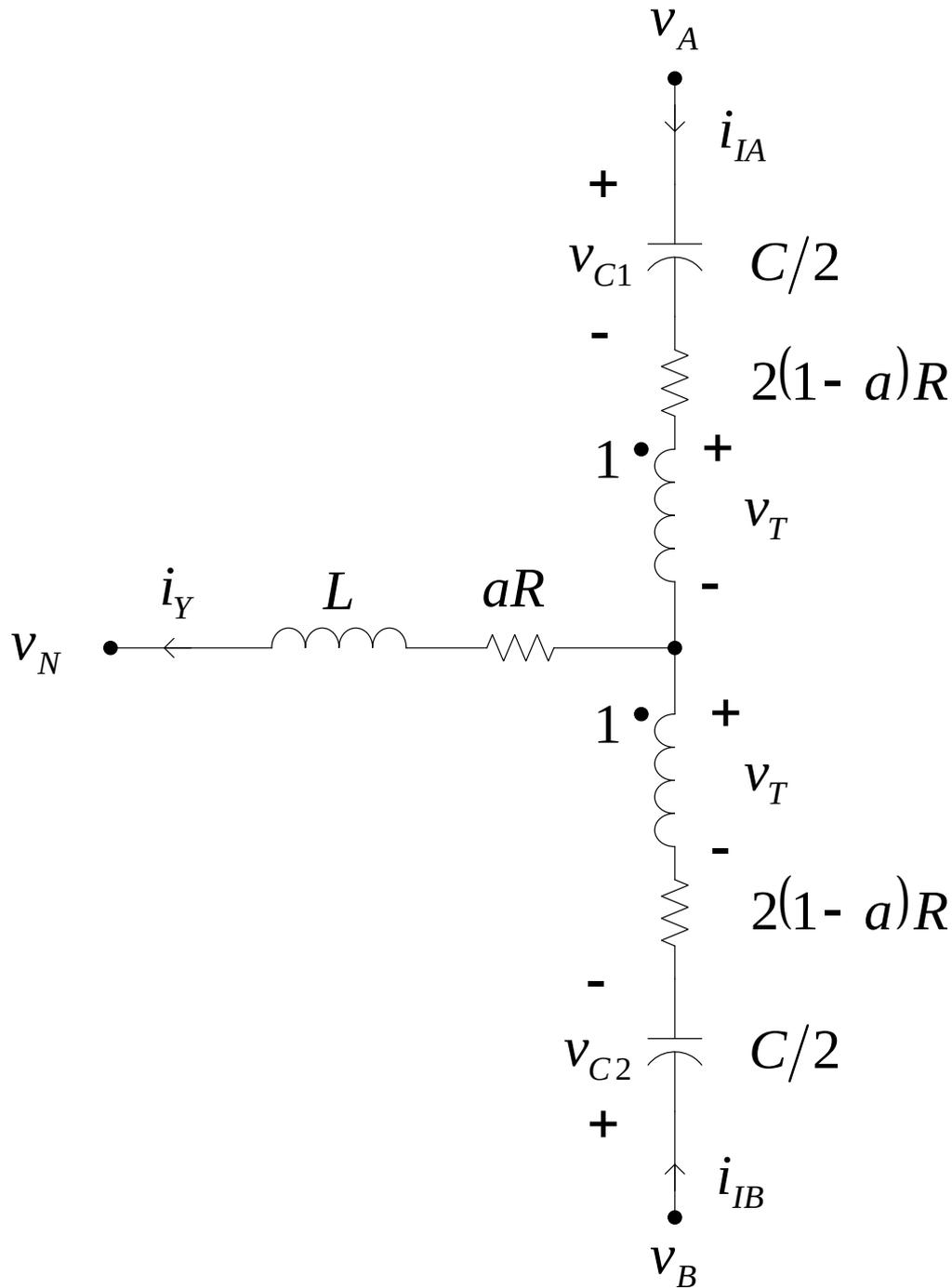
12-Jun-08 20:26

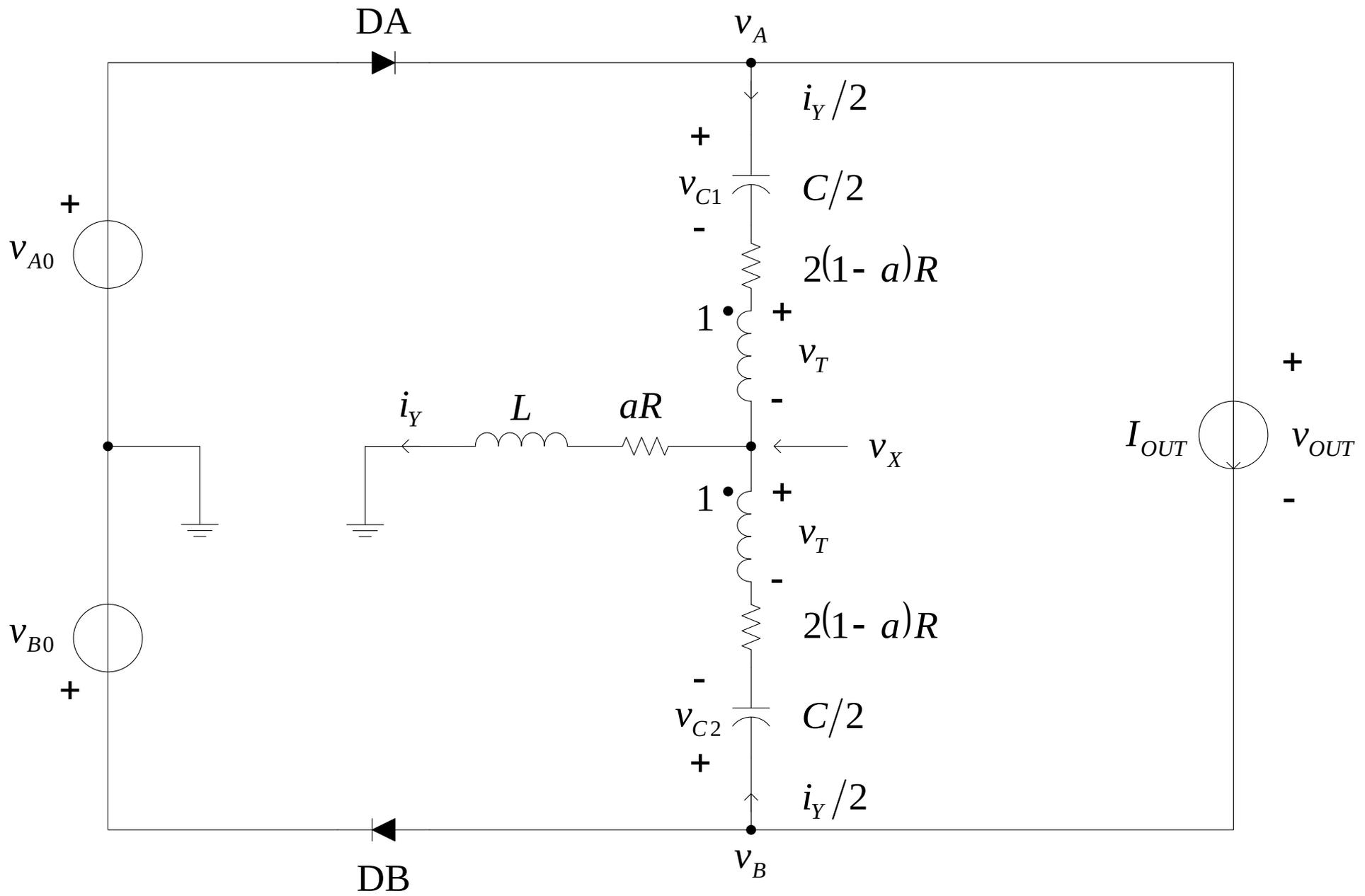
50.0004Hz

- How to analyze the circuit?
- There is no closed-form solution
- Even if there is a closed-form solution, would it be helpful?
- Numerical methods cannot be avoided
- How to make the numerical results general enough?
- Soldering iron thinking – > SPICE thinking
- What is the minimum number of parameters that affect the rectifier operation?
- How to introduce proper normalization?

Too many questions . . .







state	equations	conditions	transition if the condition is violated
1	$L \frac{di_Y}{dt} = -Ri_Y - v_C + \frac{v_{A0} + v_{B0}}{2}$	$-2I_{OUT} \leq i_Y$	to state 2
	$C \frac{dv_C}{dt} = i_Y$	$i_Y \leq 2I_{OUT}$	to state 3
2	$i_Y = -2I_{OUT}$	$v_C > \frac{v_{A0} + v_{B0}}{2} + 2RI_{OUT}$	to state 1
	$C \frac{dv_C}{dt} = -2I_{OUT}$		
3	$i_Y = 2I_{OUT}$	$v_C < \frac{v_{A0} + v_{B0}}{2} - 2RI_{OUT}$	to state 1
	$C \frac{dv_C}{dt} = 2I_{OUT}$		

$$m = \frac{v}{V_m}$$

$$R_0 = \sqrt{\frac{L}{C}} \quad \omega_R = \frac{1}{\sqrt{LC}}$$

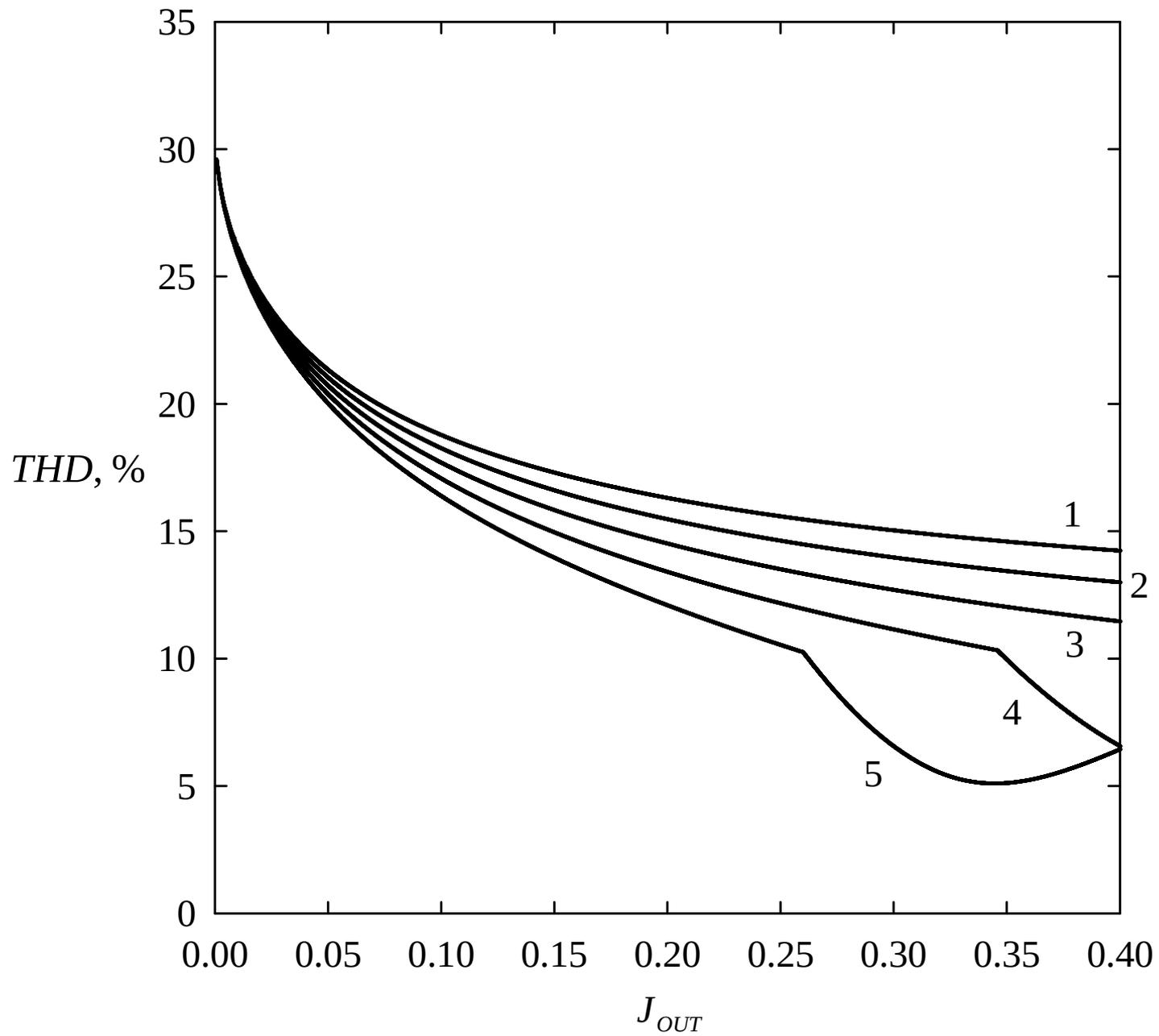
$$\rho = \frac{R}{R_0}$$

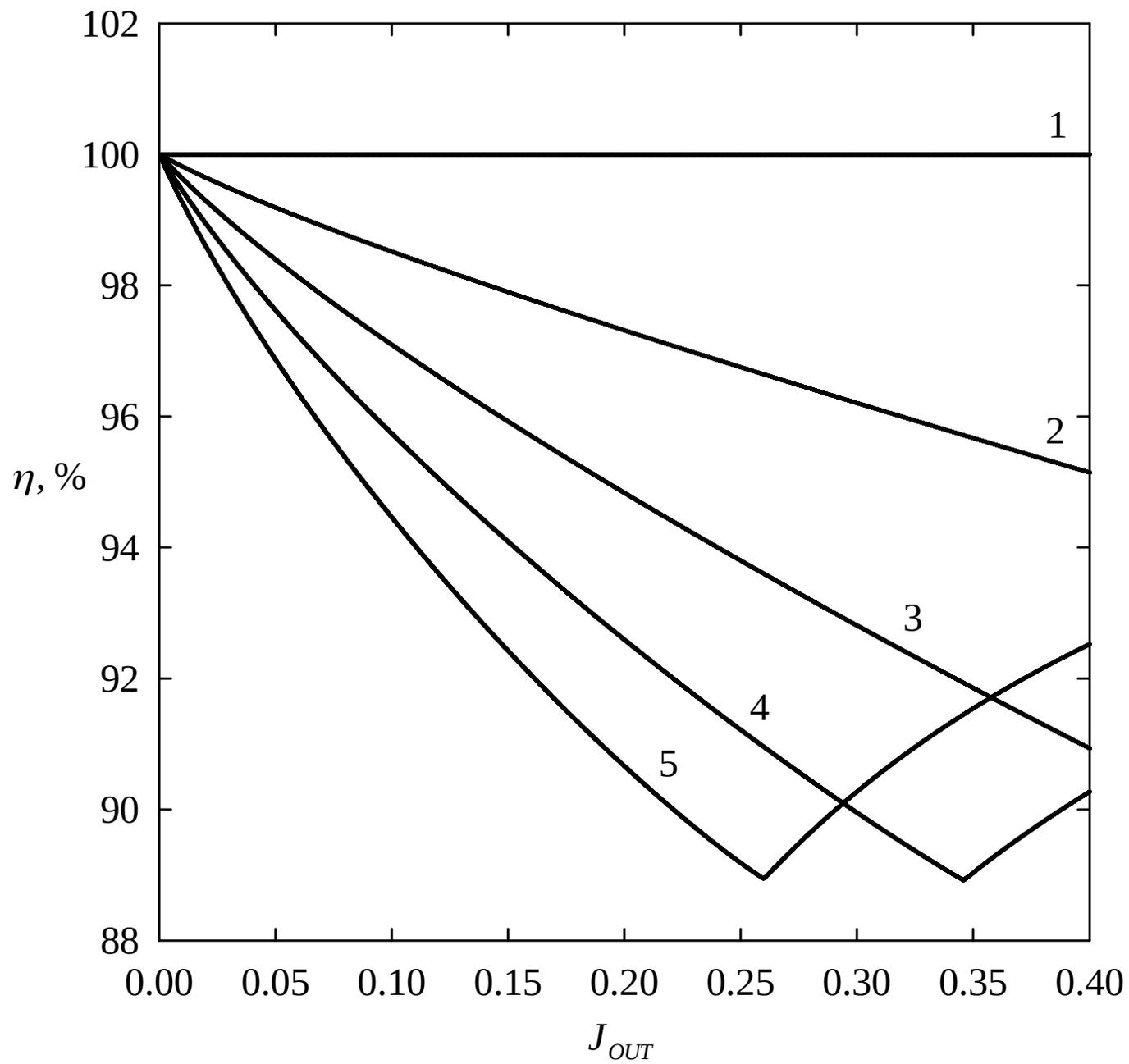
$$j = \frac{R_0}{V_m} i$$

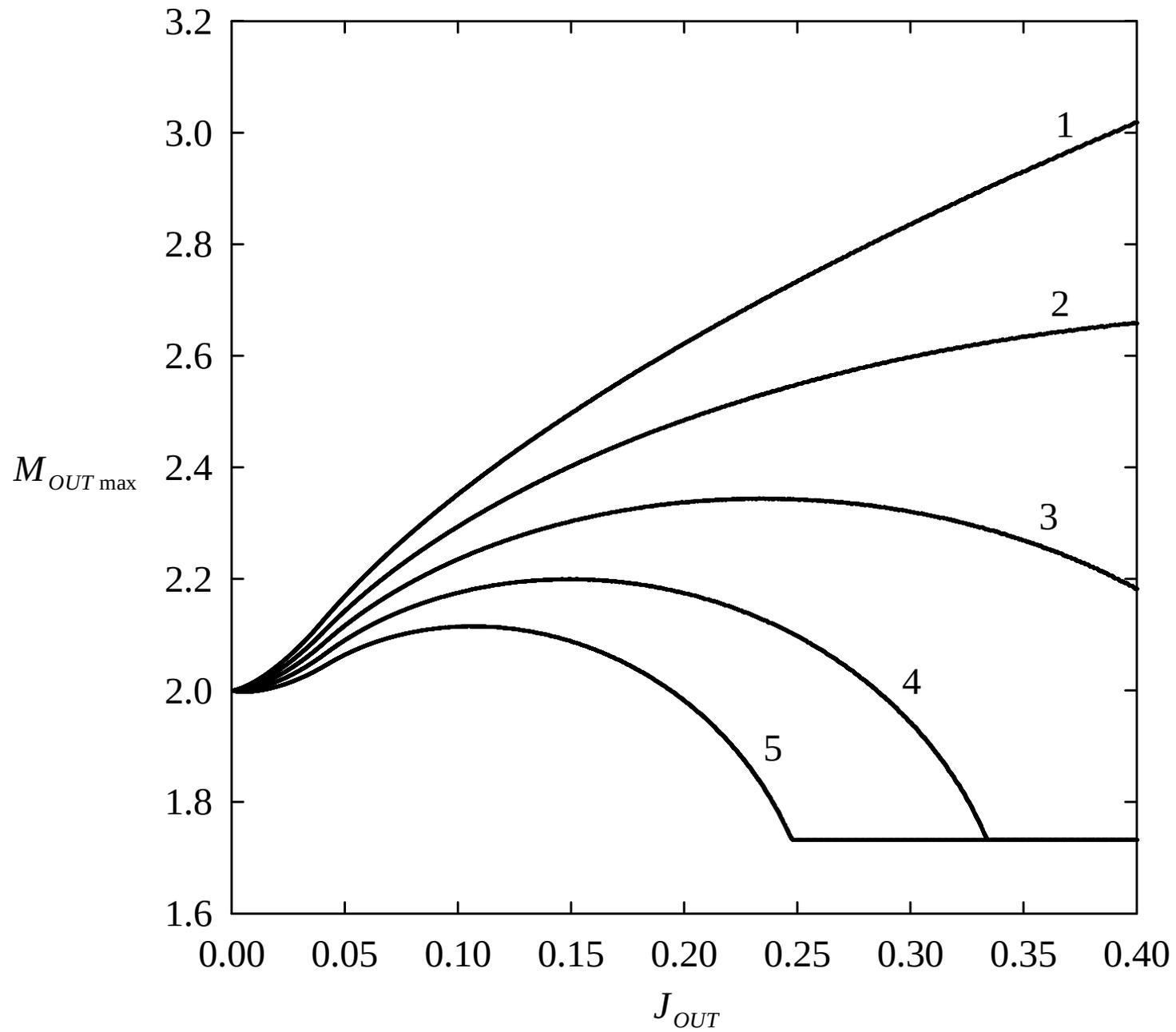
$$\phi = \omega_0 t$$

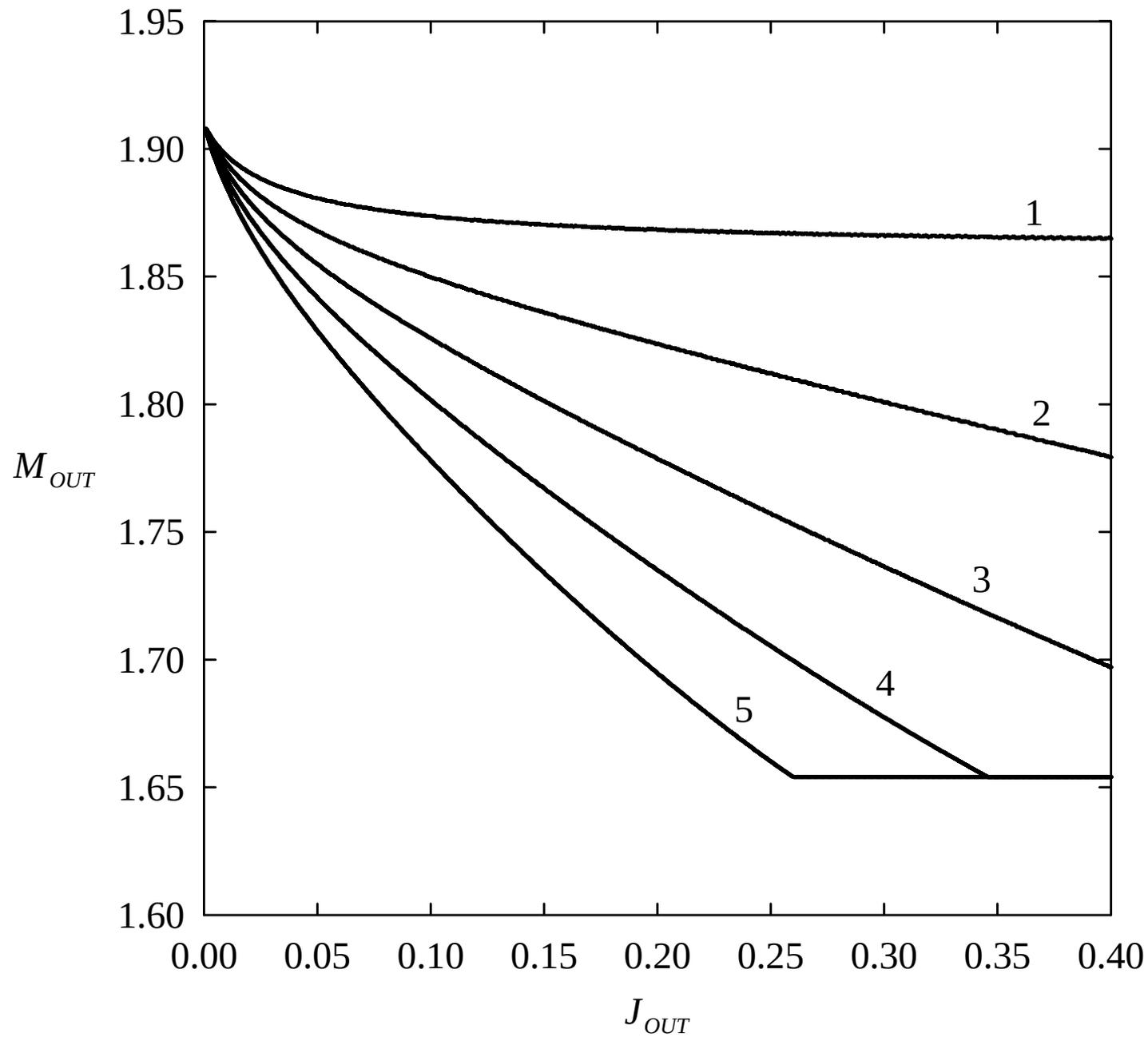
$$r = \frac{\omega_R}{3\omega_0}$$

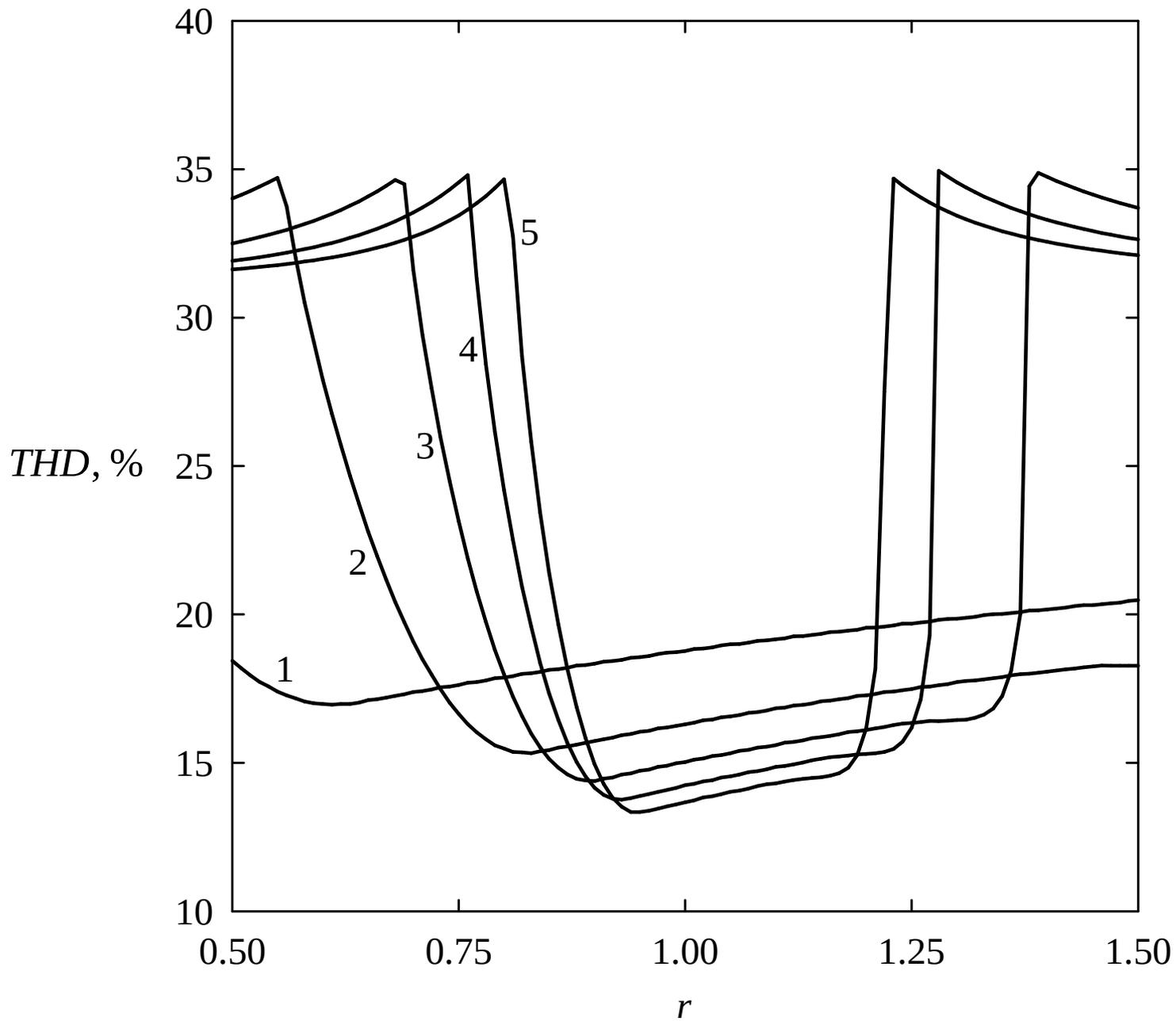
state	equations	conditions	transition if the condition is violated
1	$\frac{dj_Y}{d\varphi} = 3r \left(-\rho j_Y - m_C + \frac{m_{A0} + m_{B0}}{2} \right)$	$-2J_{OUT} \leq j_Y$	to state 2
	$\frac{dm_C}{d\varphi} = 3rj_Y$	$j_Y \leq 2J_{OUT}$	to state 3
2	$j_Y = -2J_{OUT}$	$m_C > \frac{m_{A0} + m_{B0}}{2} + 2\rho J_{OUT}$	to state 1
	$\frac{dm_C}{d\varphi} = -6rJ_{OUT}$		
3	$j_Y = 2J_{OUT}$	$m_C < \frac{m_{A0} + m_{B0}}{2} - 2\rho J_{OUT}$	to state 1
	$\frac{dm_C}{d\varphi} = 6rJ_{OUT}$		

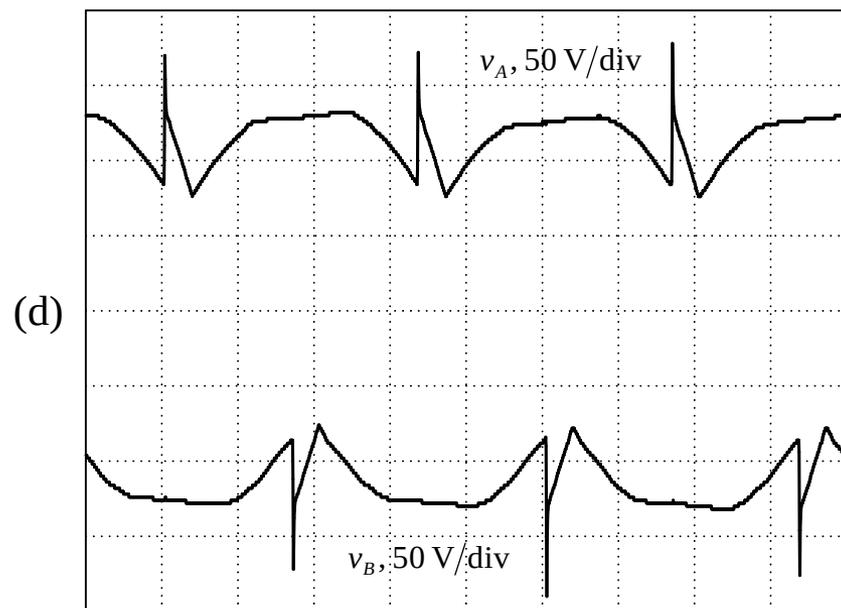
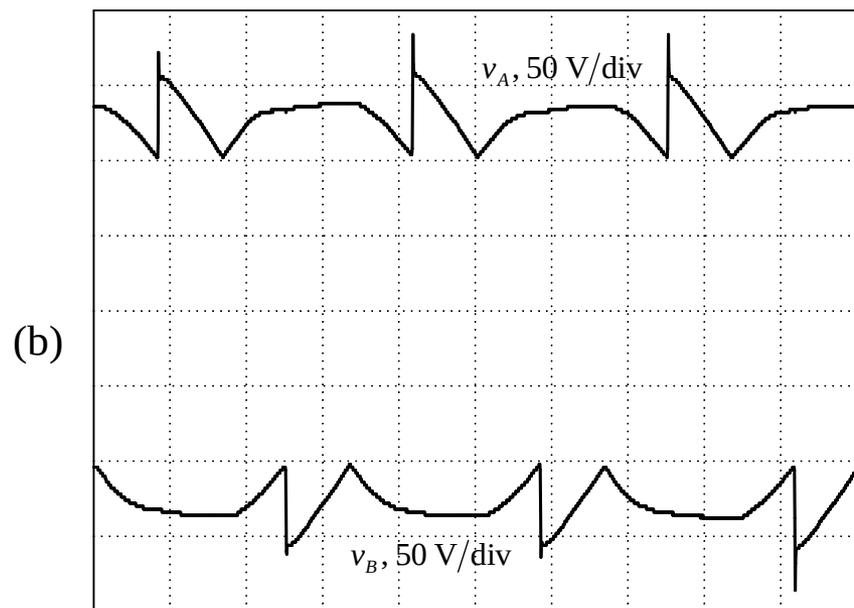
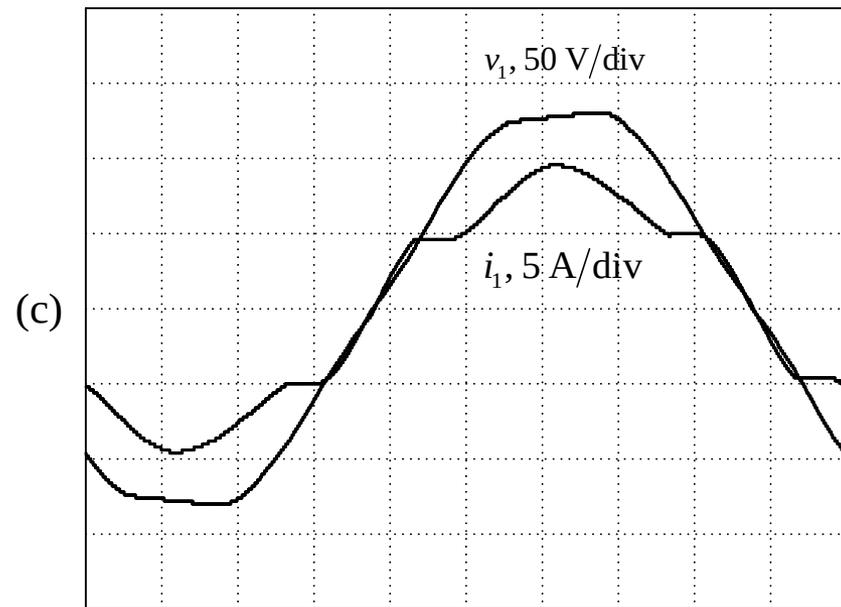
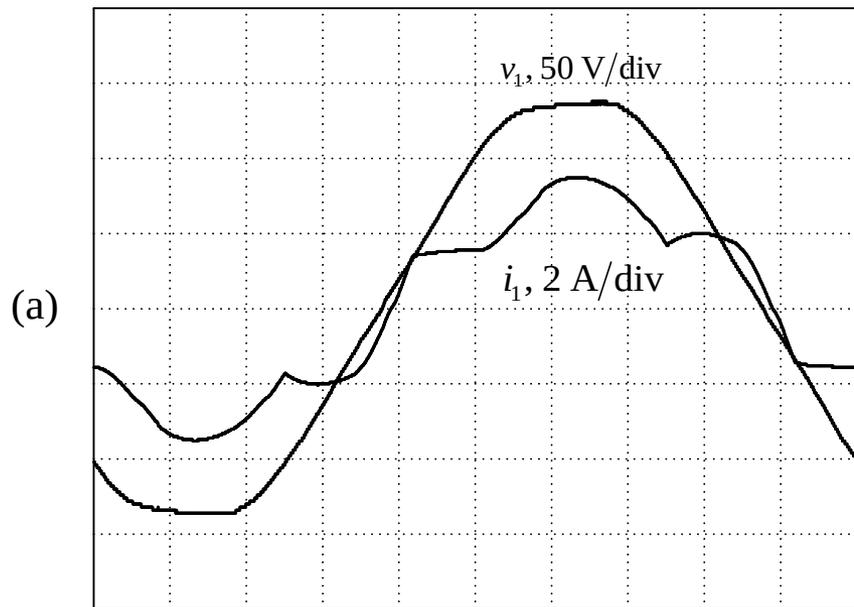


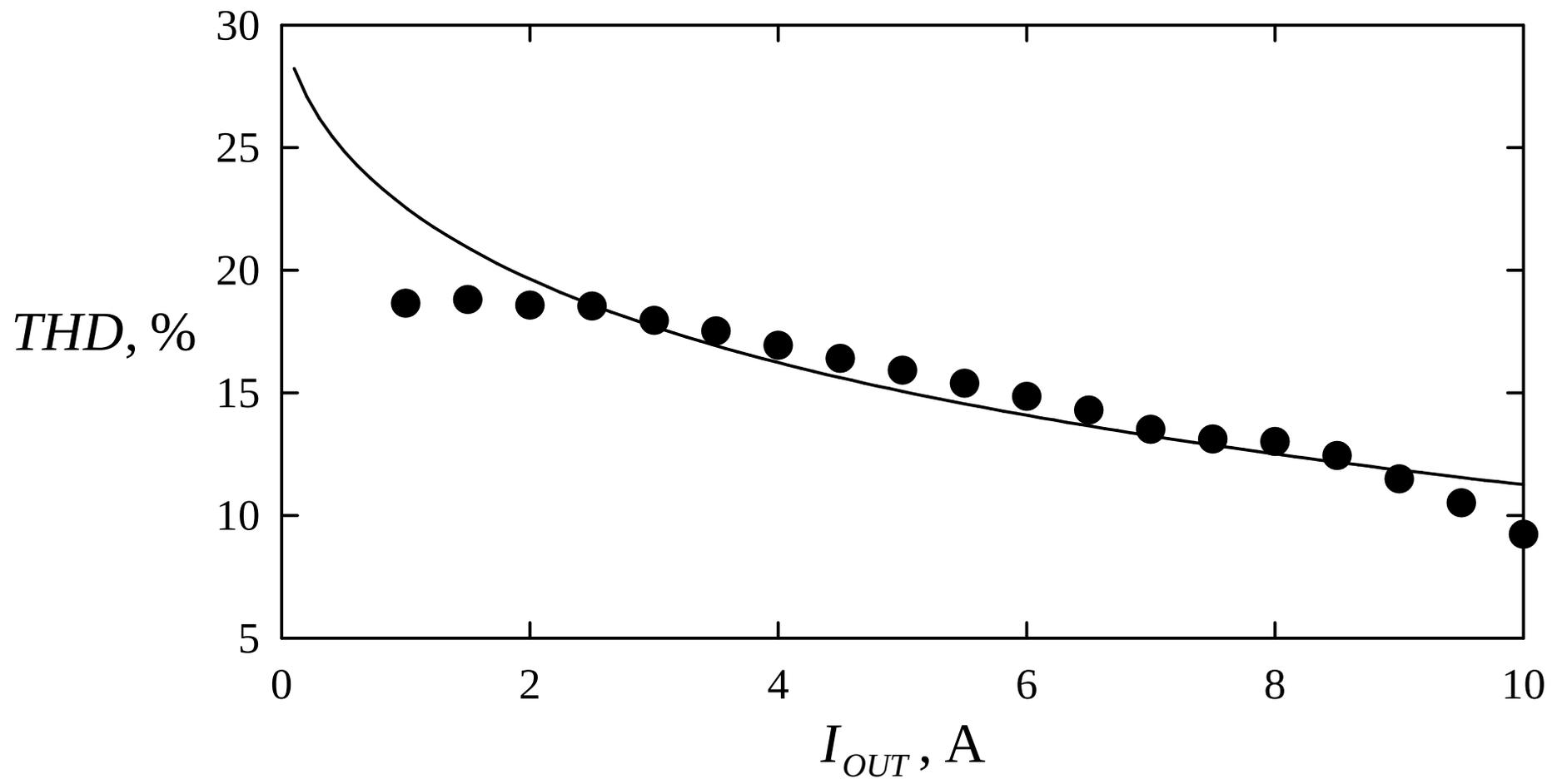


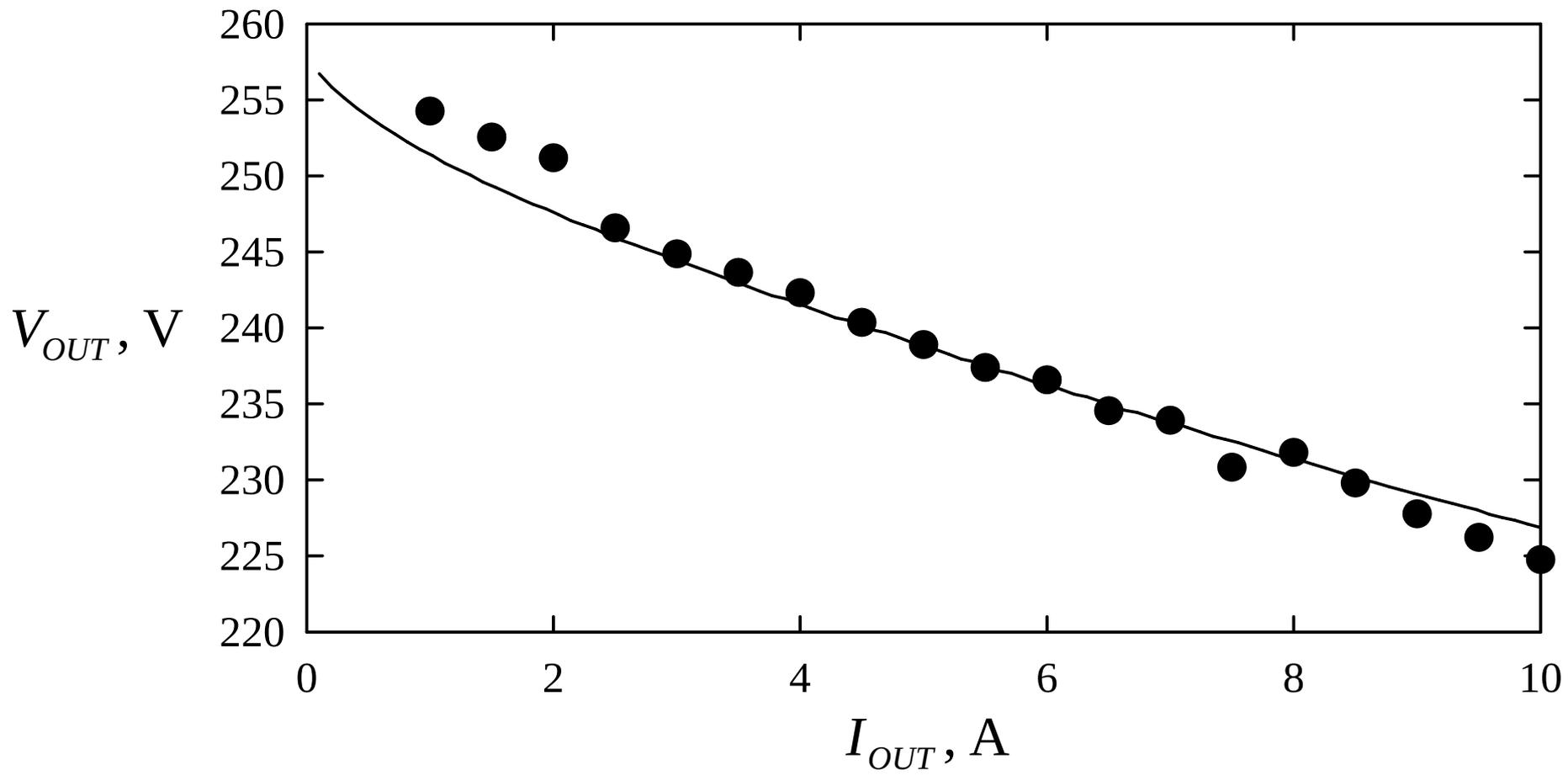












An Extrapolation Method For Accelerated Convergence to Steady State Solution of Power Electronics Circuits

Problem Statement

$$x_{k+1} = f(x_k)$$

$$\xi = f(\xi)$$

Generalized Aitken's Method

$$f(x) = \mathbf{A}\mathbf{x} + B$$

$$\xi = f(\xi) = \mathbf{A}\xi + B$$

$$x_{k+1} = f(x_k) = \mathbf{A}\mathbf{x}_k + B$$

$$x_{k+1} - \xi = A(x_k - \xi)$$

$$\xi = (I_n - A)^{-1}(x_{i+1} - \mathbf{A}\mathbf{x}_i)$$

$$e_i = x_i - x_{i-1}$$

$$x_{i+1} - x_i = A(x_i - x_{i-1})$$

$$e_2 = \mathbf{A}e_1$$

$$e_3 = \mathbf{A}e_2$$

...

$$e_{n+1} = \mathbf{A}e_n$$

$$\begin{bmatrix} e_2 \\ e_3 \\ \vdots \\ e_{n+1} \end{bmatrix} = \begin{bmatrix} \mathbf{A} & \mathbf{0}_{n \times n} & \cdots & \mathbf{0}_{n \times n} \\ \mathbf{0}_{n \times n} & \mathbf{A} & \cdots & \mathbf{0}_{n \times n} \\ \vdots & \vdots & \ddots & \vdots \\ \mathbf{0}_{n \times n} & \mathbf{0}_{n \times n} & \cdots & \mathbf{A} \end{bmatrix} \begin{bmatrix} e_1 \\ e_2 \\ \vdots \\ e_n \end{bmatrix}$$

$$\begin{bmatrix} e_2 \\ e_3 \\ \vdots \\ e_{n+1} \end{bmatrix} = \begin{bmatrix} E_1 \\ E_2 \\ \vdots \\ E_n \end{bmatrix} \begin{bmatrix} A_1^T \\ A_2^T \\ \vdots \\ A_n^T \end{bmatrix}$$

$$E_i = \begin{bmatrix} e_i^T & \mathbf{0}_{1 \times n} & \cdots & \mathbf{0}_{1 \times n} \\ \mathbf{0}_{1 \times n} & e_i^T & \cdots & \mathbf{0}_{1 \times n} \\ \vdots & \vdots & \ddots & \vdots \\ \mathbf{0}_{1 \times n} & \mathbf{0}_{1 \times n} & \cdots & e_i^T \end{bmatrix}$$

determine \mathbf{A} , determine ξ , start again

Problem with dynamic degeneration, common to the DCM

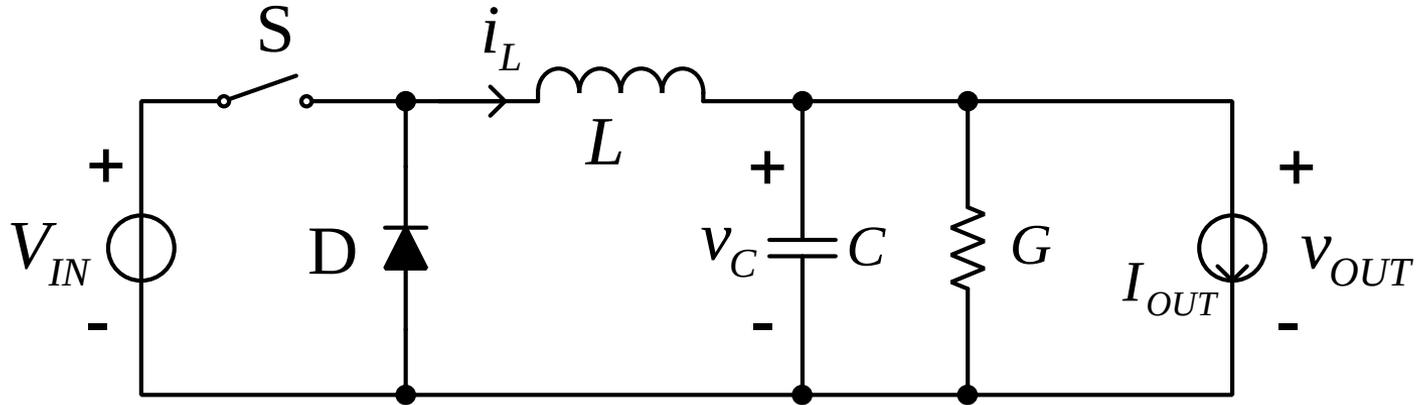
$$\begin{bmatrix} E_1 \\ E_2 \end{bmatrix} = \begin{bmatrix} e_{1,1} & e_{1,2} & 0 & 0 \\ 0 & 0 & e_{1,1} & e_{1,2} \\ e_{2,1} & e_{2,2} & 0 & 0 \\ 0 & 0 & e_{2,1} & e_{2,2} \end{bmatrix}$$

$$x_{0,2} = x_{1,2} = x_{2,2}$$

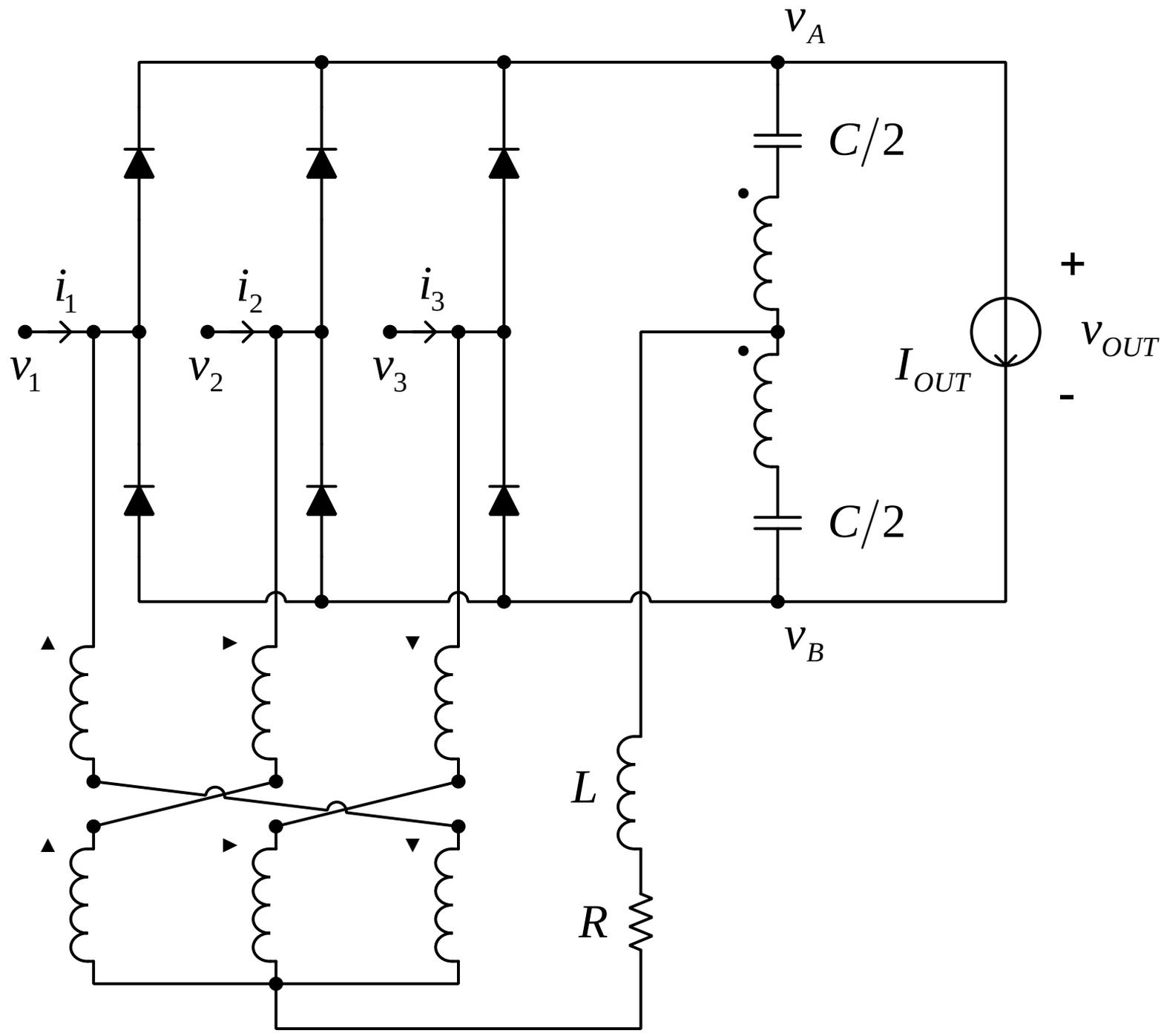
$$e_{1,2} = e_{2,2} = 0$$

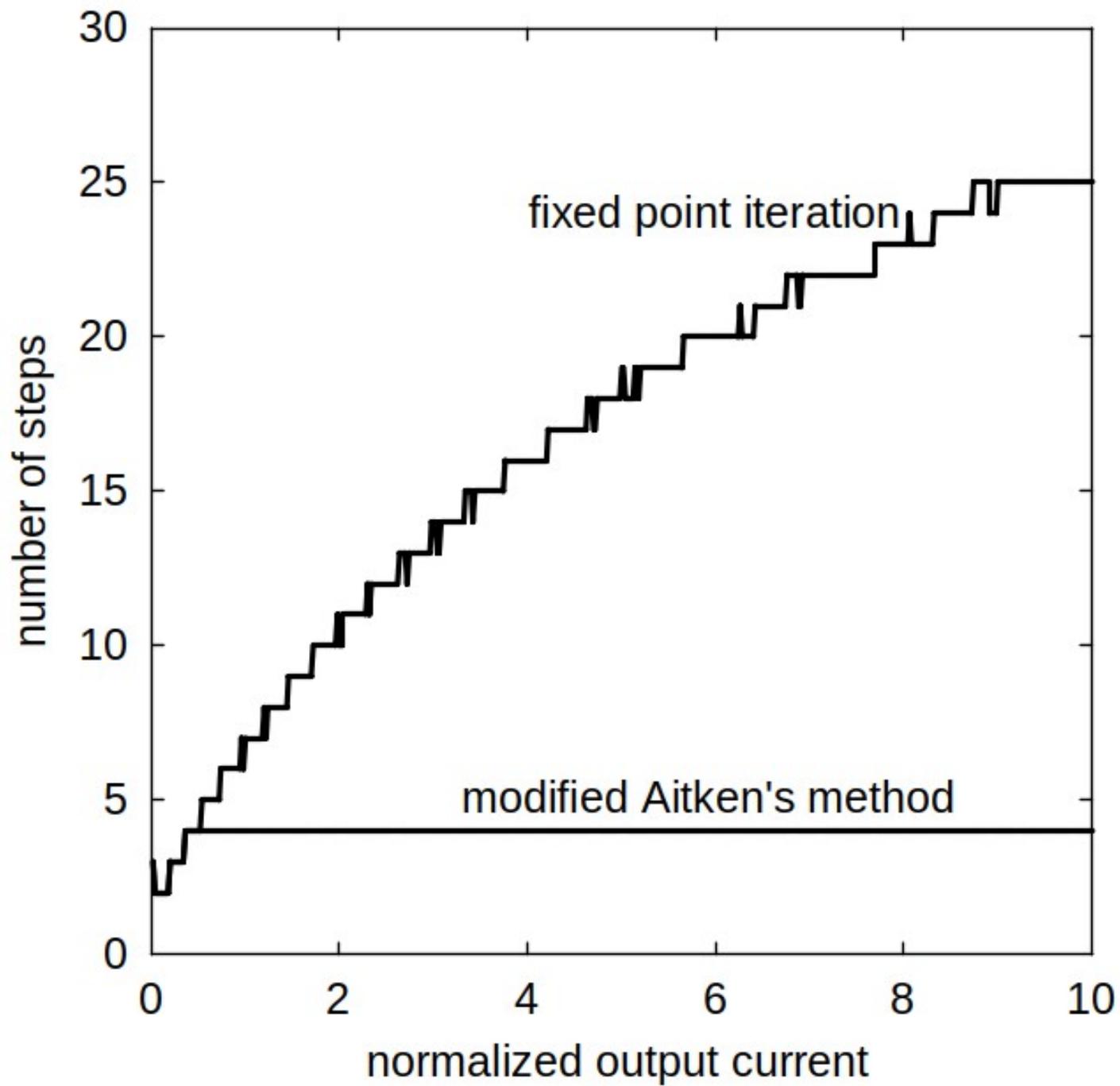
Convergence?

Simulation Results

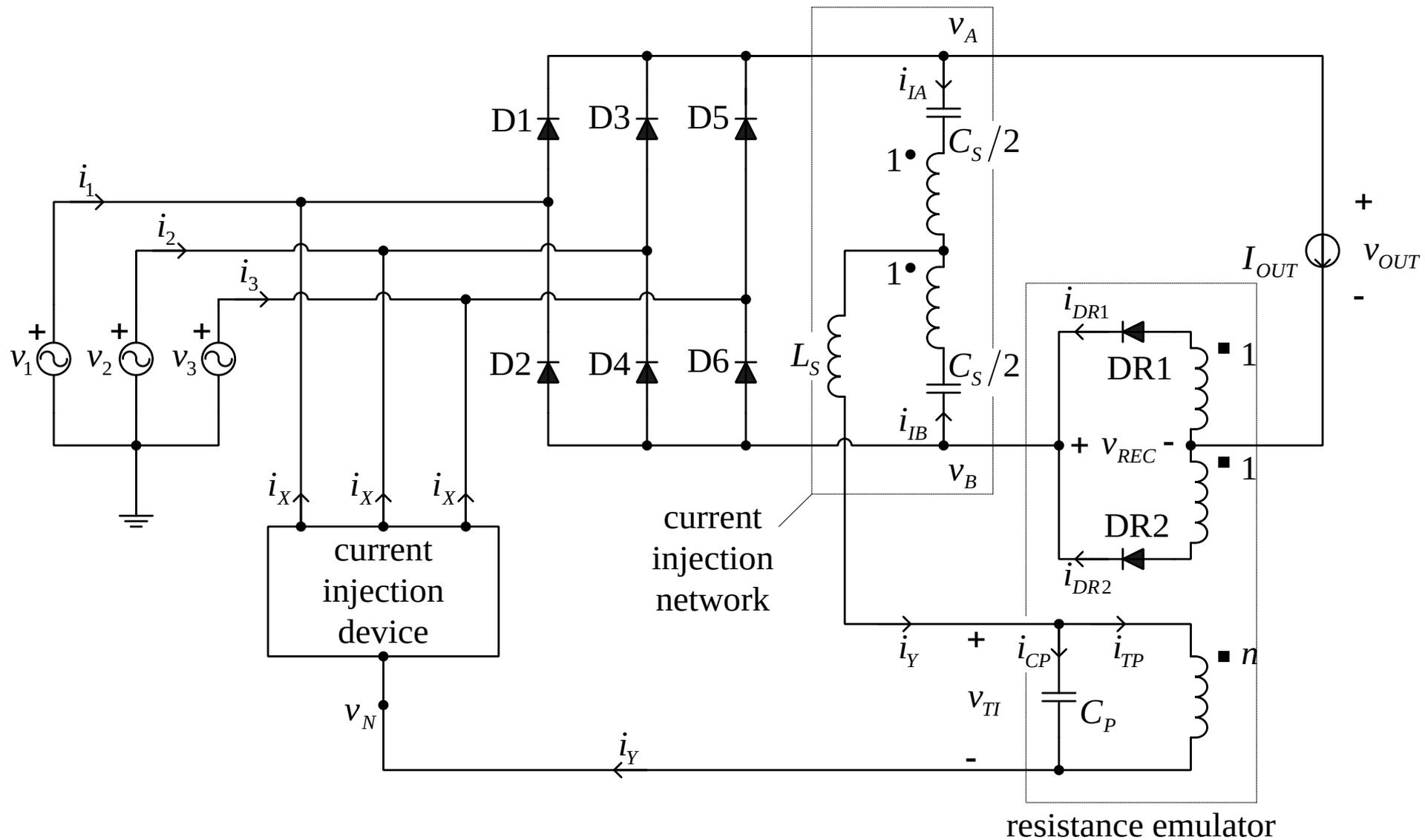


example	fixed point iteration	Aitken's method	modified Aitken's method
PWM, CCM	736	4	4
PWM, DCM	109	55	10
CMC, DCM	137	89	13
CMC, CCM	diverges	19	19

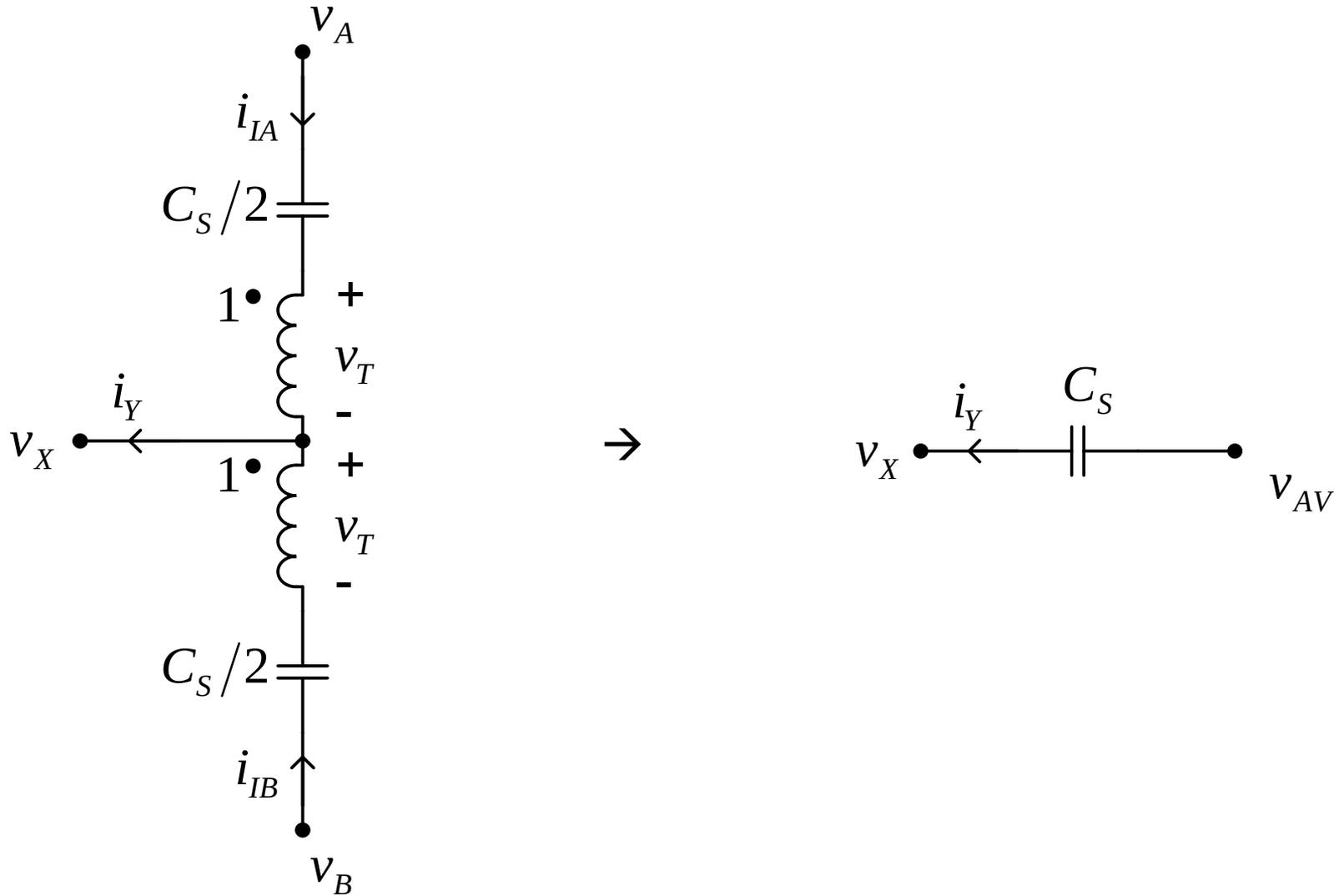


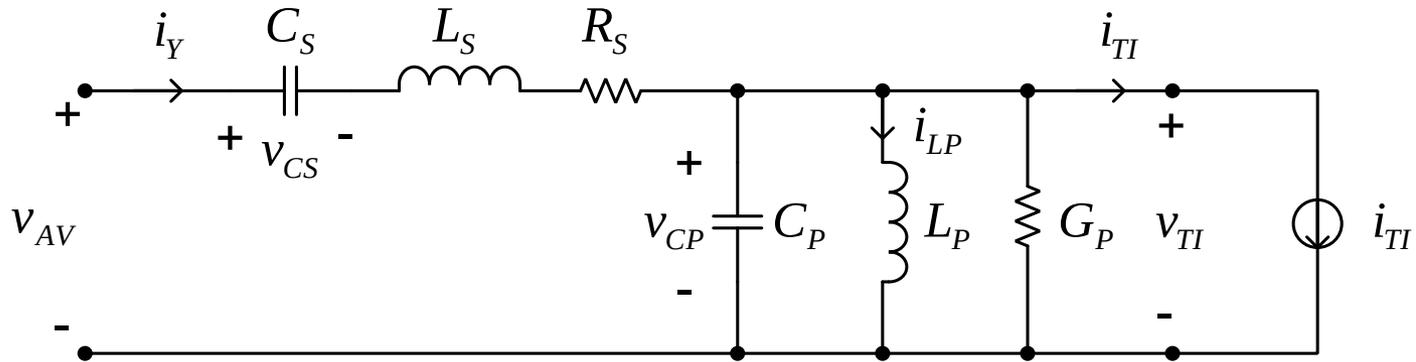


another problem:



analytical preparation, pre-analysis

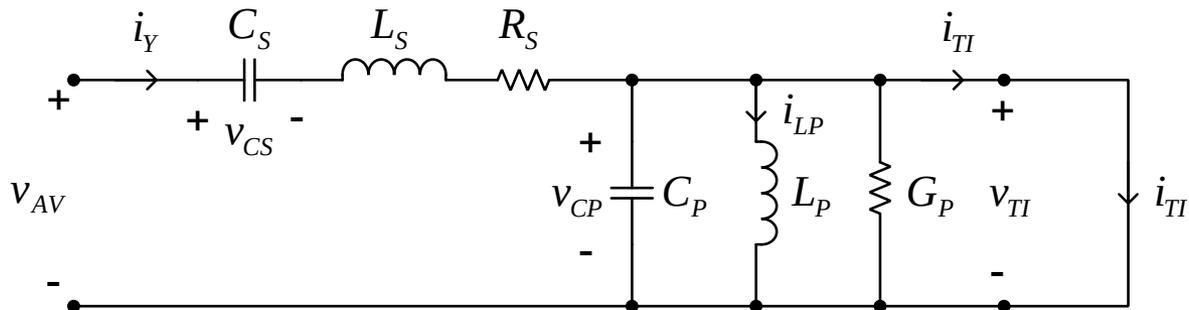




$$i_{TI} = \frac{1}{n} I_{OUT} \operatorname{sgn}(v_{CP})$$

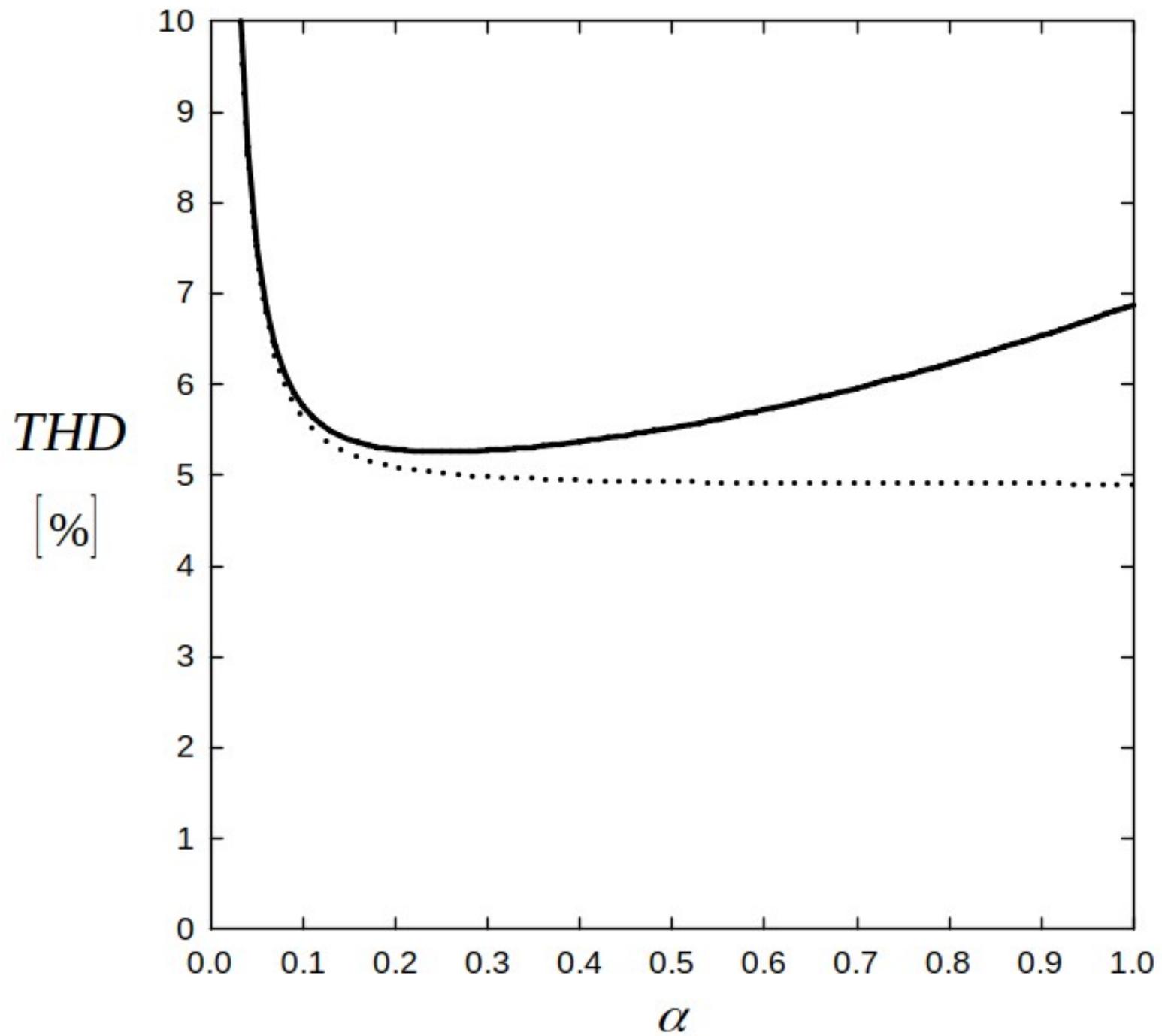
$$\frac{d}{dt} \begin{bmatrix} i_Y \\ v_{CS} \\ i_{LP} \\ v_{CP} \end{bmatrix} = \begin{bmatrix} -\frac{R_S}{L_S} & -\frac{1}{L_S} & 0 & -\frac{1}{L_S} \\ \frac{1}{C_S} & 0 & 0 & 0 \\ 0 & 0 & 0 & \frac{1}{L_P} \\ \frac{1}{C_P} & 0 & -\frac{1}{C_P} & -\frac{G_P}{C_P} \end{bmatrix} \begin{bmatrix} i_Y \\ v_{CS} \\ i_{LP} \\ v_{CP} \end{bmatrix} + \begin{bmatrix} \frac{v_{AV}}{L_S} \\ 0 \\ 0 \\ -\frac{1}{nC_P} I_{OUT} \text{sgn}(v_{CP}) \end{bmatrix}$$

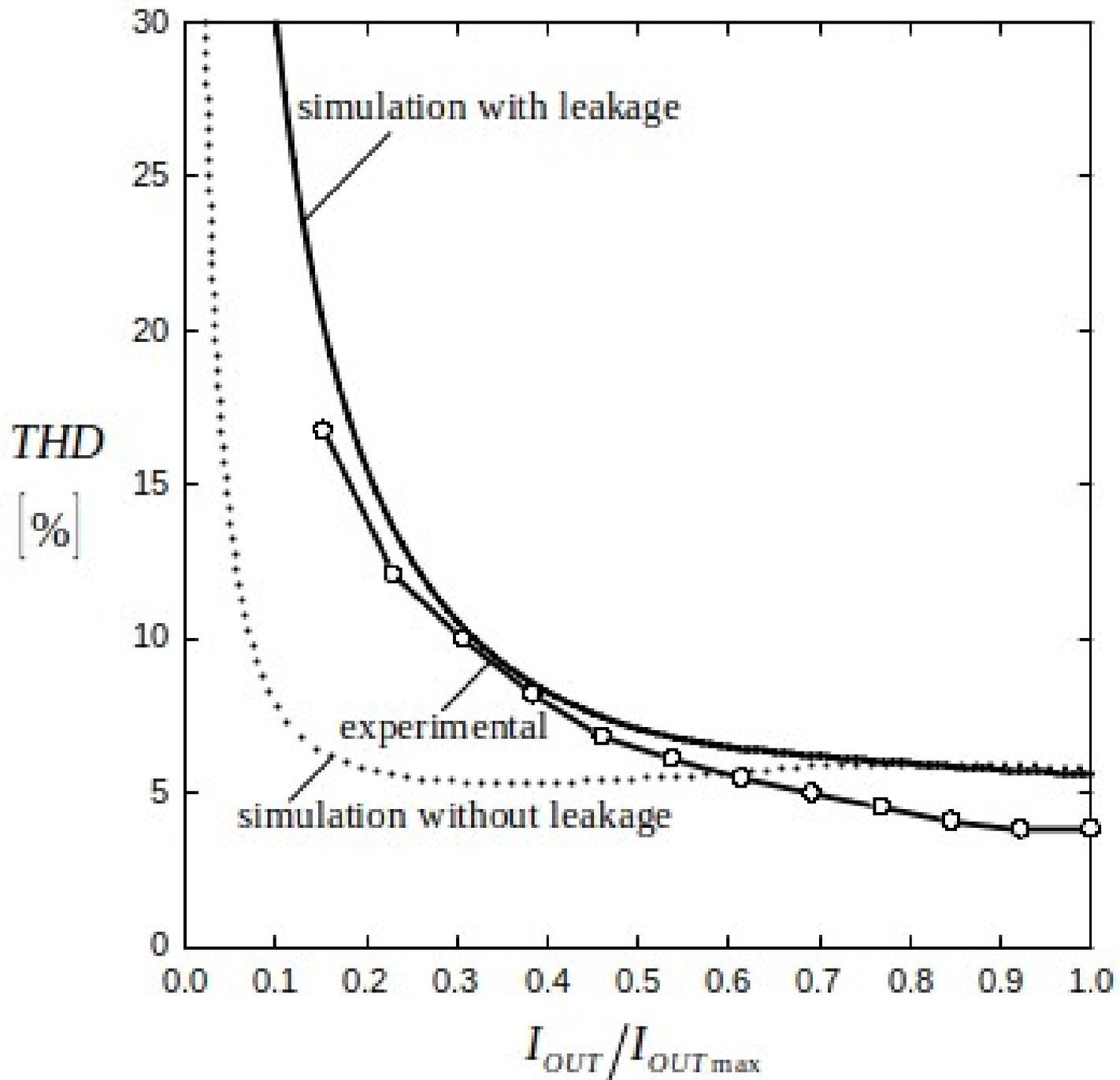
$$\frac{d}{dt} \begin{bmatrix} i_Y \\ v_{CS} \\ i_{LP} \end{bmatrix} = \begin{bmatrix} -\frac{R_S}{L_S} & -\frac{1}{L_S} & 0 \\ \frac{1}{C_S} & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} i_Y \\ v_{CS} \\ i_{LP} \end{bmatrix} + \begin{bmatrix} \frac{v_{AV}}{L_S} \\ 0 \\ 0 \end{bmatrix}$$



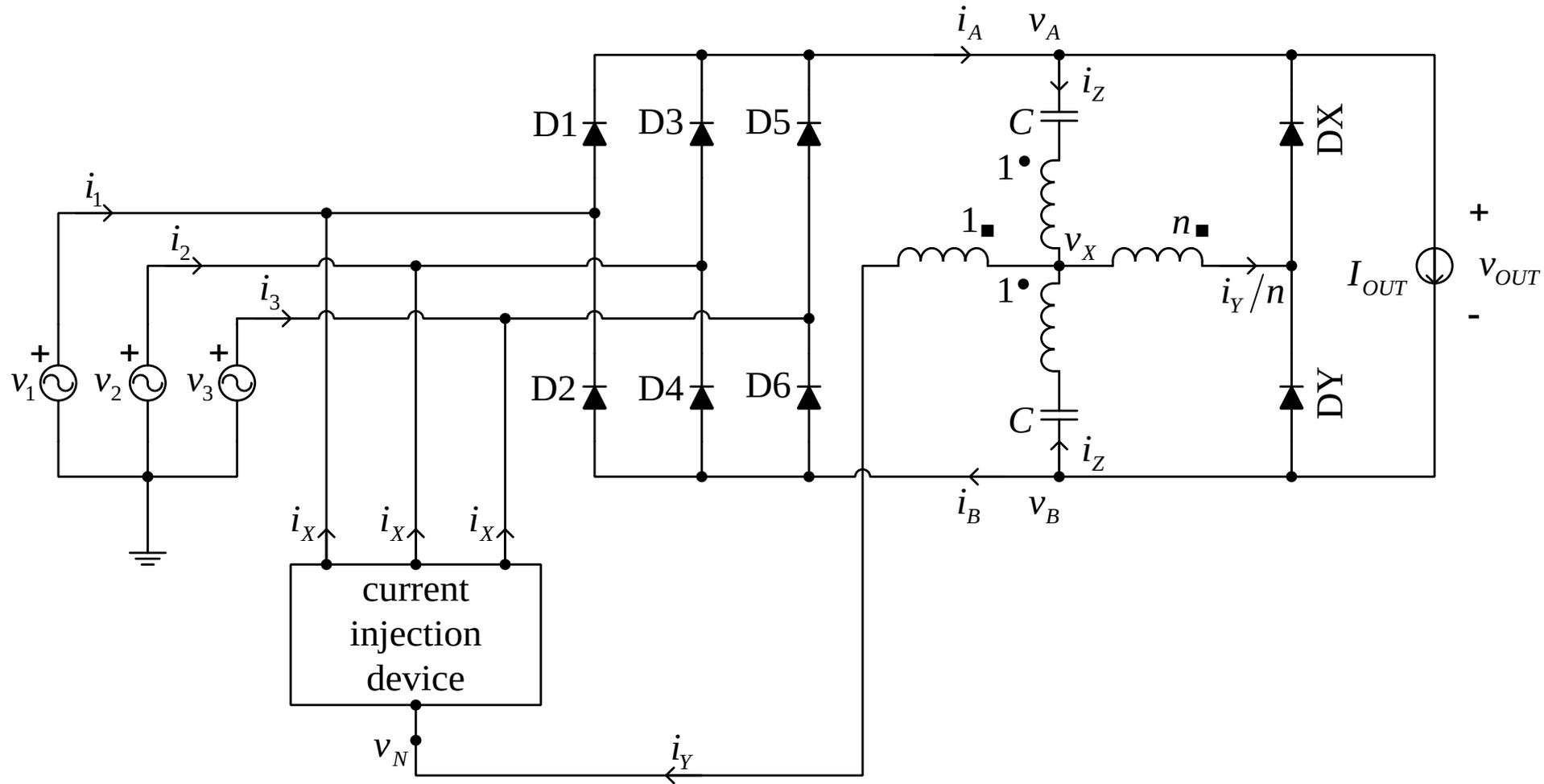
$$\frac{d}{d\phi} \begin{bmatrix} j_Y \\ m_{CS} \\ j_{LP} \\ m_{CP} \end{bmatrix} = \begin{bmatrix} -\frac{3r_S}{Q_S} & -\frac{3r_S}{\alpha\rho_S} & 0 & -\frac{3r_S}{\alpha\rho_S} \\ 3r_S\alpha\rho_S & 0 & 0 & 0 \\ 0 & 0 & 0 & \frac{3r_P}{\alpha\rho_P} \\ 3r_P\alpha\rho_P & 0 & -3r_P\alpha\rho_P & -\frac{3r_P}{Q_P} \end{bmatrix} \begin{bmatrix} j_Y \\ m_{CS} \\ j_{LP} \\ m_{CP} \end{bmatrix} + \begin{bmatrix} \frac{3r_S}{\alpha\rho_S} m_{AV} \\ 0 \\ 0 \\ -3r_P\alpha\rho_P \frac{\text{sgn}(m_{CP})}{n} \end{bmatrix}$$

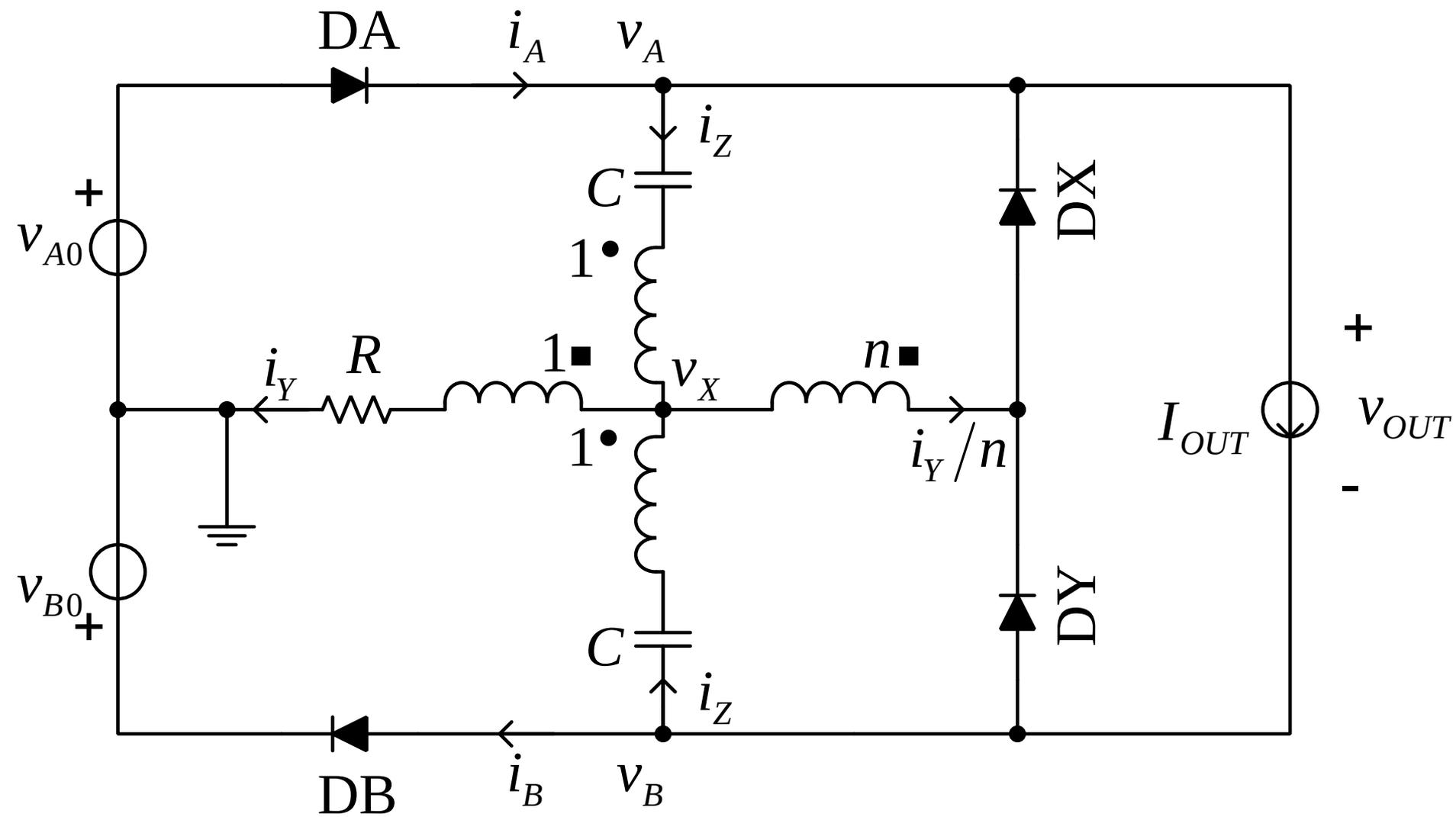
$$\frac{d}{d\phi} \begin{bmatrix} j_Y \\ m_{CS} \\ j_{LP} \\ m_{CP} \end{bmatrix} = \begin{bmatrix} 0 & -\frac{\pi\sqrt{3}}{\alpha} & 0 & -\frac{\pi\sqrt{3}}{\alpha} \\ \frac{3\sqrt{3}}{\pi}\alpha & 0 & 0 & 0 \\ 0 & 0 & 0 & \frac{\pi^2\sqrt{3}}{\alpha} \\ \frac{3\sqrt{3}}{\pi^2}\alpha & 0 & -\frac{3\sqrt{3}}{\pi^2}\alpha & 0 \end{bmatrix} \begin{bmatrix} j_Y \\ m_{CS} \\ j_{LP} \\ m_{CP} \end{bmatrix} + \begin{bmatrix} \frac{\pi\sqrt{3}}{\alpha} m_{AV} \\ 0 \\ 0 \\ -\frac{9\sqrt{3}}{8\pi}\alpha \text{sgn}(m_{CP}) \end{bmatrix}$$





and another problem (a fresh one, PESC 2008):





analyzed as a resistive circuit!

normalization:

$$m_X = \frac{v_X}{V_m}$$

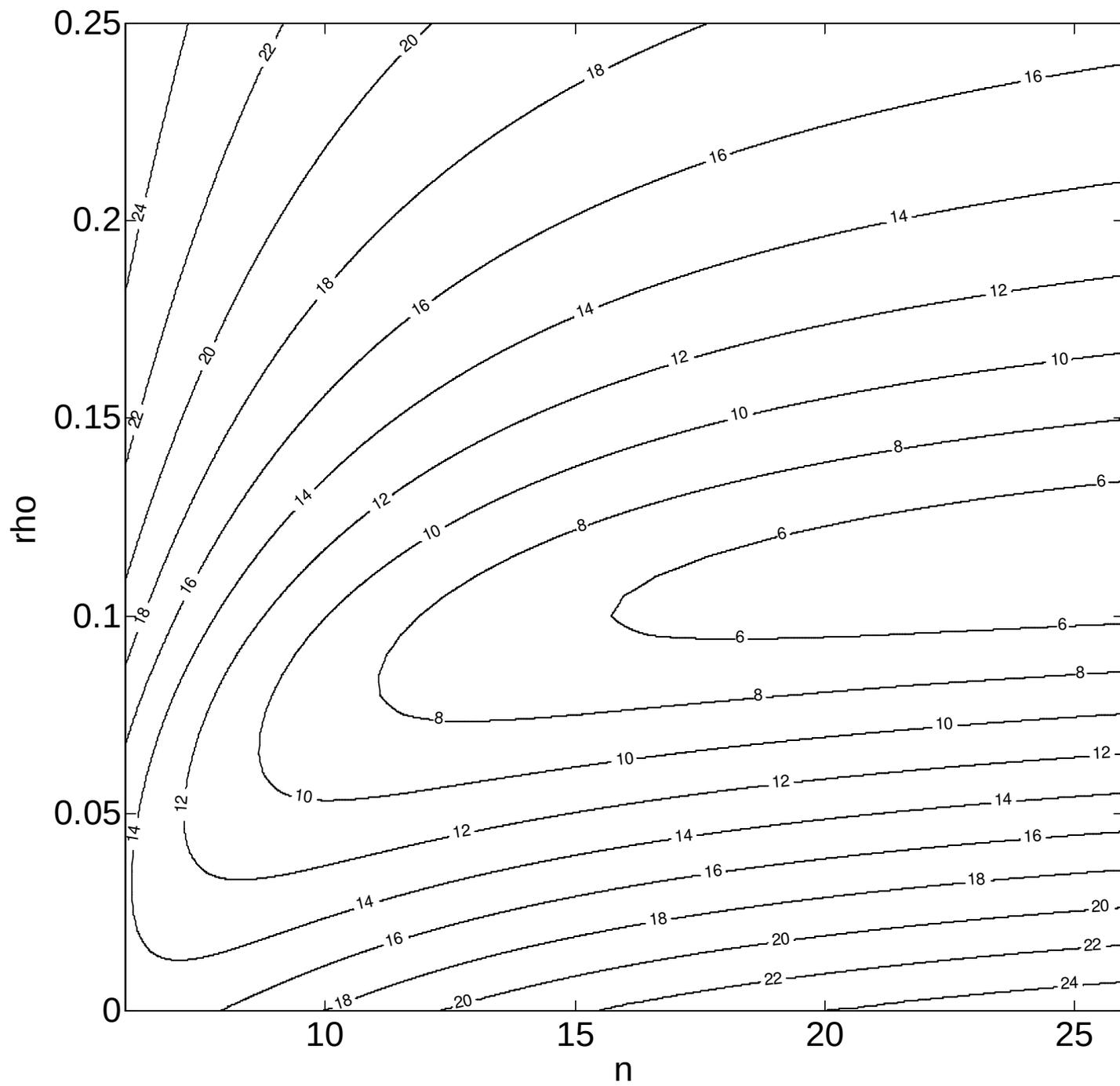
$$j_X = \frac{i_X}{I_{OUT}}$$

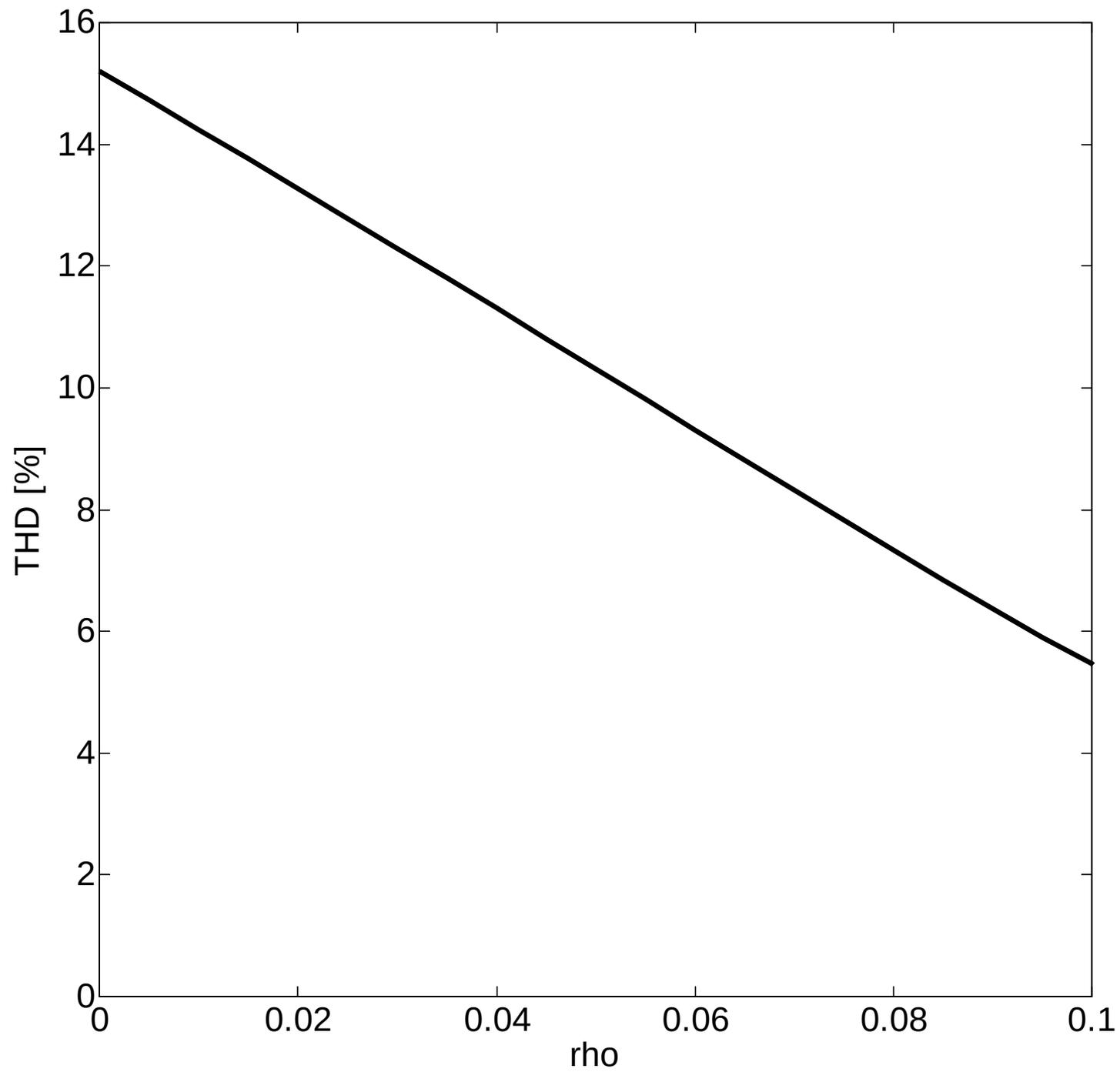
$$\rho = R \frac{I_{OUT}}{V_m}$$

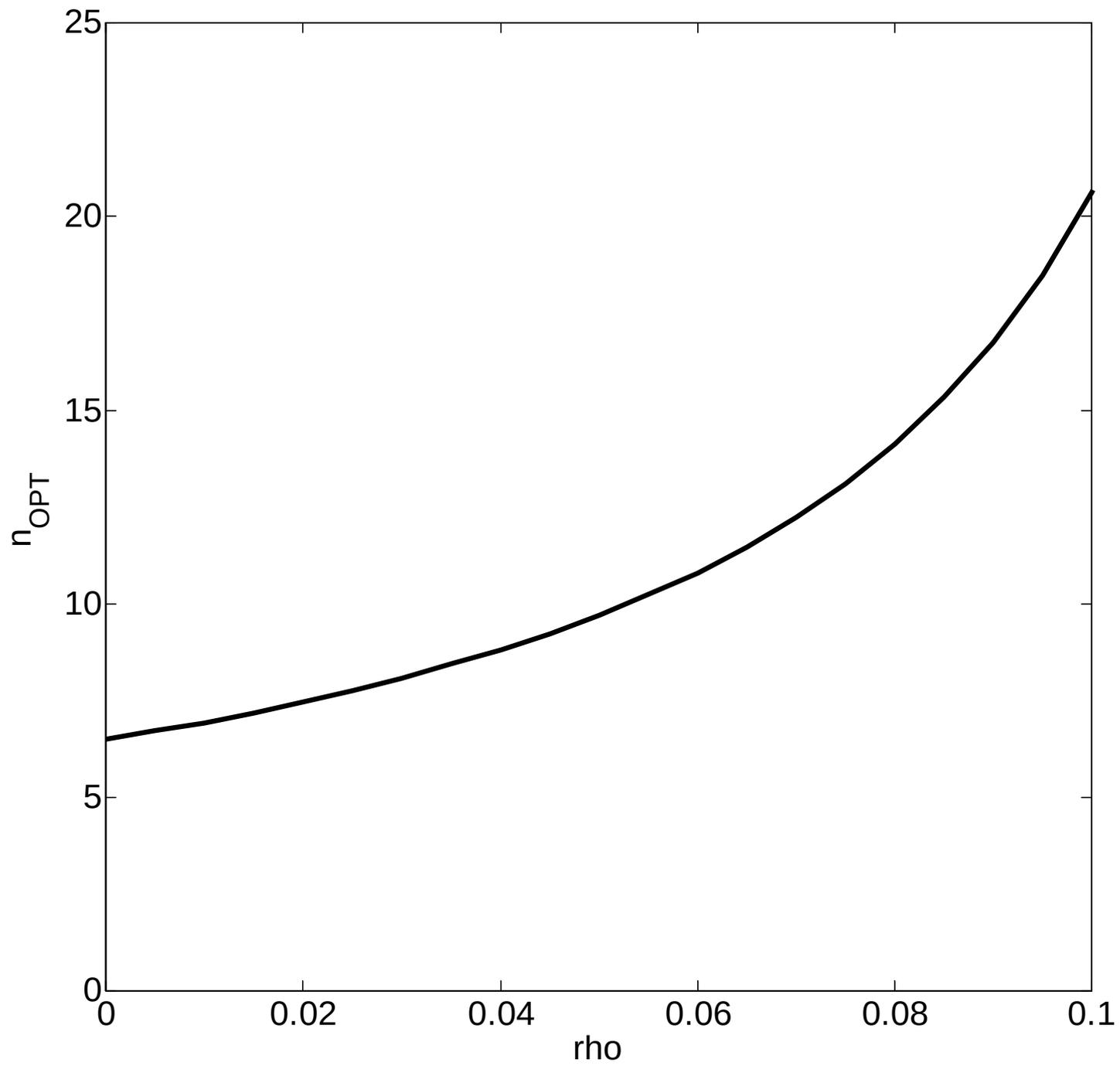
state	DA	DB	DX	DY	j_Y	j_A	j_B
-2	0	1	0	1	$j_Y = -\frac{2n}{n+1}$	0	$-j_Y$
-1	1	1	0	1	$j_Y = \frac{1}{2n\rho} ((n+1)m_{A0} + (n-1)m_{B0})$	$1 + \frac{n+1}{2n} j_Y$	$1 - \frac{n-1}{2n} j_Y$
0	1	1	0	0	$j_Y = 0$	1	1
1	1	1	1	0	$j_Y = \frac{1}{2n\rho} ((n-1)m_{A0} + (n+1)m_{B0})$	$1 + \frac{n-1}{2n} j_Y$	$1 - \frac{n+1}{2n} j_Y$
2	1	0	1	0	$j_Y = \frac{2n}{n+1}$	j_Y	0

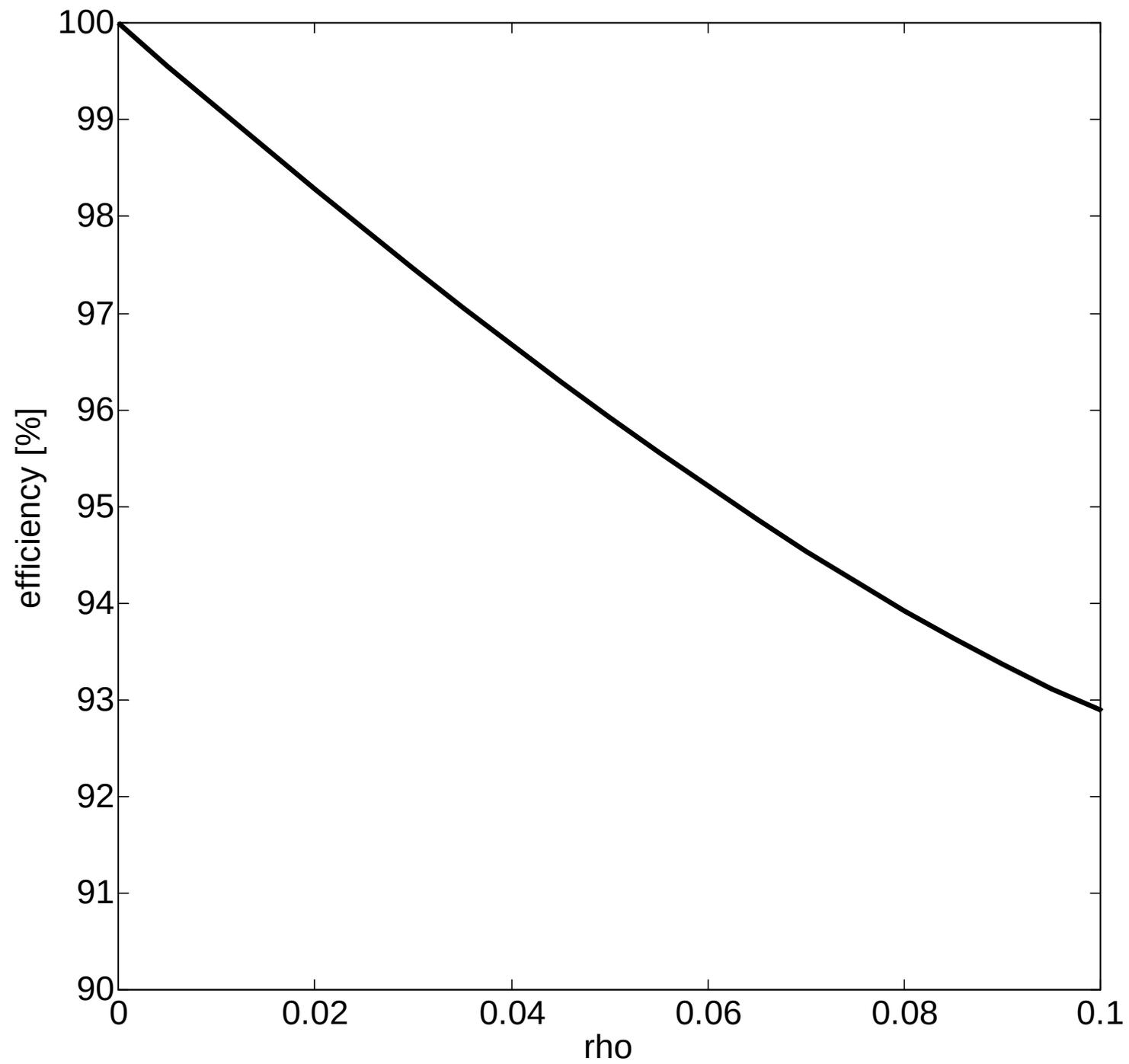
state	condition 1	transition	condition 2	transition
-2	$(n+1)m_{A0} + (n-1)m_{B0} + \frac{4n^2}{n+1}\rho < 0$	-1		
-1	$(n+1)m_{A0} + (n-1)m_{B0} < 0$	0	$(n+1)m_{A0} + (n-1)m_{B0} + \frac{4n^2}{n+1}\rho > 0$	-2
0	$(n-1)m_{A0} + (n+1)m_{B0} < 0$	1	$(n+1)m_{A0} + (n-1)m_{B0} > 0$	-1
1	$(n-1)m_{A0} + (n+1)m_{B0} > 0$	0	$(n-1)m_{A0} + (n+1)m_{B0} - \frac{4n^2}{n+1}\rho < 0$	2
2	$(n-1)m_{A0} + (n+1)m_{B0} - \frac{4n^2}{n+1}\rho > 0$	1		

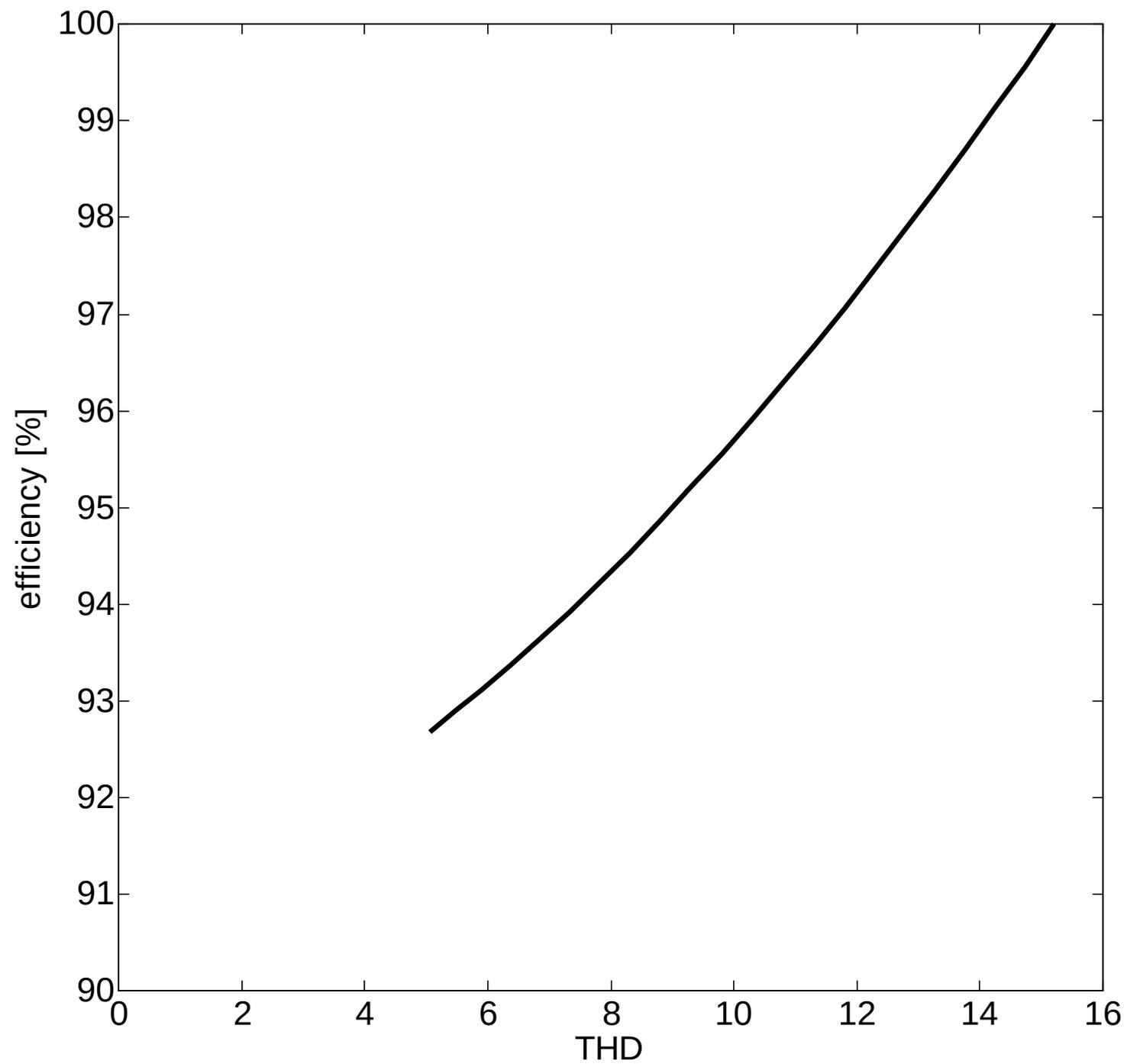
state	m_A	m_B
-2	$-\frac{n-1}{n+1}m_{B0} - \frac{4n^2}{(n+1)^2}\rho$	m_{B0}
-1	m_{A0}	m_{B0}
0	m_{A0}	m_{B0}
1	m_{A0}	m_{B0}
2	m_{A0}	$-\frac{n-1}{n+1}m_{A0} + \frac{4n^2}{(n+1)^2}\rho$

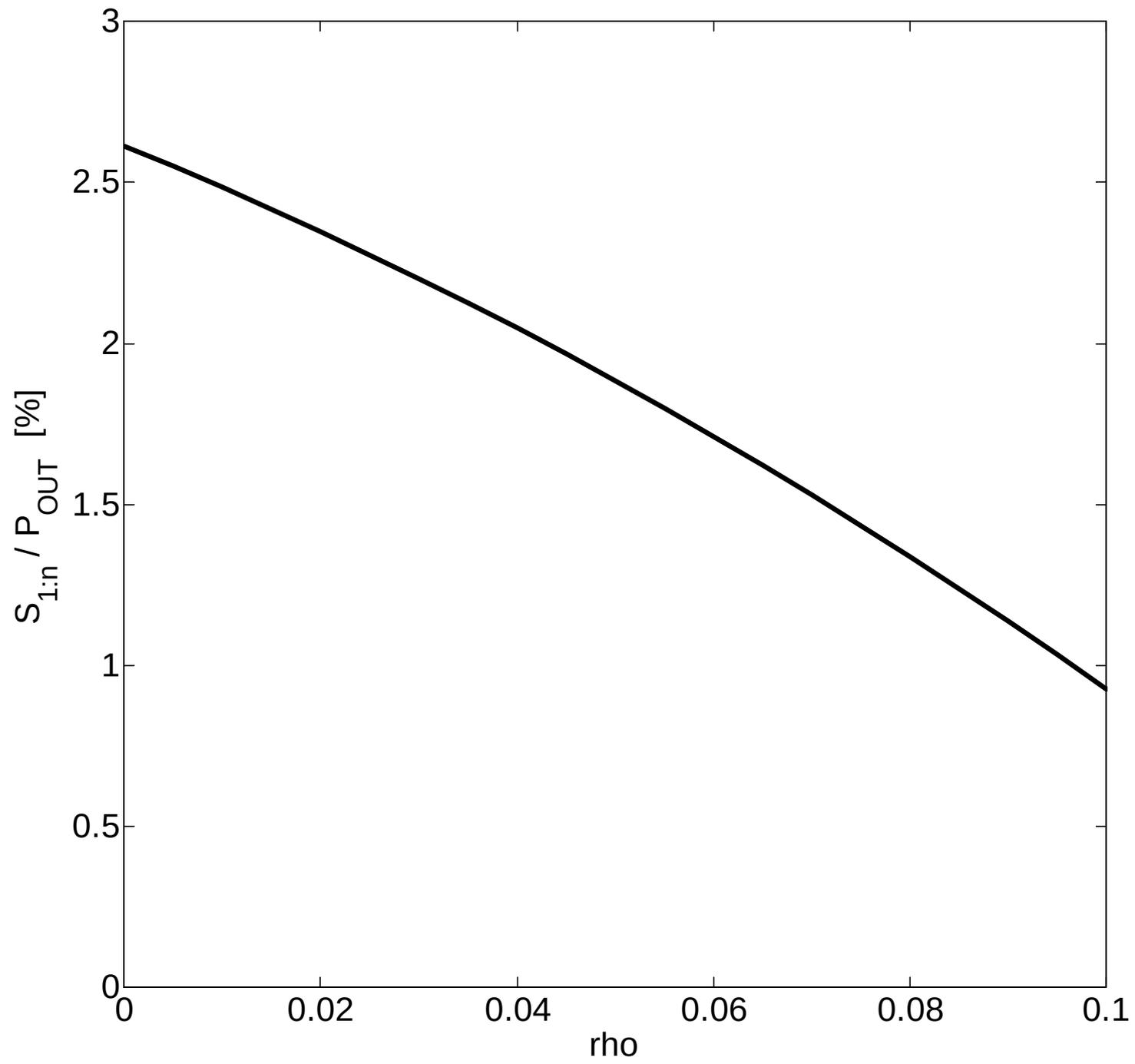


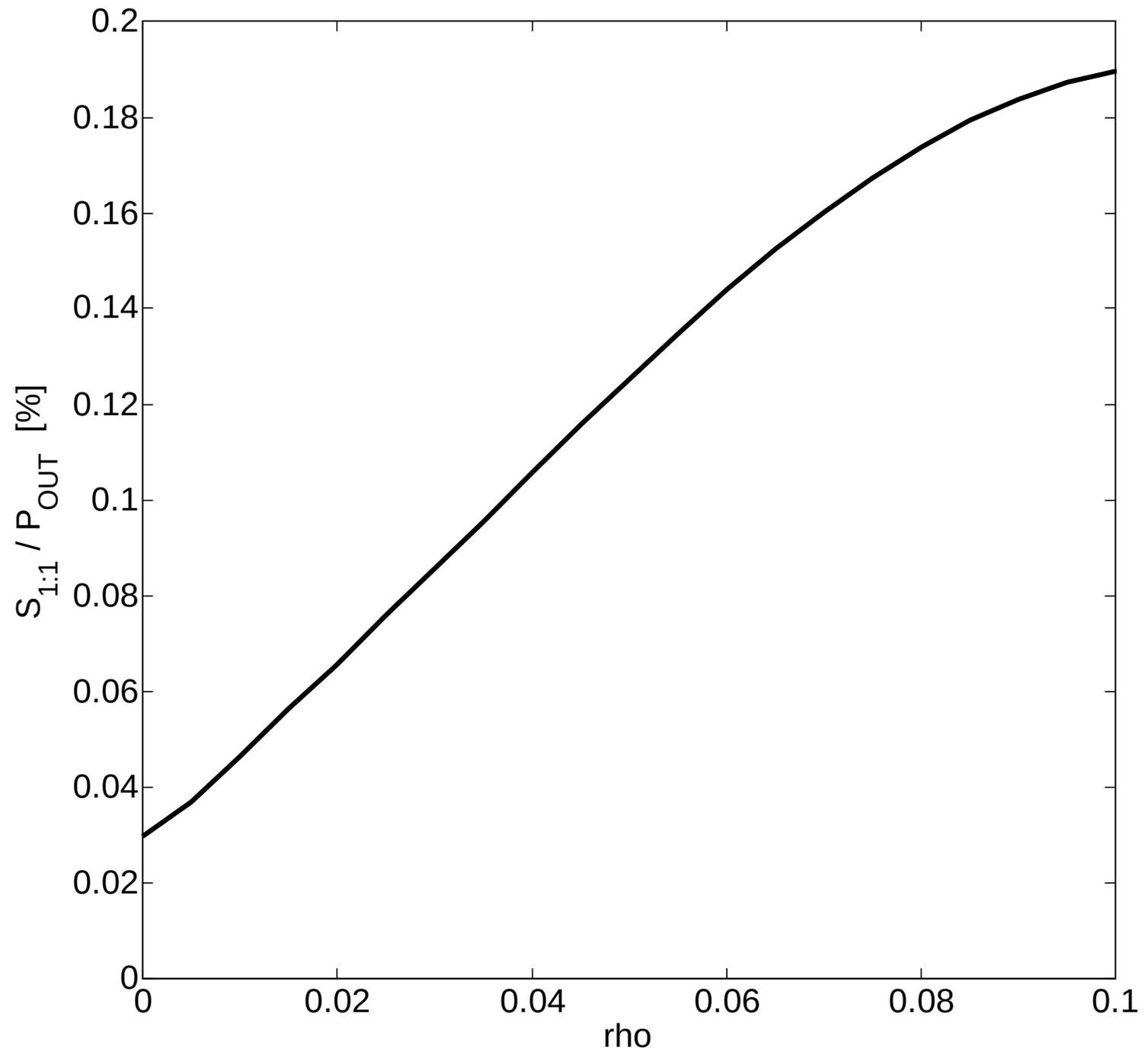


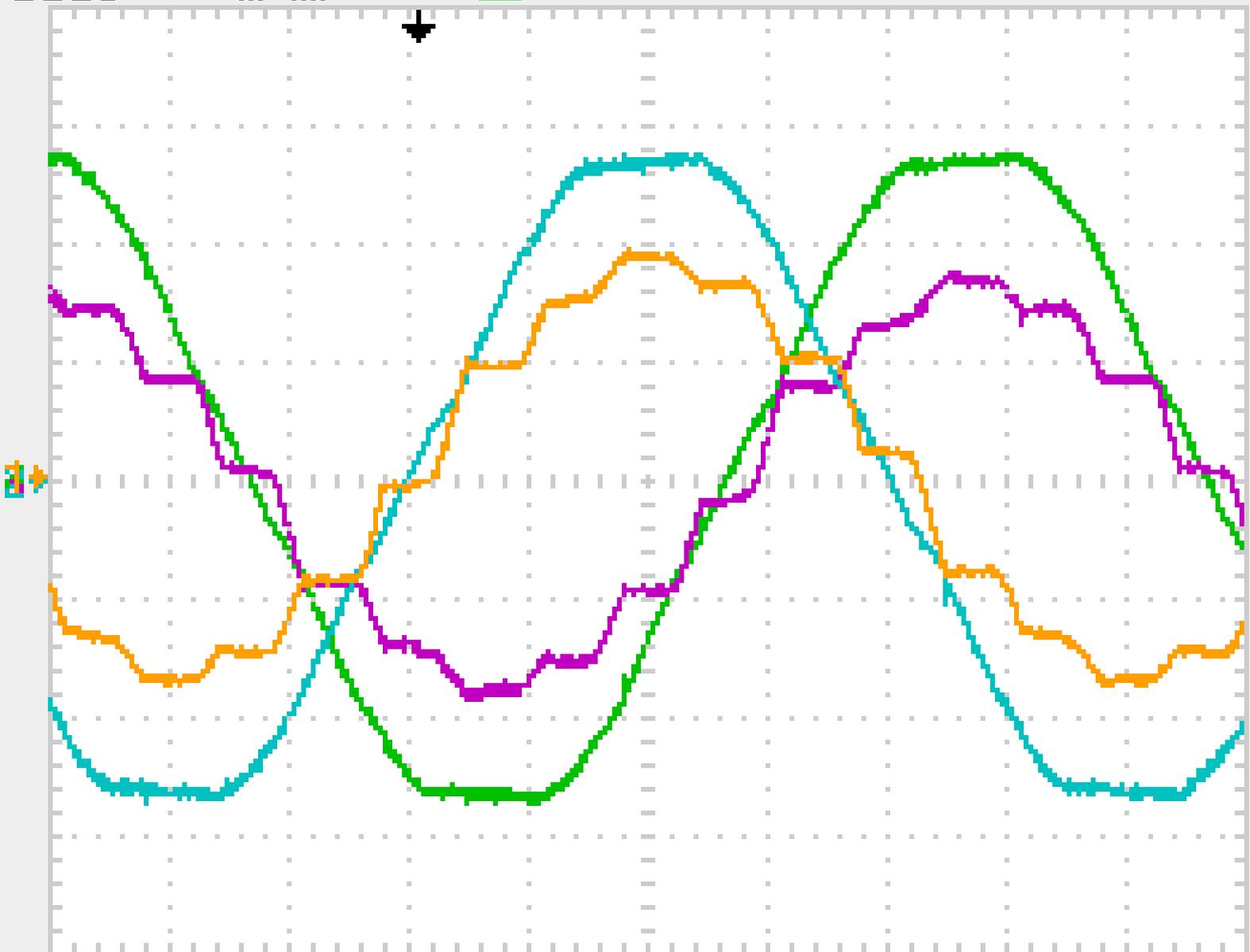












Coupling

DC

BW Limit

On

20MHz

Volts/Div

Coarse

Probe

1X

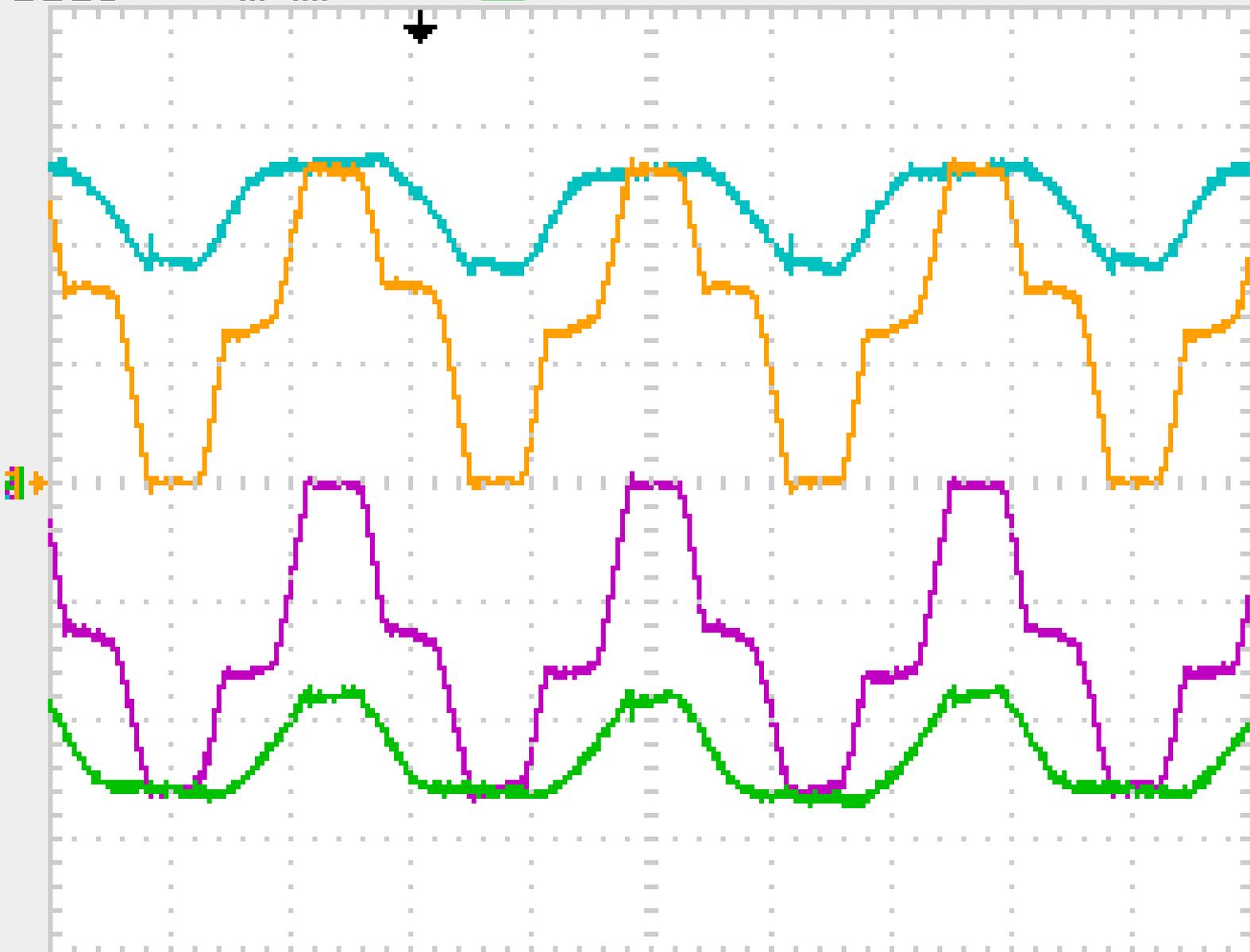
Voltage

Invert

Off

CH1 500mV/Div CH2 50.0V/Div M 2.50ms Ext / 0.00V

CH3 500mV/Div CH4 50.0V/Div 4-Nov-07 16:35 49.9959Hz



Coupling

DC

BW Limit

On

20MHz

Volts/Div

Coarse

Probe

1X

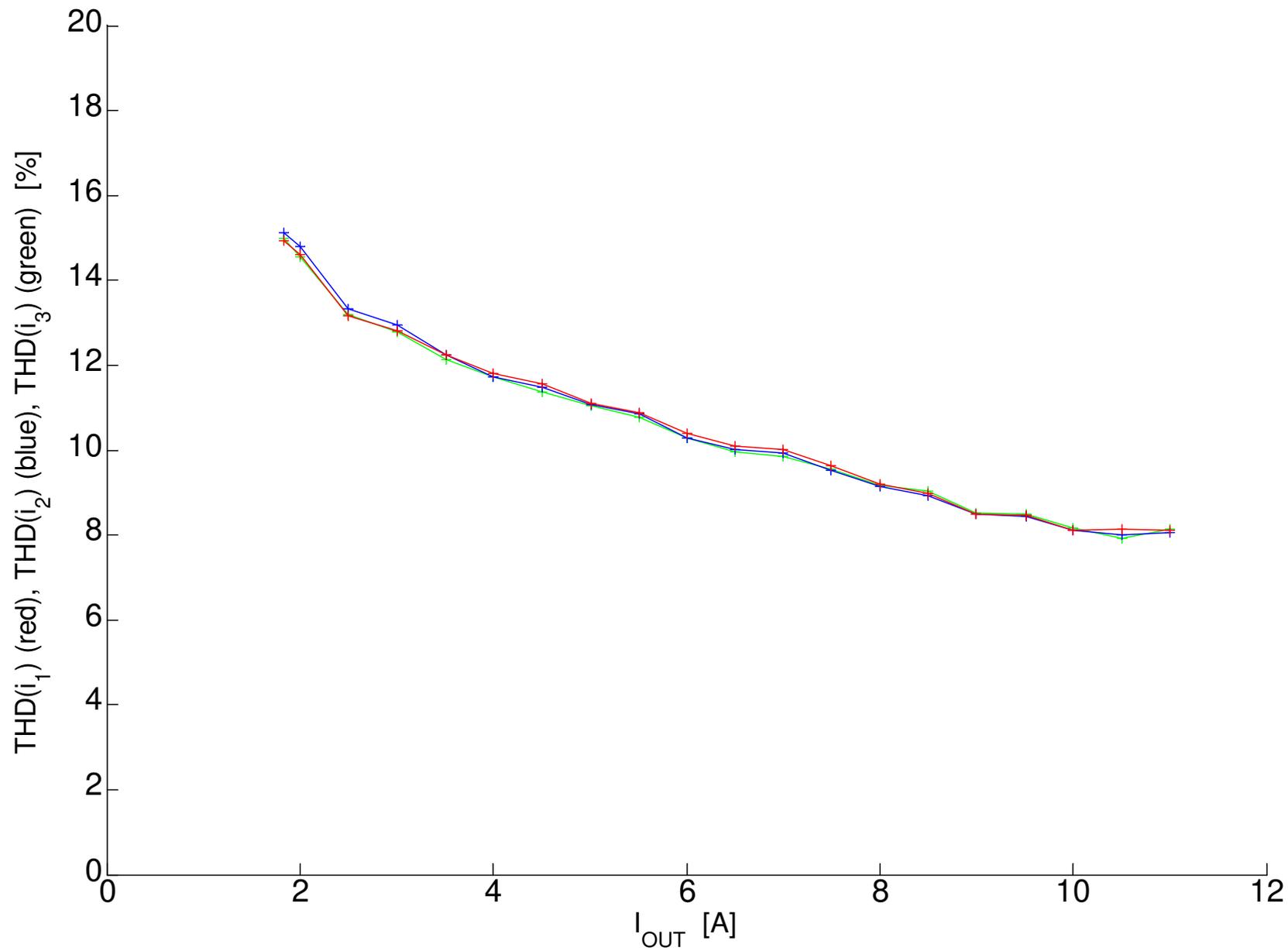
Voltage

Invert

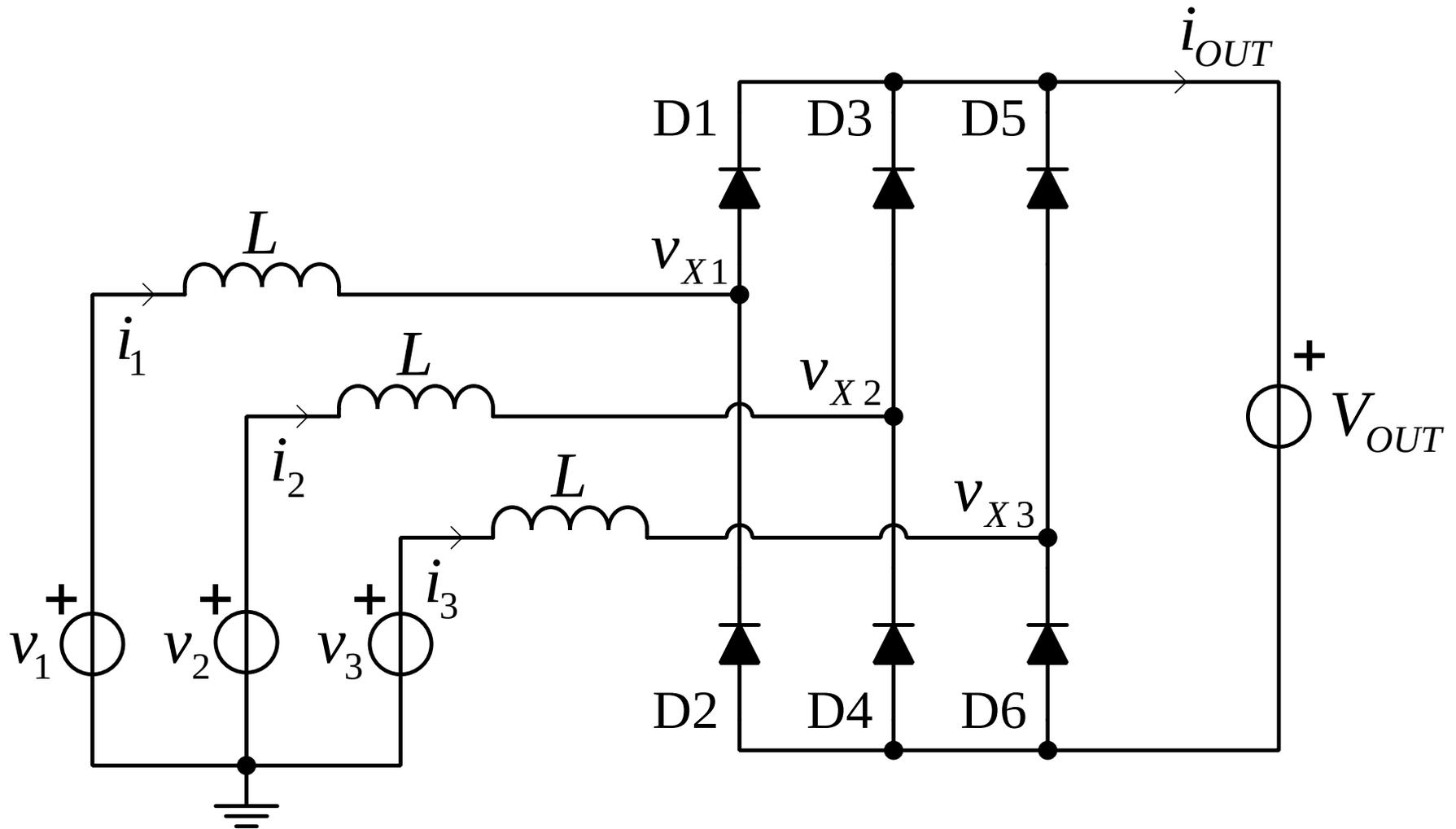
Off

CH1 50.0mV/Div CH2 50.0V/Div M 2.50ms Ext / 0.00V

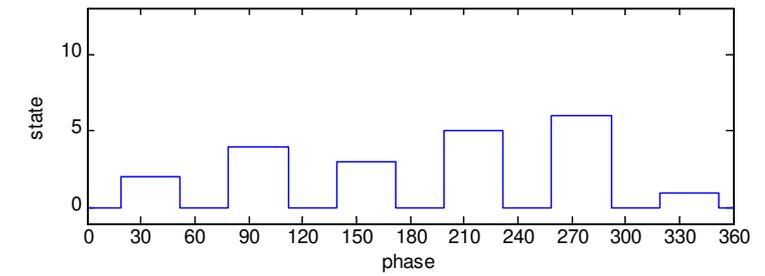
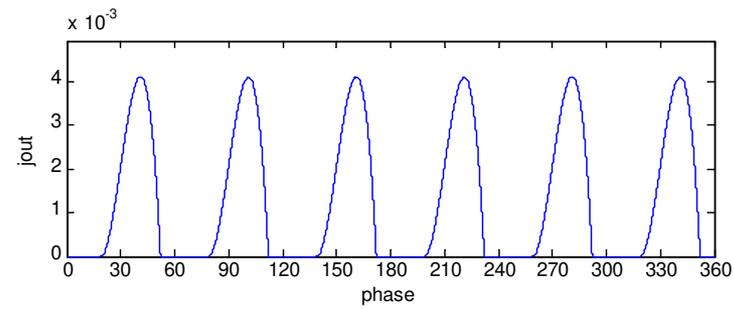
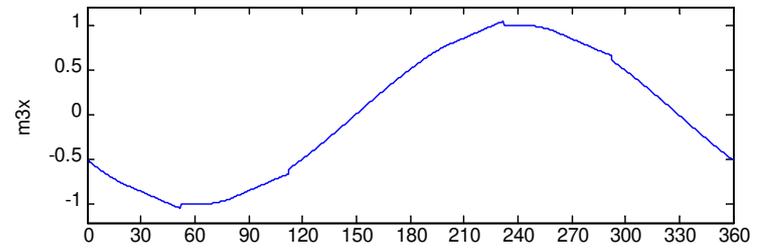
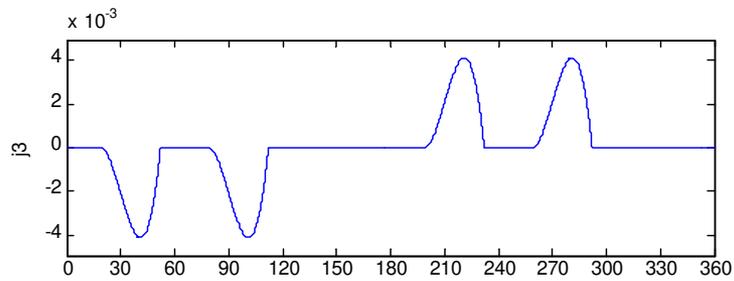
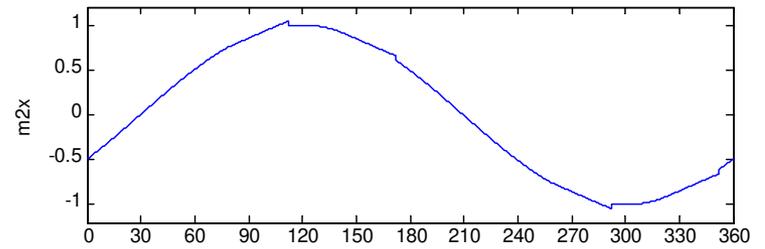
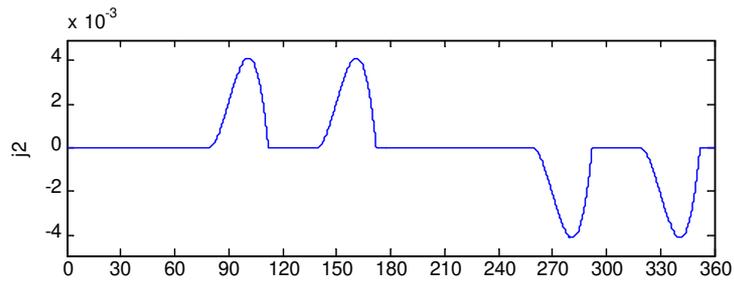
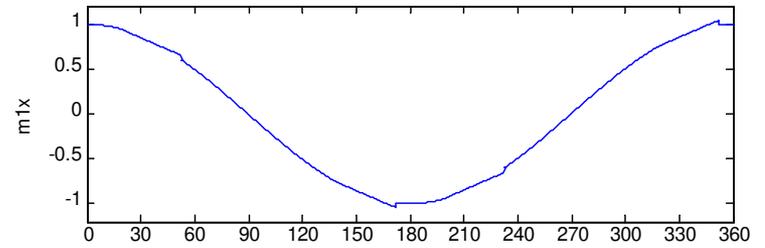
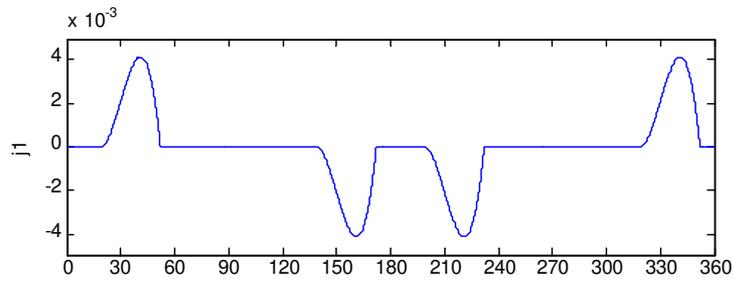
CH3 50.0mV/Div CH4 50.0V/Div 4-Nov-07 17:10 49.9808Hz

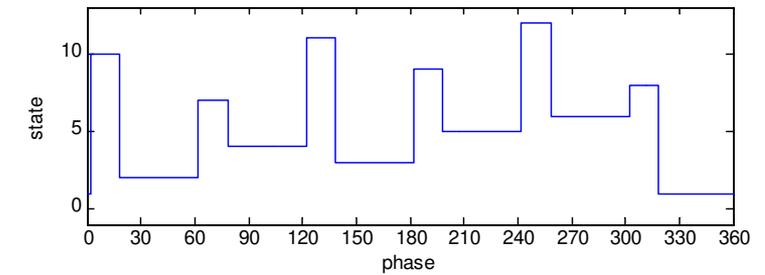
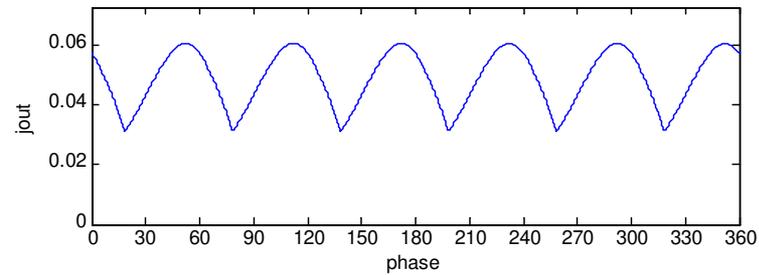
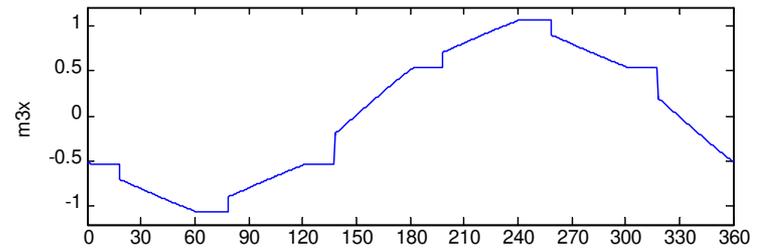
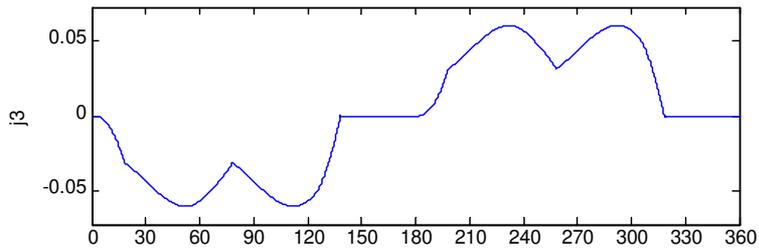
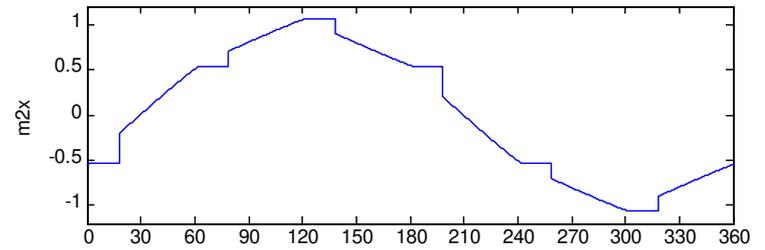
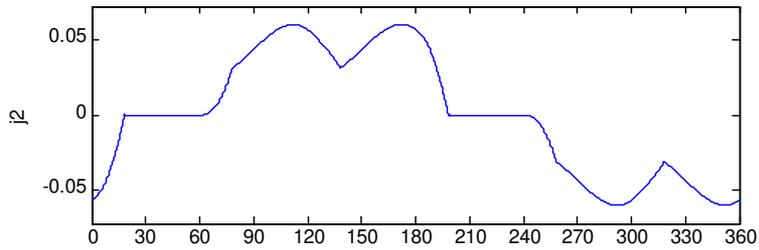
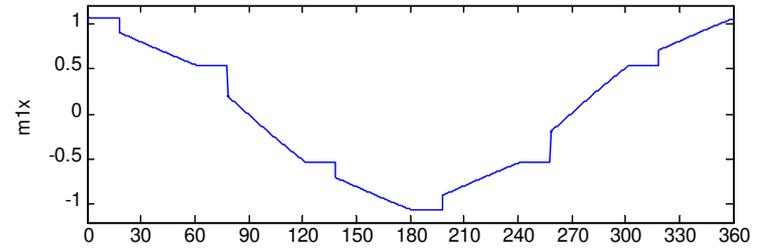
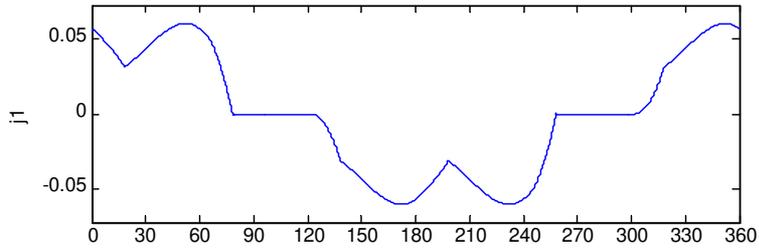


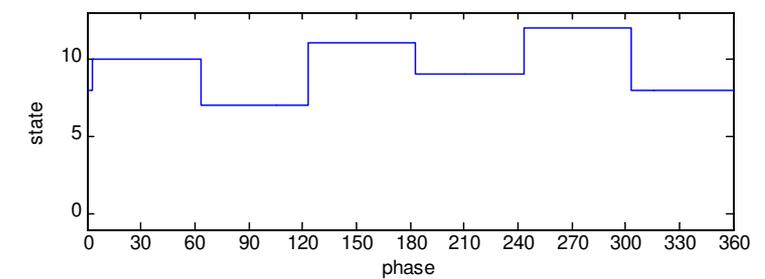
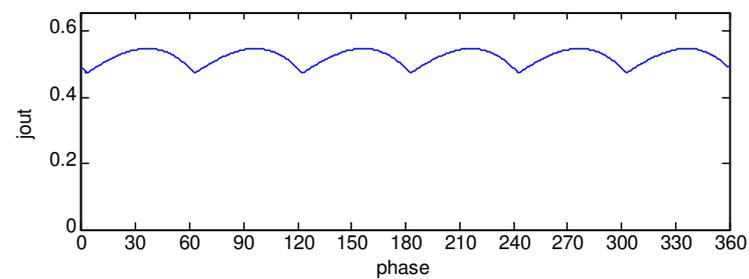
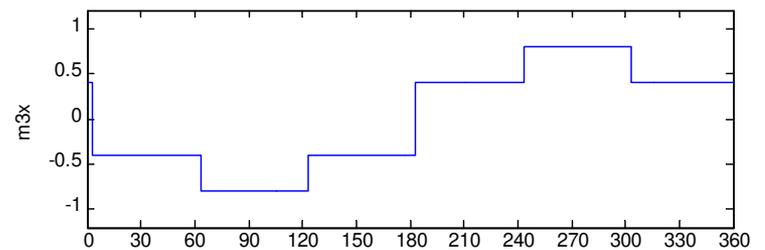
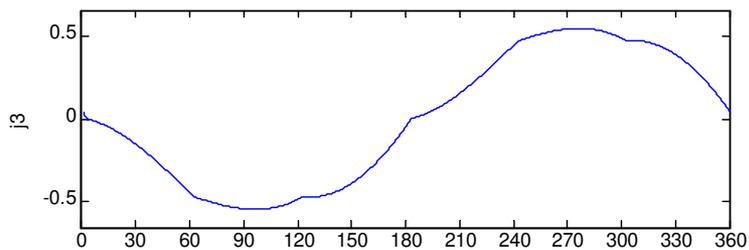
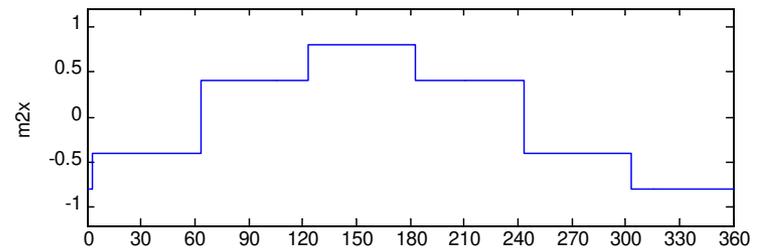
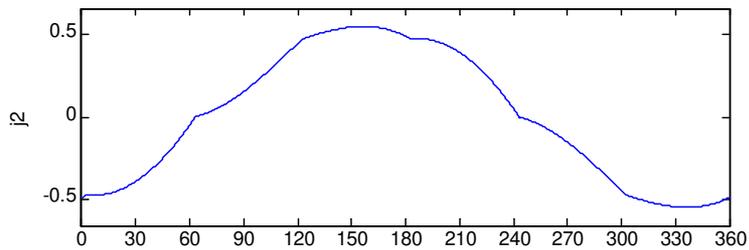
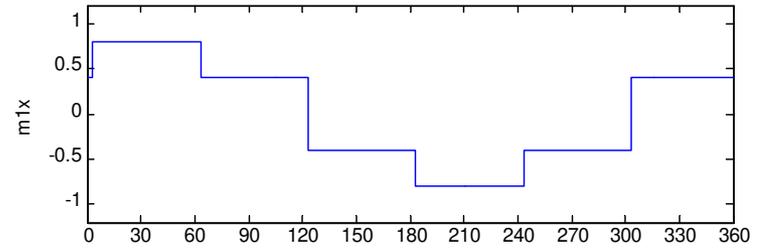
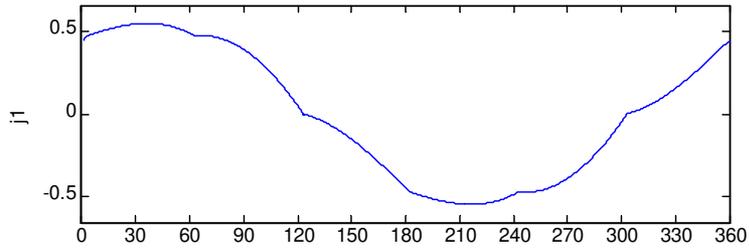
an application in education:

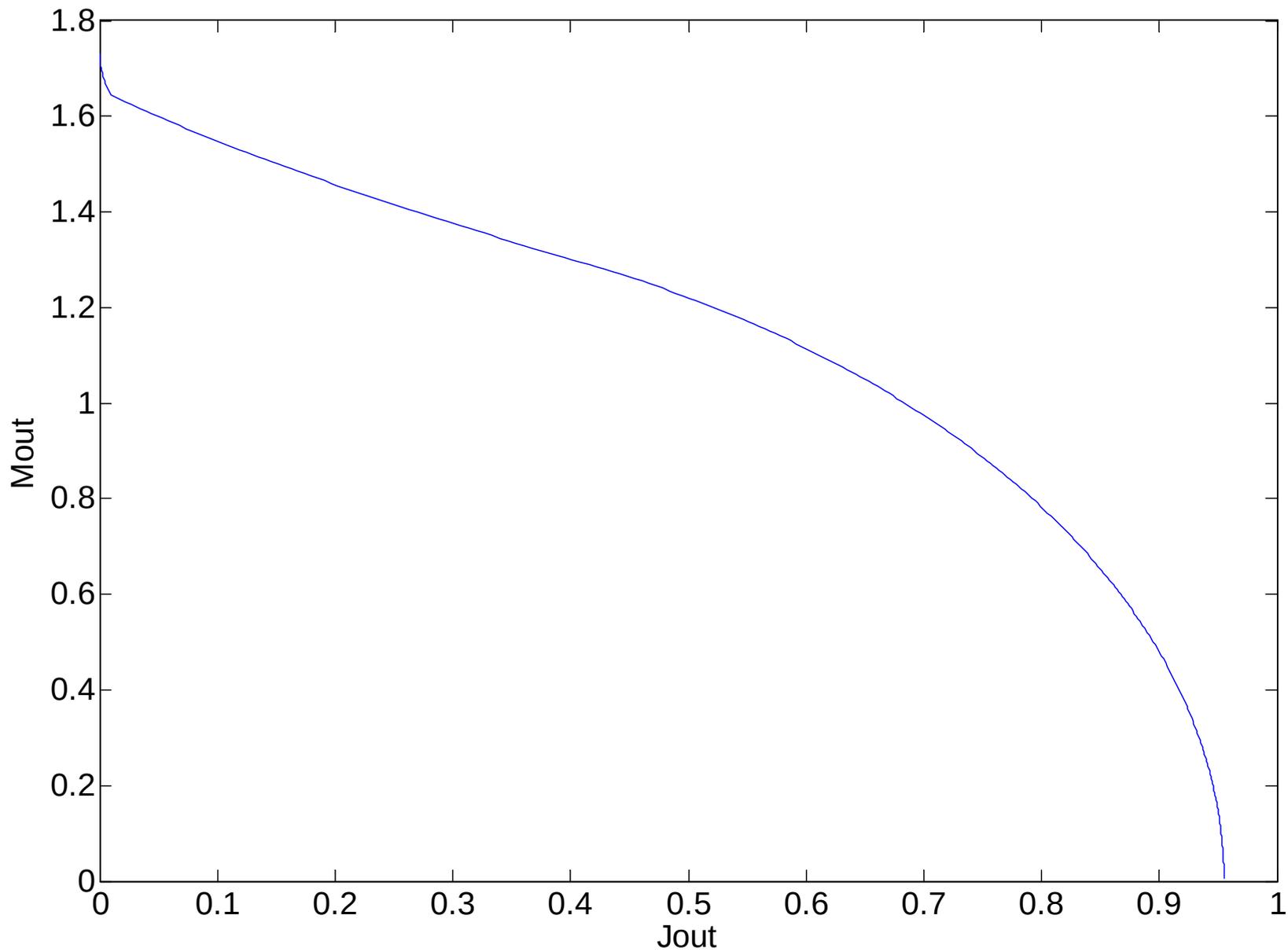


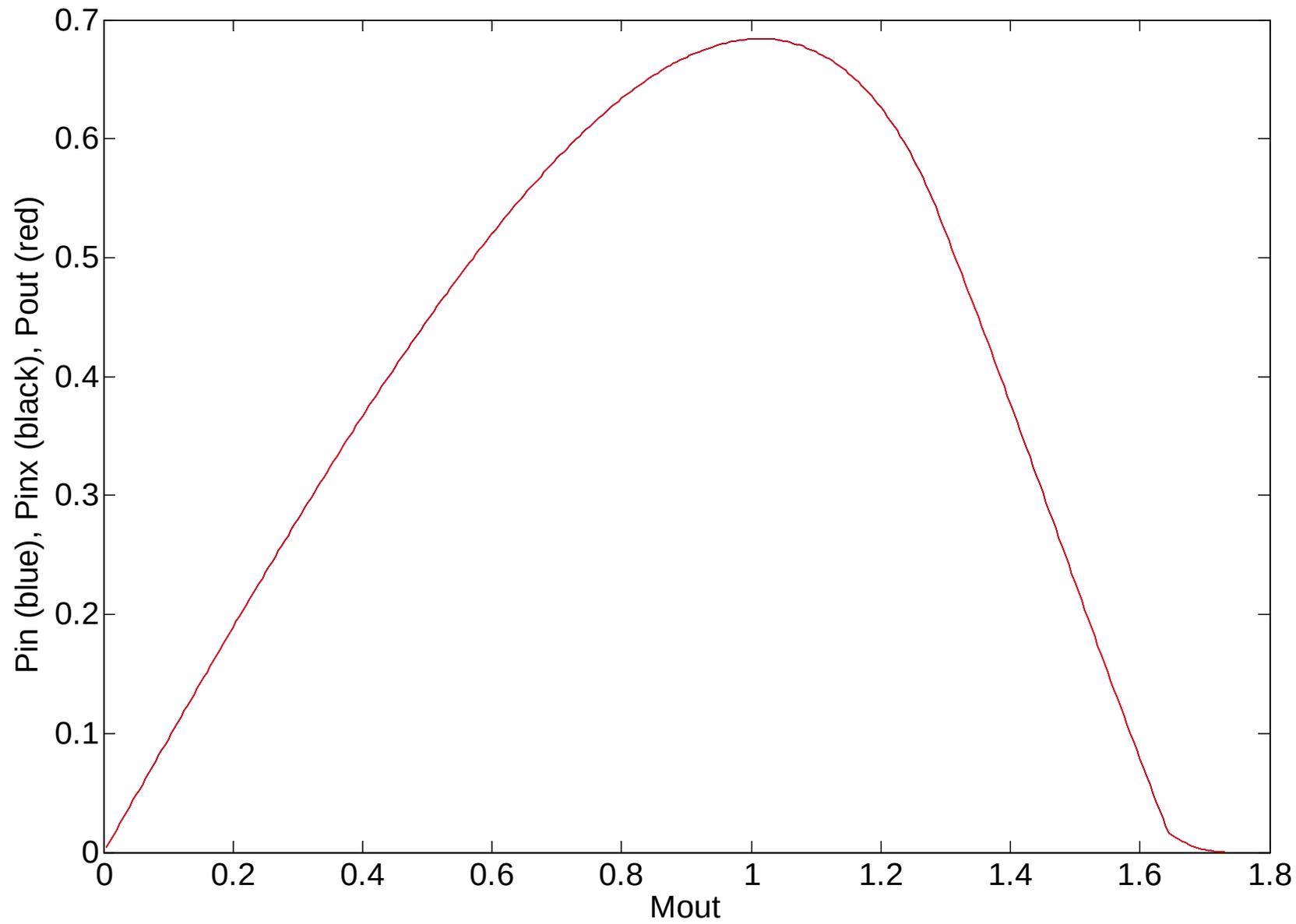
13 states, state transition rules, 2nd order circuit, . . .

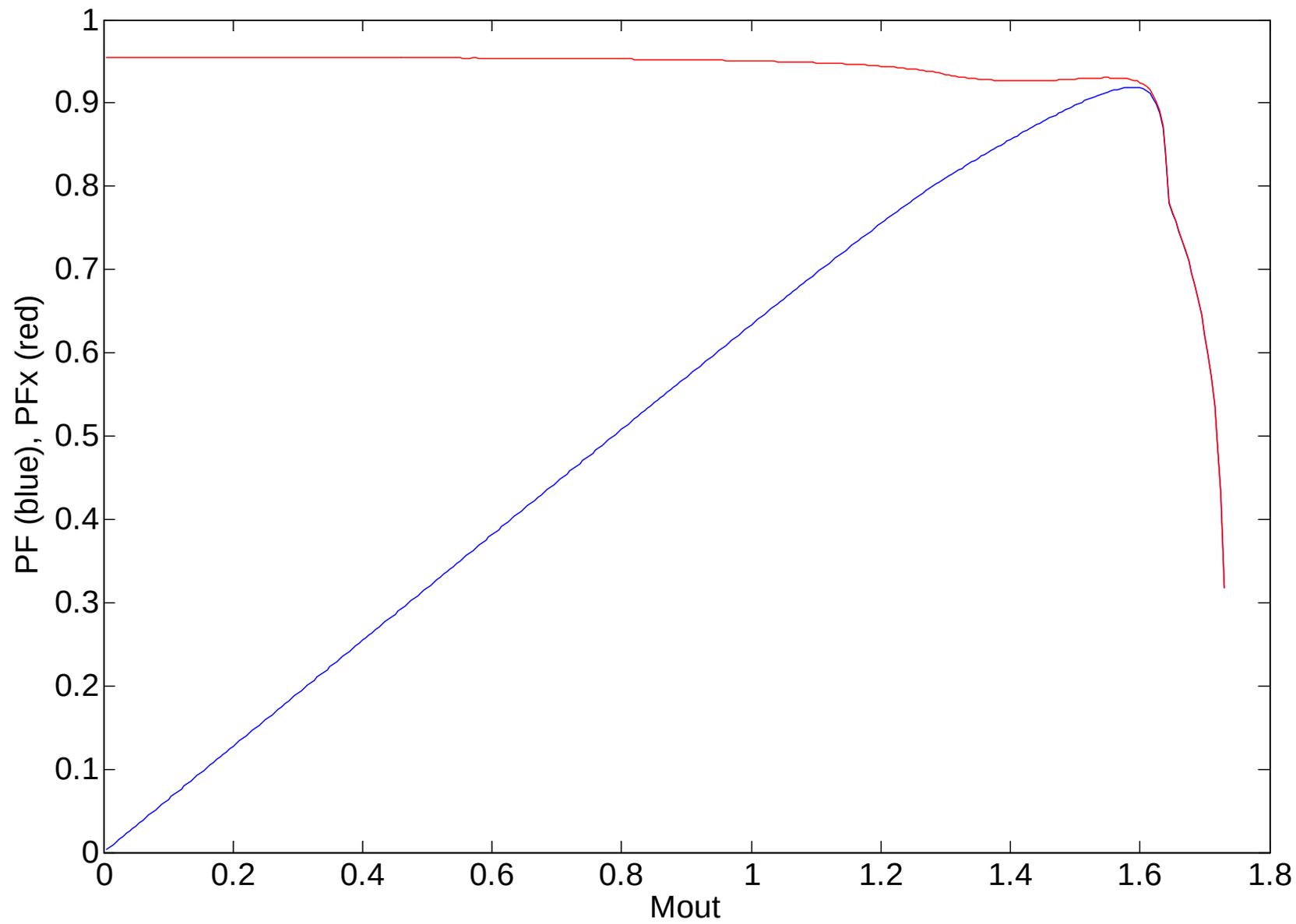


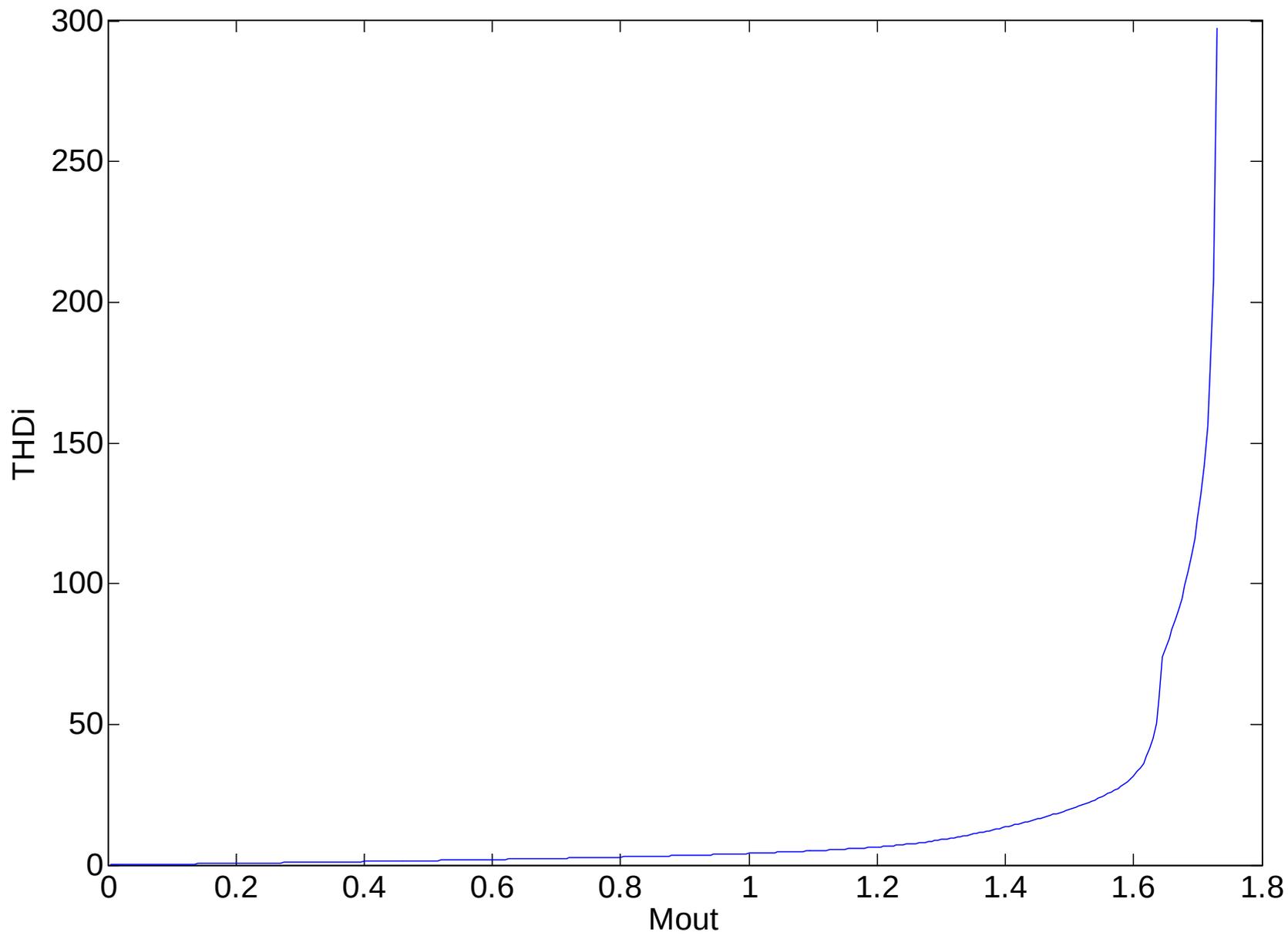


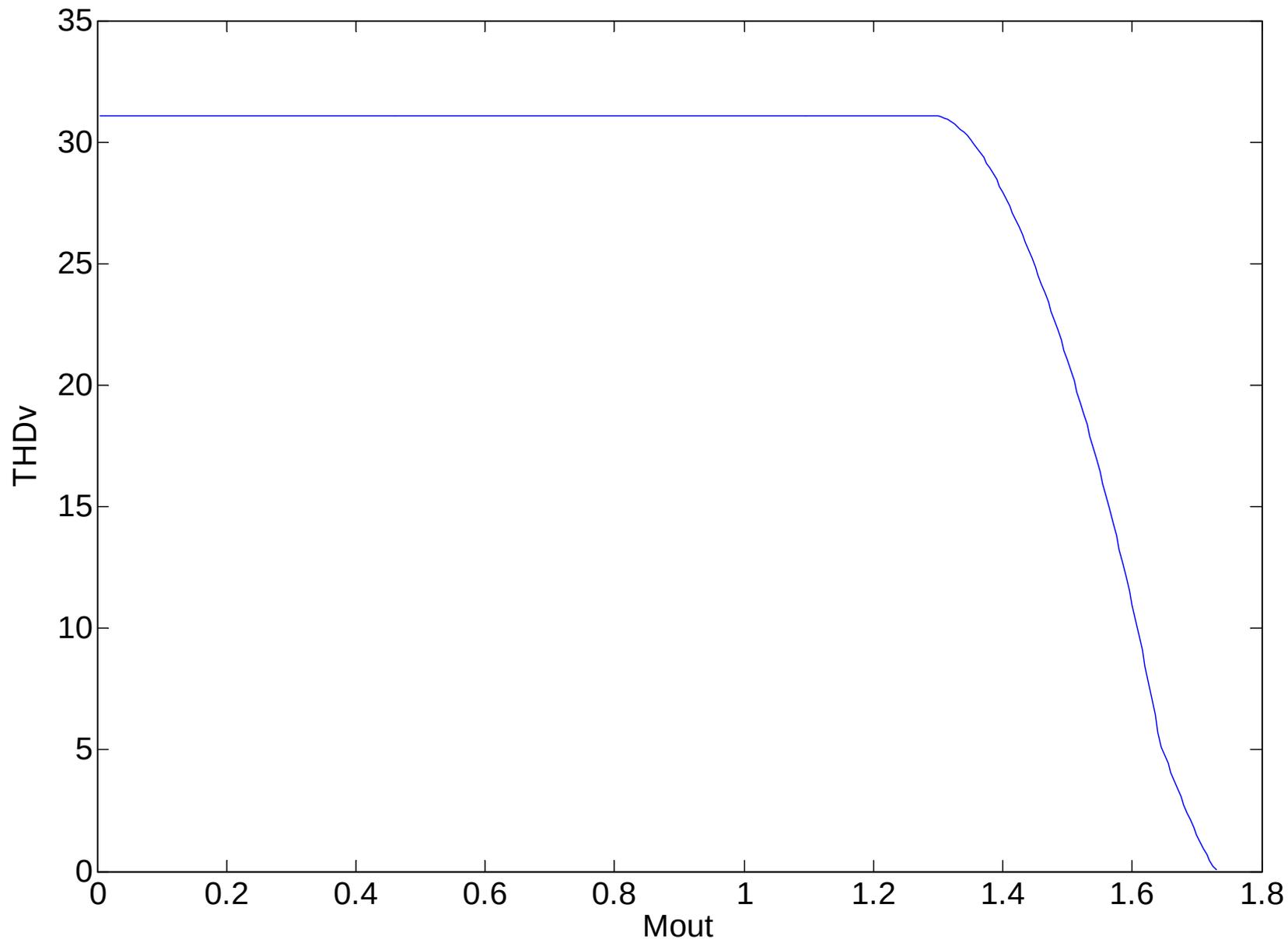


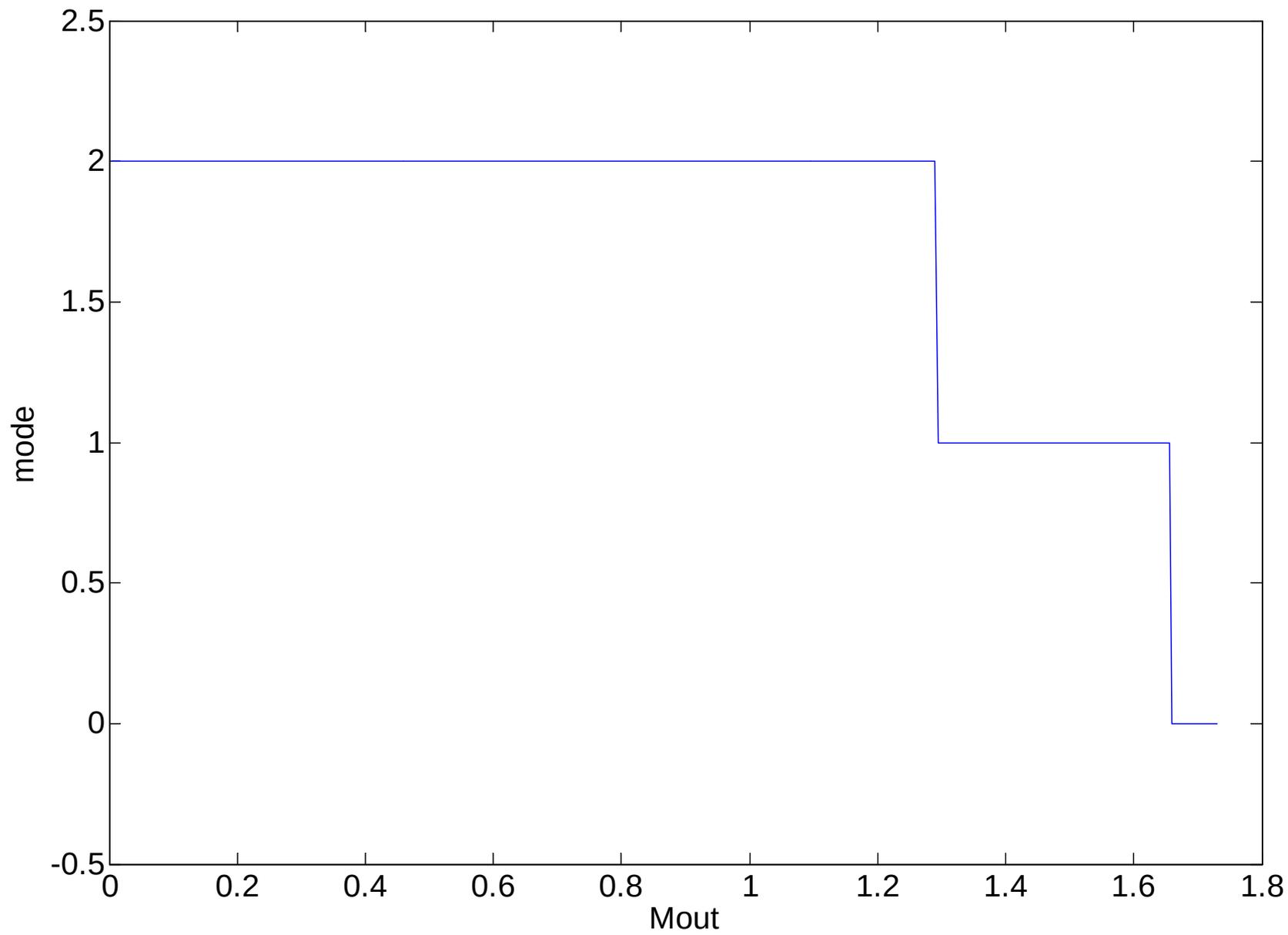












Conclusions

- How to make simulation results general enough?
- Normalization
- Analytical preparation
- Numerical simulation
- Diagrams
- Steady state acceleration algorithm
- Problems with dynamic (algebraic) degeneration
- Analysis of dynamic circuits as resistive circuits
- Application of simulation in education
- Analytical solutions versus numerical solutions?