## Switching Current Injection Device

## aim ...

- current injection devices inject to all three of the phases...
- but only one phase really needs injection!
- is there a way to inject only where needed?
- is there a way to get rid of the current injection device?
- something smaller, lighter, cheaper, ...


## let's get back to basic current injection ...



## as always, assume ...

$$
m_{1}=\cos \left(\omega_{0} t\right)
$$

$$
\begin{aligned}
m_{2} & =\cos \left(\omega_{0} t-\frac{2 \pi}{3}\right) \\
m_{3} & =\cos \left(\omega_{0} t-\frac{4 \pi}{3}\right)
\end{aligned}
$$

## diodes ...








## currents without injection





## what do we need?

## conclusion:

current injection is needed in the phase whose voltage is neither minimal neither maximal in the considered time point

## let's do it!



## how to operate the switches?



## some Boole algebra ...

$$
\begin{aligned}
& s_{1}=\neg d_{1} \wedge \neg d_{2} \\
& s_{2}=\neg d_{3} \wedge \neg d_{4} \\
& s_{3}=\neg d_{5} \wedge \neg d_{6}
\end{aligned}
$$

## and the voltages are defined ...





## $m_{A}$, waveform



## $m_{A}$, analytical

$$
m_{A}=\max \left(m_{1}, m_{2}, m_{3}\right)
$$

$$
\begin{gathered}
m_{A}=M_{A 0}+\sum_{k=1}^{\infty} M_{A, k} \cos \left(3 k \omega_{0} t\right) \\
M_{A 0}=\frac{3 \sqrt{3}}{2 \pi} \\
M_{A, k}=\frac{3 \sqrt{3}}{\pi} \frac{(-1)^{k+1}}{9 k^{2}-1}
\end{gathered}
$$

$m_{A}$, spectrum, real part

$m_{B}$, waveform


$$
m_{B}=\min \left(m_{1}, m_{2}, m_{3}\right)
$$

$$
\begin{gathered}
m_{B}=M_{B 0}+\sum_{k=1}^{\infty} M_{B, k} \cos \left(3 k \omega_{0} t\right) \\
M_{B 0}=-\frac{3 \sqrt{3}}{2 \pi} \\
M_{B, k}=\frac{3 \sqrt{3}}{\pi} \frac{1}{9 k^{2}-1}
\end{gathered}
$$

$m_{B}$, spectrum, real part

$m_{C}$, waveform, this is a new one..


## $m_{C}$, analytical

$$
\begin{gathered}
m_{C}=-m_{A}-m_{B} \\
m_{C}=\sum_{k=1,3,5, \ldots}^{\infty} M_{C, k} \cos \left(3 k \omega_{0} t\right) \\
M_{C, k}=-\frac{6 \sqrt{3}}{\pi} \frac{1}{9 k^{2}-1}
\end{gathered}
$$

this is a (sort of) new spectrum to deal with ...
and there is no DC component in it ...
$m_{C}$, spectrum, real part

since we are already here, $m_{A V}$, waveform


## $m_{A V}$, analytical

$$
\begin{gathered}
m_{A V}=\frac{m_{A}+m_{B}}{2}=-\frac{m_{C}}{2} \\
m_{A V}=\sum_{k=1,3,5, \ldots}^{\infty} M_{A V, k} \cos \left(3 k \omega_{0} t\right) \\
M_{A V, k}=\frac{3 \sqrt{3}}{\pi} \frac{1}{9 k^{2}-1}
\end{gathered}
$$

$m_{A V}$, spectrum, real part

$m_{\text {OUT AC }}$, waveform


$$
\begin{gathered}
m_{\text {OUT AC }}=m_{\text {OUT }}-M_{\text {OUT }}=m_{A}-M_{A, 0}-m_{B}+M_{B, 0} \\
m_{\text {OUT } A C}=\sum_{k=2,4,6, \ldots}^{\infty} M_{\text {OUT } A C, k} \cos \left(3 k \omega_{0} t\right) \\
M_{\text {OUT AC }, k}=-\frac{6 \sqrt{3}}{\pi} \frac{1}{9 k^{2}-1}
\end{gathered}
$$

## $m_{O U T A C}$, spectrum, real part



## we would like to have ...

$$
\begin{gathered}
j_{1}=\cos \left(\omega_{0} t\right) \\
j_{2}=\cos \left(\omega_{0} t-\frac{2 \pi}{3}\right) \\
j_{3}=\cos \left(\omega_{0} t-\frac{4 \pi}{3}\right)
\end{gathered}
$$

a note, again: normalized amplitude is 1 ; if actual amplitude is $I_{m}$, the normalization is

$$
j_{X} \triangleq \frac{i_{X}}{I_{m}}
$$

## and there is a way to get it ...



## analysis . . .

- regarding the inputs, diodes and switches perform useless function
- each phase observes $R_{E}$, which is perfect!
- which of the resistors is the hottest one?
- or better to ask, which one is the coldest?
- let's separate AC and DC, you already know the trick...
... the trick...

hint: $L=R^{2} C$, if $\infty$ is too big


## voila!



## analysis ...

- ... and we have the circuit!
- not a long mathematical derivation?
- actually, it's invented right now, while working on this presentation (April 15, 2012, 00:06:53)
- not the first experience of this kind, 1999, ...


## something similar published in ...

Predrag Pejović

## "A Novel Low Harmonic Three Phase Rectifier"

IEEE Transactions on Circuits and Systems I:
Fundamental Theory and Applications, vol. 49, no. 7, pp. 955-965, July 2002
after lots of trouble ...
although, the derivation presented here is much shorter, and the circuit is slightly different ...

## currents . . .

$$
\begin{gathered}
j_{A}=m_{A} \\
j_{B}=-m_{B} \\
j_{C}=-m_{C}
\end{gathered}
$$

analytical description, spectra, waveforms, ...

$$
\begin{gathered}
j_{I A}=j_{A}-J_{O U T} \\
j_{I B}=J_{O U T}-j_{B} \\
J_{O U T}=\frac{3 \sqrt{3}}{2 \pi}
\end{gathered}
$$

power and efficiency ...

$$
\begin{gathered}
P_{I N}=\frac{3}{2} \\
P_{\text {OUT }}=M_{\text {OUT }} J_{\text {OUT }}=\frac{3 \sqrt{3}}{\pi} \times \frac{3 \sqrt{3}}{2 \pi}=\frac{27}{2 \pi^{2}} \\
P_{I N J}=P_{I N}-P_{\text {OUT }}=\frac{3 \pi^{2}-27}{2 \pi^{2}} \\
\eta=\frac{P_{\text {OUT }}}{P_{I N}}=\frac{9}{\pi^{2}} \approx 91.19 \%
\end{gathered}
$$

already familiar with the results?

## remains to be done ...

- how hot each resistor is?
- other current injection networks?
- the third harmonic current injection?
- how to build the switching current injection device?
- in the meantime, some waveforms and spectra...
$j_{A}, j_{B}$, and $j_{C}$





## $j_{A}$, analytical

$$
\begin{gathered}
j_{A}=\max \left(j_{1}, j_{2}, j_{3}\right) \\
j_{A}=J_{A 0}+\sum_{k=1}^{\infty} J_{A, k} \cos \left(3 k \omega_{0} t\right) \\
J_{A 0}=\frac{3 \sqrt{3}}{2 \pi}=J_{O U T} \\
J_{A, k}=\frac{3 \sqrt{3}}{\pi} \frac{(-1)^{k+1}}{9 k^{2}-1}
\end{gathered}
$$

the same as $m_{A}$
$j_{A}$, spectrum, real part


## $j_{B}$, analytical

$$
\begin{gathered}
j_{B}=-\min \left(j_{1}, j_{2}, j_{3}\right) \\
j_{B}=J_{B 0}+\sum_{k=1}^{\infty} J_{B, k} \cos \left(3 k \omega_{0} t\right) \\
J_{B 0}=\frac{3 \sqrt{3}}{2 \pi}=J_{O U T} \\
J_{B, k}=-\frac{3 \sqrt{3}}{\pi} \frac{1}{9 k^{2}-1}
\end{gathered}
$$

the same as $-m_{B}$

## $j_{B}$, spectrum, real part



## $j_{C}$, analytical

$$
\begin{gathered}
j_{C}=j_{A}-j_{B} \\
j_{C}=\sum_{k=1,3,5, \ldots}^{\infty} J_{C, k} \cos \left(3 k \omega_{0} t\right) \\
J_{C, k}=\frac{6 \sqrt{3}}{\pi} \frac{1}{9 k^{2}-1}
\end{gathered}
$$

proportional to $m_{A V}$ and $-m_{C}$; going to be important

## $j_{C}$, spectrum, real part



## $j_{I A}, j_{I B}$, and $j_{C}$





## currents ...

$$
\begin{gathered}
j_{I A}=j_{A}-J_{O U T} \\
j_{I B}=J_{O U T}-j_{B} \\
J_{O U T}=\frac{3 \sqrt{3}}{2 \pi}
\end{gathered}
$$

analytical description, spectra, ...

## $j_{I A}$, analytical

$$
\begin{gathered}
j_{I A}=\max \left(j_{1}, j_{2}, j_{3}\right)-\frac{3 \sqrt{3}}{\pi} \\
j_{I A}=\sum_{k=1}^{\infty} J_{A, k} \cos \left(3 k \omega_{0} t\right) \\
J_{I A, k}=\frac{3 \sqrt{3}}{\pi} \frac{(-1)^{k+1}}{9 k^{2}-1}
\end{gathered}
$$

## $j_{I A}$, spectrum, real part



## $j_{I B}$, analytical

$$
\begin{gathered}
j_{I B}=\frac{3 \sqrt{3}}{2 \pi}-\min \left(j_{1}, j_{2}, j_{3}\right) \\
j_{I B}=\sum_{k=1}^{\infty} J_{B, k} \cos \left(3 k \omega_{0} t\right) \\
J_{I B, k}=\frac{3 \sqrt{3}}{\pi} \frac{1}{9 k^{2}-1}
\end{gathered}
$$

## $j_{I B}$, spectrum, real part



## detour: $j_{R 1}, j_{R 2}$, and $j_{R 3}$





## $j_{R 1}$, spectrum, real part


$j_{S 1}, j_{S 2}$, and $j_{S 3}$



$j_{S 1}$, spectrum, real part


## $j_{1}, j_{2}$, and $j_{3}$





## how $j_{1}$ is obtained?




how $j_{1}$ is obtained: spectral approach


## how $j_{2}$ is obtained?





## how $j_{3}$ is obtained?





## $j_{\text {odd }}$ and $j_{\text {even }}$

matter of convenience:

$$
\begin{aligned}
& j_{I A}=j_{\text {odd }}+j_{\text {even }} \\
& j_{I B}=j_{\text {odd }}-j_{\text {even }} \\
& j_{\text {odd }}=\frac{j_{I A}+j_{I B}}{2} \\
& j_{\text {even }}=\frac{j_{I A}-j_{I B}}{2}
\end{aligned}
$$

## $j_{\text {odd }}$, spectrum

$$
\begin{gathered}
j_{\text {odd }}=\sum_{k=1,3,5 \ldots}^{\infty} J_{o d d, k} \cos \left(3 k \omega_{0} t\right) \\
J_{\text {odd }, k}=\frac{3 \sqrt{3}}{\pi} \frac{1}{9 k^{2}-1}
\end{gathered}
$$

## $j_{\text {odd }}$, spectrum, real part



## $j_{\text {even }}$, spectrum

$$
\begin{gathered}
j_{\text {even }}=\sum_{k=2,4,6 \ldots}^{\infty} J_{\text {even }, k} \cos \left(3 k \omega_{0} t\right) \\
J_{\text {even }, k}=-\frac{3 \sqrt{3}}{\pi} \frac{1}{9 k^{2}-1}
\end{gathered}
$$

proportional to $m_{O U T ~ A C}$; going to be important

## $j_{\text {even }}$, spectrum, real part


so, we have SCIN $\# 0$


## SCIN \#0, heat $\ldots$

wxMaxima ... I won't be able to complete this job manually, it's too boring ...

$$
\begin{gathered}
J_{I A R M S}=J_{I B R M S}=\frac{\sqrt{4 \pi^{2}+3 \pi \sqrt{3}-54}}{2 \pi \sqrt{2}} \\
P_{\text {vertical } R_{E}}=\left(J_{I A R M S}\right)^{2}=\frac{4 \pi^{2}+3 \pi \sqrt{3}-54}{8 \pi^{2}} \approx 0.0228
\end{gathered}
$$

vertically placed $R_{E}$ resistors are not so hot

$$
\begin{gathered}
J_{C R M S}=\sqrt{\frac{2 \pi-3 \sqrt{3}}{4 \pi}} \\
P_{\text {horizontal } R_{E}}=\left(J_{C R M S}\right)^{2}=\frac{2 \pi-3 \sqrt{3}}{4 \pi} \approx 0.0865
\end{gathered}
$$

horizontally placed $R_{E}$ is much hotter, almost 4 times; why $4 \times$ ?

## a few words about SCIN \#0

- transformer not required
- too many resistors
- one resistor takes the most of the power


## some relations ...

$$
\begin{gathered}
\frac{m_{A V}}{j_{o d d}}=1 \\
\frac{m_{C}}{j_{o d d}}=-2 \\
\frac{m_{\text {OUT } A C}}{j_{\text {even }}}=2
\end{gathered}
$$

to be used while creating new current injection networks ...

## SCIN \#1, separatism



## $R_{1}$ and $R_{2}$

$$
\begin{gathered}
R_{1}=\frac{m_{A V}-m_{C}}{2 j_{\text {odd }}} R_{E}=\frac{3}{2} R_{E} \\
R_{2}=\frac{m_{\text {OUT } A C}}{j_{\text {even }}}=2 R_{E}
\end{gathered}
$$

and now you know why "some relations" were needed for

## some power accounting, again ...

$$
\begin{gathered}
J_{o d d R M S}=\frac{\sqrt{2 \pi-3 \sqrt{3}}}{4 \sqrt{\pi}} \\
P_{1}=\frac{3}{2}\left(J_{o d d ~ R M S}\right)^{2}=\frac{3}{8}\left(2-\frac{3 \sqrt{3}}{\pi}\right) \approx 0.1298 \\
J_{\text {even } R M S}=\frac{\sqrt{3}}{4 \pi} \sqrt{2 \pi^{2}+3 \pi \sqrt{3}-36} \\
P_{2}=\frac{3\left(2 \pi^{2}+3 \pi \sqrt{3}-36\right)}{8 \pi^{2}} \approx 0.0024
\end{gathered}
$$

dissipation is dominant on $R_{1}$

## an idea makes a new idea ...

- power on $R_{2}$ is small...
- let's get rid of $R_{2}$ !
- result: the same as the $3^{\text {rd }}$ harmonic CIN $\# 3$ for $Q=0$, $T H D \approx 4 \%$
- could we get $Q \neq 0$ ?
- definitely!
- do we need it?
- well, maybe for passive resistance emulation, to be talked about later


## SCIN $\# 2$, the $3^{\text {rd }}$ harmonic one



## SCIN \#2, parameters

$$
\begin{gathered}
3 \omega_{0}=\frac{1}{\sqrt{L C}} \\
j_{C}=\frac{1}{2} \cos \left(3 \omega_{0} t\right) \\
M_{A V, 1}-M_{C, 1}=\frac{9 \sqrt{3}}{8 \pi} \\
R=\frac{9 \sqrt{3}}{4 \pi} \frac{V_{m}}{I_{O U T}}
\end{gathered}
$$

- $R \quad 9 \times \uparrow$
- $i_{R} 3 \times \downarrow$
- $v_{R} 3 \times \uparrow$
- $p_{R}=v_{R} i_{R}$ remains the same, no way to save any power


## renormalization

- input currents not sinusoidal any more
- convenient to use $I_{\text {base }}=I_{O U T}$, instead of $I_{\text {base }}=I_{m}$
- mutual relation?
- $I_{\text {OUT }}=\frac{3 \sqrt{3}}{2 \pi} I_{m} \approx 0.82699 I_{m}$, but only when the input currents are sinusoidal


## $j_{A}, j_{B}$, and $j_{C}$, the $3^{\text {rd }}$ harmonic injection


$j_{A}$, spectrum, real part, the $3^{\text {rd }}$ harmonic injection

$j_{B}$, spectrum, real part, the $3^{\text {rd }}$ harmonic injection

$j_{C}$, spectrum, real part, the $3^{\text {rd }}$ harmonic injection


## $j_{I A}, j_{I B}$, and $j_{C}$, the $3^{\text {rd }}$ harmonic injection




$j_{I A}$, spectrum, real part, the $3^{\text {rd }}$ harmonic injection

$j_{I B}$, spectrum, real part, the $3^{\text {rd }}$ harmonic injection

$j_{R 1}, j_{R 2}$, and $j_{R 3}$, the $3^{\text {rd }}$ harmonic injection



$j_{R 1}$, spectrum, real part, the $3^{\text {rd }}$ harmonic injection

$j_{S 1}, j_{S 2}$, and $j_{S 3}$, the $3^{\text {rd }}$ harmonic injection



$j_{S 1}$, spectrum, real part, the $3^{\text {rd }}$ harmonic injection

$j_{1}, j_{2}$, and $j_{3}$, the $3^{\text {rd }}$ harmonic injection



how $j_{1}$ is obtained, the $3^{\text {rd }}$ harmonic injection



how $j_{1}$ is obtained, the $3^{\text {rd }}$ harmonic injection

how $j_{2}$ is obtained, the $3^{\text {rd }}$ harmonic injection



how $j_{3}$ is obtained, the $3^{\text {rd }}$ harmonic injection




## a note about RMSs

$$
\begin{gathered}
J_{A, 0}=J_{B, 0}=1 \\
J_{A R M S}=J_{B R M S}=\sqrt{\frac{33}{32}} \approx 1.0155
\end{gathered}
$$

much better than for magnetic current injection devices (where the increase was about $13 \%$ )

## a note about bidirectional switches ...



- turned out to be easier than expected
- control problems, interaction with the diode bridge
- interphase shorts should not be allowed to occur ...
- primarily in the diode bridge!


## conclusions

- switching current injection device
- injection only where needed
- bidirectional switches required
- control of the switches, switching at $2 f_{0} \ldots$
- three current injection networks proposed, although there are many more
- the optimal and the third harmonic current injection
- "dominant" resistor suitable for resistance emulation
- three times lower currents in comparison to magnetic current injection devices
- lower RMS of the diode bridge load currents ...


## "future work"

- how to restore the power taken by the current injection network?

