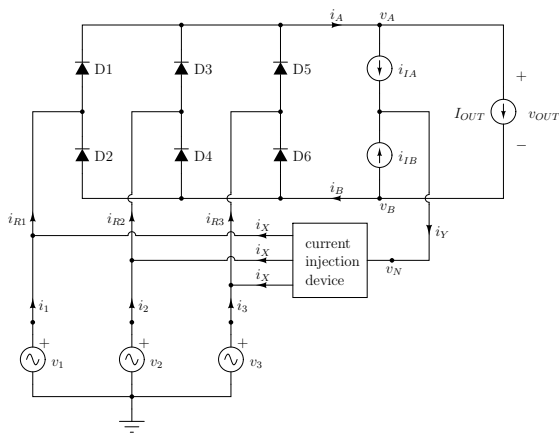


aim ...

## Switching Current Injection Device

- ▶ current injection devices inject to all three of the phases ...
- ▶ but only one phase really needs injection!
- ▶ is there a way to inject only where needed?
- ▶ is there a way to get rid of the current injection device?
- ▶ something smaller, lighter, cheaper, ...

let's get back to basic current injection ...



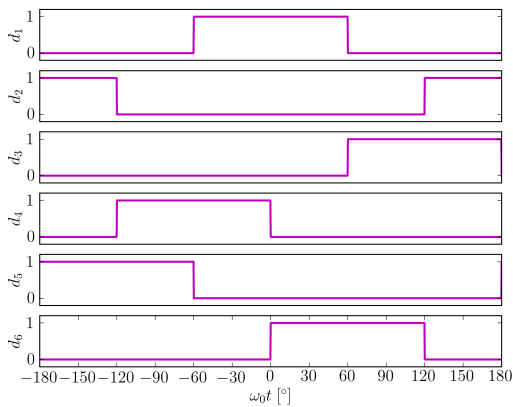
as always, assume ...

$$m_1 = \cos(\omega_0 t)$$

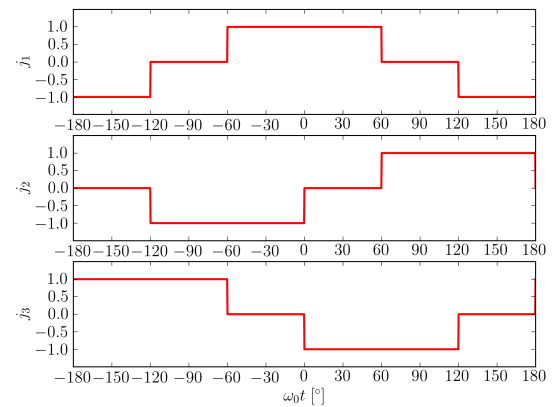
$$m_2 = \cos\left(\omega_0 t - \frac{2\pi}{3}\right)$$

$$m_3 = \cos\left(\omega_0 t - \frac{4\pi}{3}\right)$$

diodes ...



currents without injection

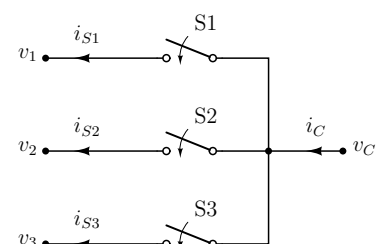


what do we need?

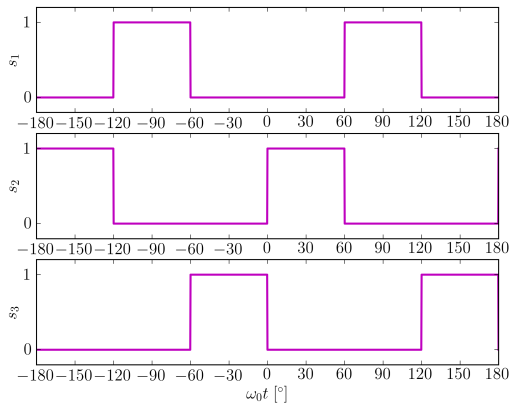
**conclusion:**

current injection is needed in the phase whose voltage is neither minimal neither maximal in the considered time point

let's do it!



how to operate the switches?



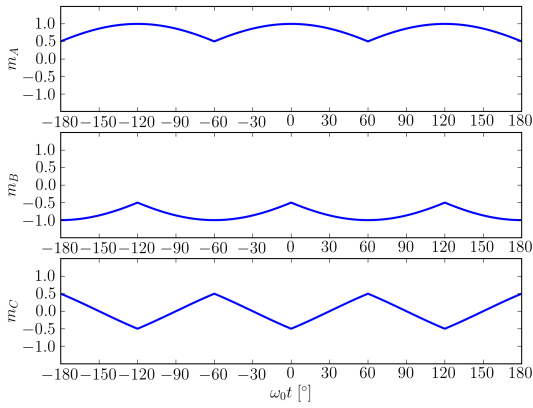
some Boole algebra ...

$$s_1 = \neg d_1 \wedge \neg d_2$$

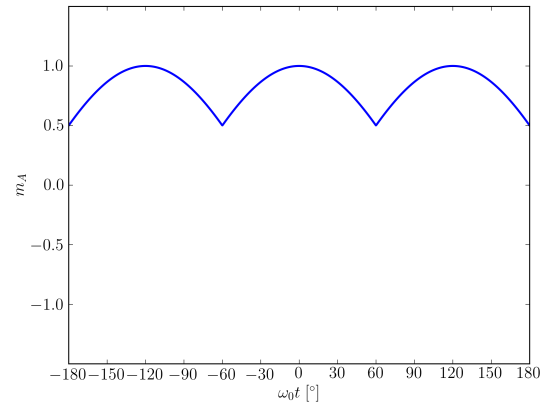
$$s_2 = \neg d_3 \wedge \neg d_4$$

$$s_3 = \neg d_5 \wedge \neg d_6$$

and the voltages are defined ...



$m_A$ , waveform



$m_A$ , analytical

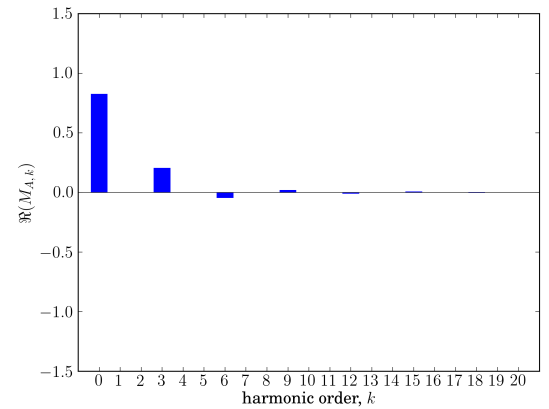
$$m_A = \max(m_1, m_2, m_3)$$

$$m_A = M_{A0} + \sum_{k=1}^{\infty} M_{A,k} \cos(3k\omega_0 t)$$

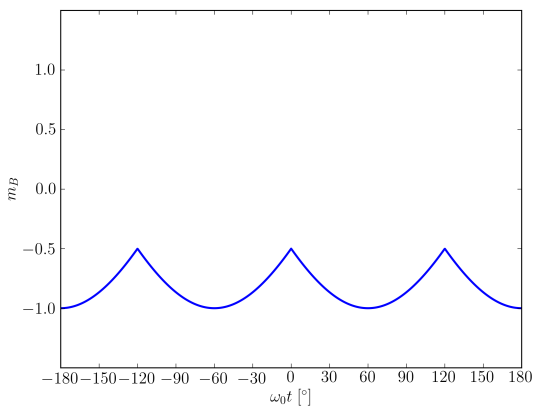
$$M_{A0} = \frac{3\sqrt{3}}{2\pi}$$

$$M_{A,k} = \frac{3\sqrt{3}}{\pi} \frac{(-1)^{k+1}}{9k^2 - 1}$$

$m_A$ , spectrum, real part



$m_B$ , waveform



$m_B$ , analytical

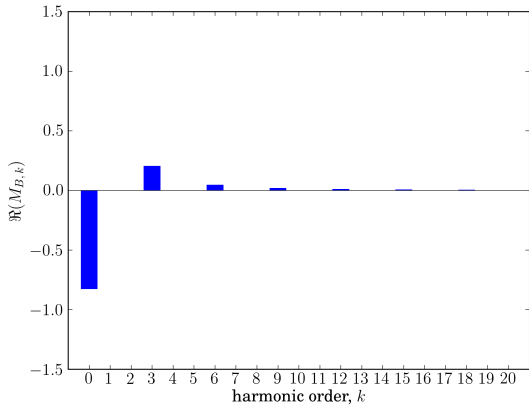
$$m_B = \min(m_1, m_2, m_3)$$

$$m_B = M_{B0} + \sum_{k=1}^{\infty} M_{B,k} \cos(3k\omega_0 t)$$

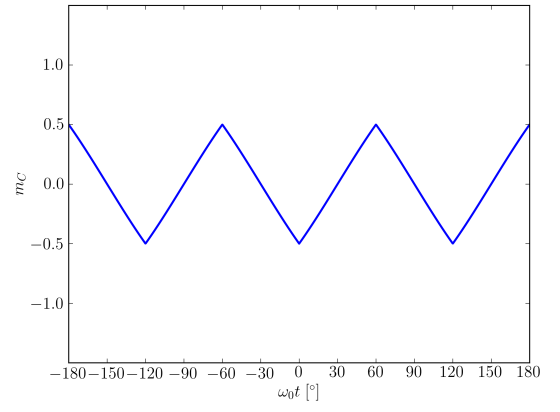
$$M_{B0} = -\frac{3\sqrt{3}}{2\pi}$$

$$M_{B,k} = \frac{3\sqrt{3}}{\pi} \frac{1}{9k^2 - 1}$$

$m_B$ , spectrum, real part



$m_C$ , waveform, this is a new one ...



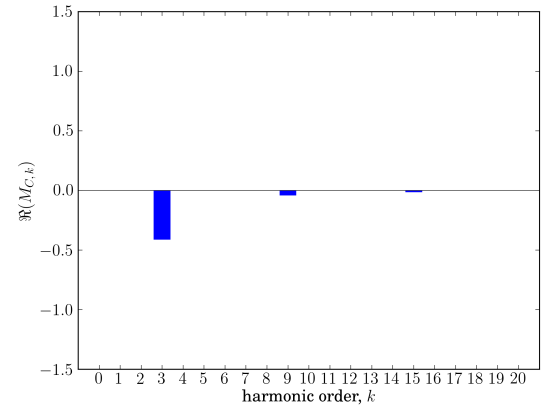
$m_C$ , analytical

$$m_C = -m_A - m_B$$
$$m_C = \sum_{k=1,3,5,\dots}^{\infty} M_{C,k} \cos(3k\omega_0 t)$$
$$M_{C,k} = -\frac{6\sqrt{3}}{\pi} \frac{1}{9k^2 - 1}$$

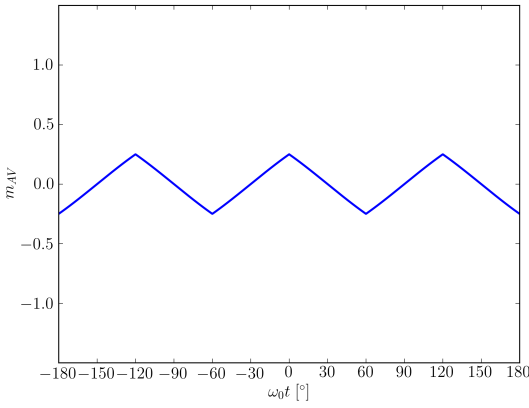
this is a (sort of) new spectrum to deal with ...

and there is no DC component in it ...

$m_C$ , spectrum, real part



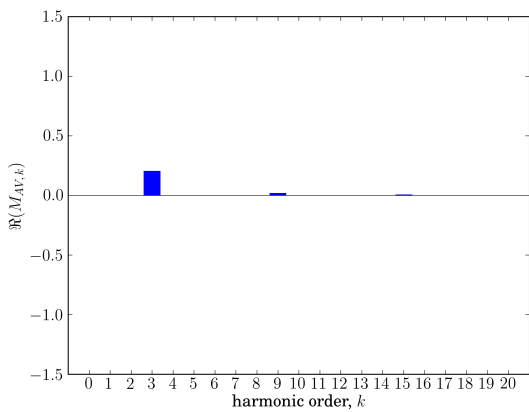
since we are already here,  $m_{AV}$ , waveform



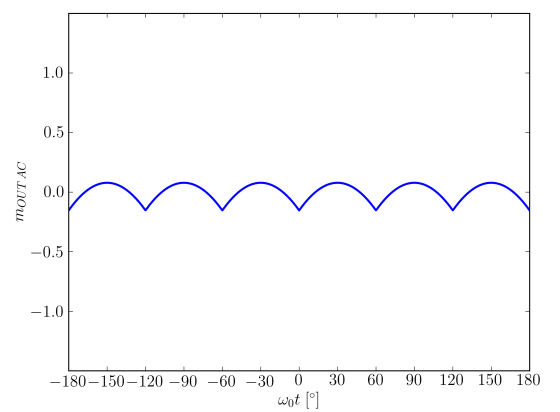
$m_{AV}$ , analytical

$$m_{AV} = \frac{m_A + m_B}{2} = -\frac{m_C}{2}$$
$$m_{AV} = \sum_{k=1,3,5,\dots}^{\infty} M_{AV,k} \cos(3k\omega_0 t)$$
$$M_{AV,k} = \frac{3\sqrt{3}}{\pi} \frac{1}{9k^2 - 1}$$

$m_{AV}$ , spectrum, real part



$m_{OUTAC}$ , waveform



$$m_{OUT\ AC} = m_{OUT} - M_{OUT} = m_A - M_{A,0} - m_B + M_{B,0}$$

$$m_{OUT\ AC} = \sum_{k=2,4,6,\dots}^{\infty} M_{OUT\ AC,k} \cos(3k\omega_0 t)$$

$$M_{OUT\ AC,k} = -\frac{6\sqrt{3}}{\pi} \frac{1}{9k^2 - 1}$$

we would like to have ...

$$j_1 = \cos(\omega_0 t)$$

$$j_2 = \cos\left(\omega_0 t - \frac{2\pi}{3}\right)$$

$$j_3 = \cos\left(\omega_0 t - \frac{4\pi}{3}\right)$$

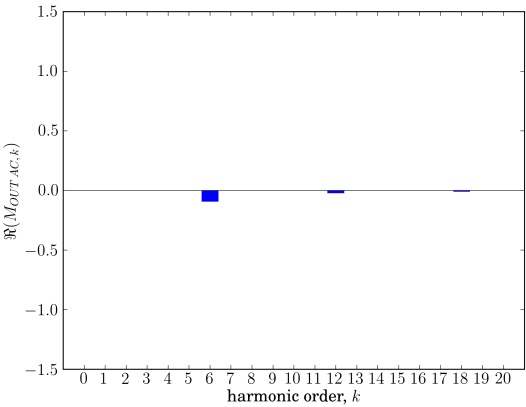
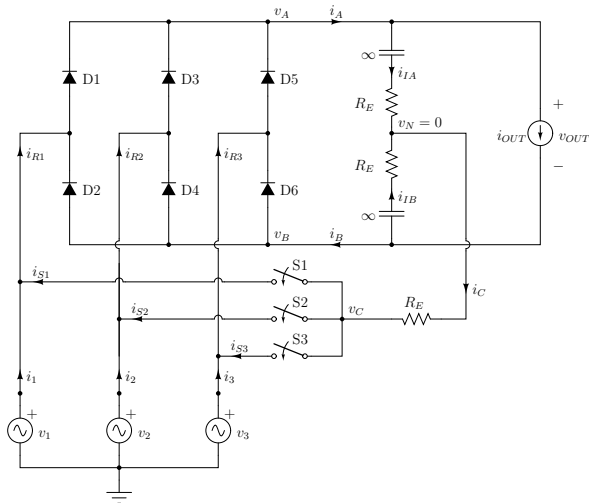
a note, again: normalized amplitude is 1; if actual amplitude is *I<sub>m</sub>*, the normalization is

$$j_X \triangleq \frac{i_X}{I_m}$$

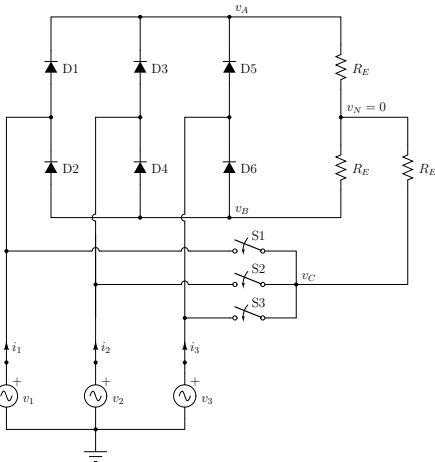
analysis ...

- ▶ regarding the inputs, diodes and switches perform useless function
- ▶ each phase observes *R<sub>E</sub>*, which is perfect!
- ▶ which of the resistors is the hottest one?
- ▶ or better to ask, which one is the coldest?
- ▶ let's separate AC and DC, you already know the trick ...

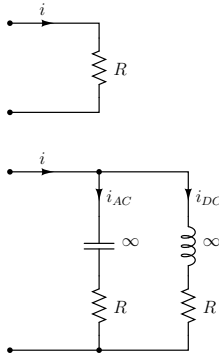
voila!



and there is a way to get it ...



... the trick ...



hint:  $L = R^2 C$ , if  $\infty$  is too big

analysis ...

- ▶ ... and we have the circuit!
- ▶ not a long mathematical derivation?
- ▶ actually, it's invented right now, while working on this presentation (April 15, 2012, 00:06:53)
- ▶ not the first experience of this kind, 1999, ...

Predrag Pejović

“A Novel Low Harmonic Three Phase Rectifier”

*IEEE Transactions on Circuits and Systems I: Fundamental Theory and Applications*,  
vol. 49, no. 7, pp. 955–965, July 2002

after lots of trouble ...

although, the derivation presented here is much shorter, and the circuit is slightly different ...

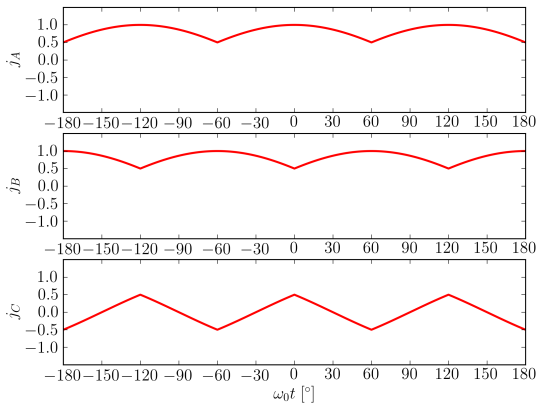
power and efficiency ...

$$P_{IN} = \frac{3}{2}$$
$$P_{OUT} = M_{OUT} J_{OUT} = \frac{3\sqrt{3}}{\pi} \times \frac{3\sqrt{3}}{2\pi} = \frac{27}{2\pi^2}$$
$$P_{INJ} = P_{IN} - P_{OUT} = \frac{3\pi^2 - 27}{2\pi^2}$$

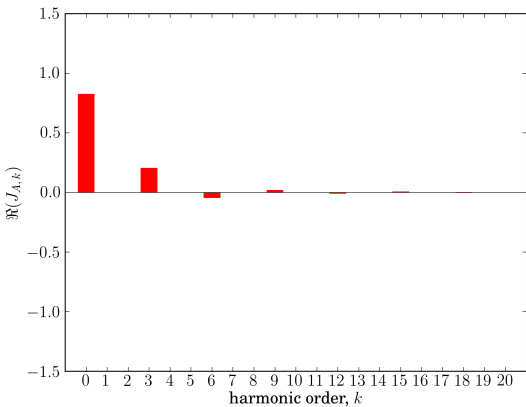
$$\eta = \frac{P_{OUT}}{P_{IN}} = \frac{9}{\pi^2} \approx 91.19\%$$

already familiar with the results?

$j_A$ ,  $j_B$ , and  $j_C$



$j_A$ , spectrum, real part



$j_A$ , analytical

$$j_A = \max(j_1, j_2, j_3)$$
$$j_A = J_{A0} + \sum_{k=1}^{\infty} J_{A,k} \cos(3k\omega_0 t)$$
$$J_{A0} = \frac{3\sqrt{3}}{2\pi} = J_{OUT}$$
$$J_{A,k} = \frac{3\sqrt{3}}{\pi} \frac{(-1)^{k+1}}{9k^2 - 1}$$

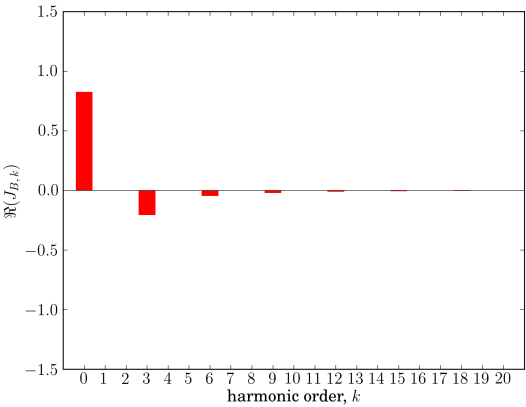
the same as  $m_A$

$j_B$ , analytical

$$j_B = -\min(j_1, j_2, j_3)$$
$$j_B = J_{B0} + \sum_{k=1}^{\infty} J_{B,k} \cos(3k\omega_0 t)$$
$$J_{B0} = \frac{3\sqrt{3}}{2\pi} = J_{OUT}$$
$$J_{B,k} = -\frac{3\sqrt{3}}{\pi} \frac{1}{9k^2 - 1}$$

the same as  $-m_B$

$j_B$ , spectrum, real part



$j_C$ , analytical

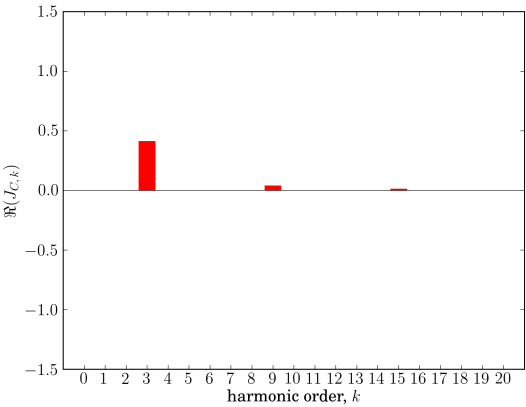
$$j_C = j_A - j_B$$

$$j_C = \sum_{k=1,3,5,\dots}^{\infty} J_{C,k} \cos(3k\omega_0 t)$$

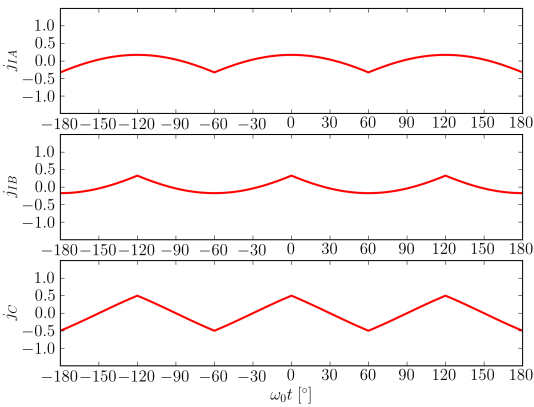
$$J_{C,k} = \frac{6\sqrt{3}}{\pi} \frac{1}{9k^2 - 1}$$

proportional to  $m_{AV}$  and  $-m_C$ ; going to be important

$j_C$ , spectrum, real part



$j_{IA}$ ,  $j_{IB}$ , and  $j_C$



currents ...

$$j_{IA} = j_A - J_{OUT}$$

$$j_{IB} = J_{OUT} - j_B$$

$$J_{OUT} = \frac{3\sqrt{3}}{2\pi}$$

analytical description, spectra, ...

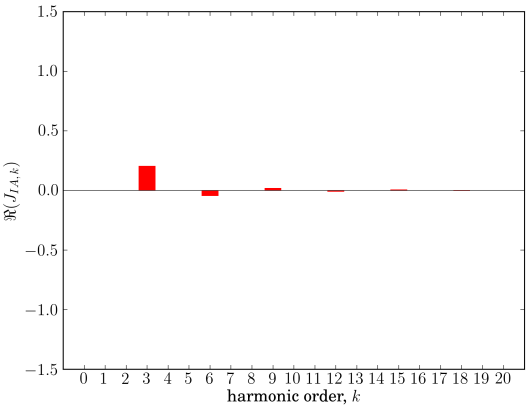
$j_{IA}$ , analytical

$$j_{IA} = \max(j_1, j_2, j_3) - \frac{3\sqrt{3}}{\pi}$$

$$j_{IA} = \sum_{k=1}^{\infty} J_{A,k} \cos(3k\omega_0 t)$$

$$J_{IA,k} = \frac{3\sqrt{3}}{\pi} \frac{(-1)^{k+1}}{9k^2 - 1}$$

$j_{IA}$ , spectrum, real part



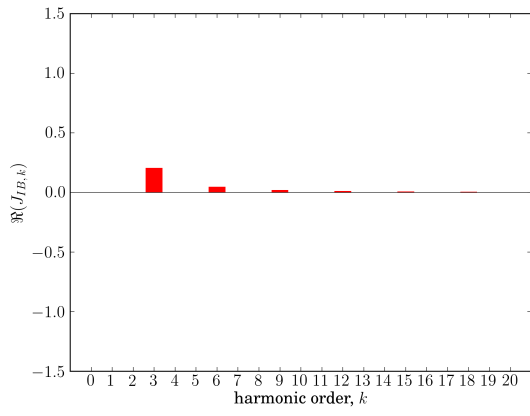
$j_{IB}$ , analytical

$$j_{IB} = \frac{3\sqrt{3}}{2\pi} - \min(j_1, j_2, j_3)$$

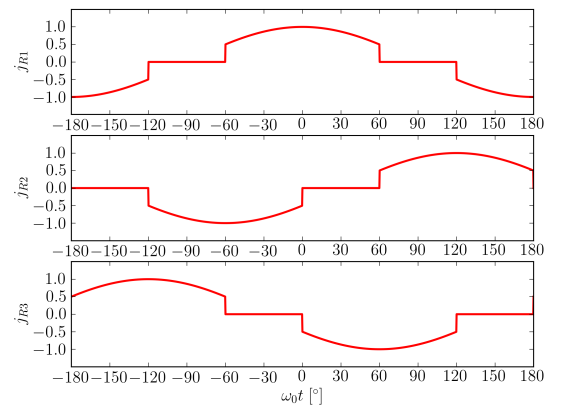
$$j_{IB} = \sum_{k=1}^{\infty} J_{B,k} \cos(3k\omega_0 t)$$

$$J_{IB,k} = \frac{3\sqrt{3}}{\pi} \frac{1}{9k^2 - 1}$$

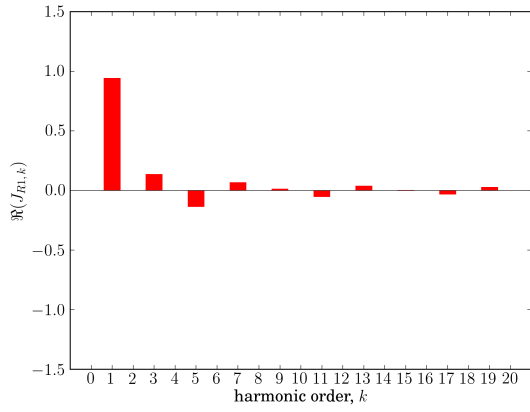
$j_{IB}$ , spectrum, real part



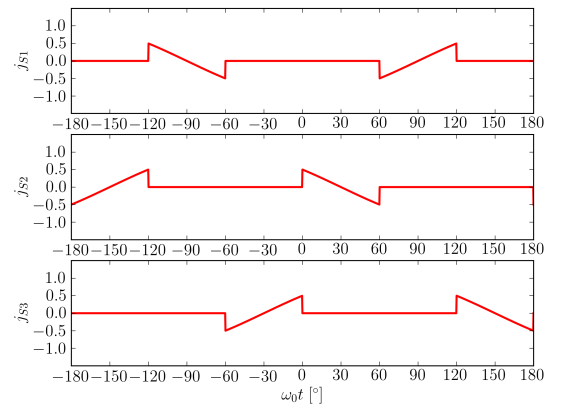
detour:  $j_{R1}$ ,  $j_{R2}$ , and  $j_{R3}$



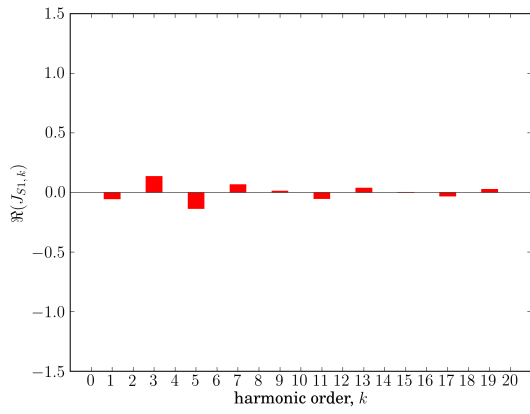
$j_{R1}$ , spectrum, real part



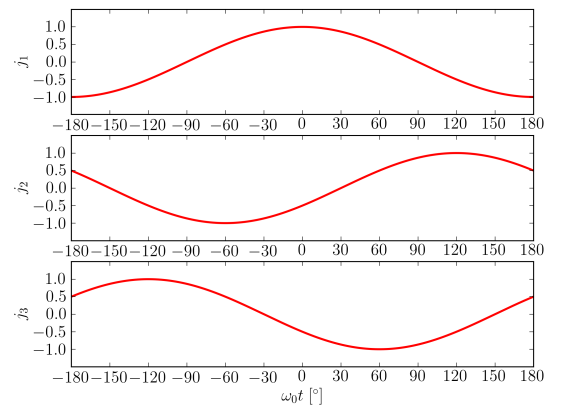
$j_{S1}$ ,  $j_{S2}$ , and  $j_{S3}$



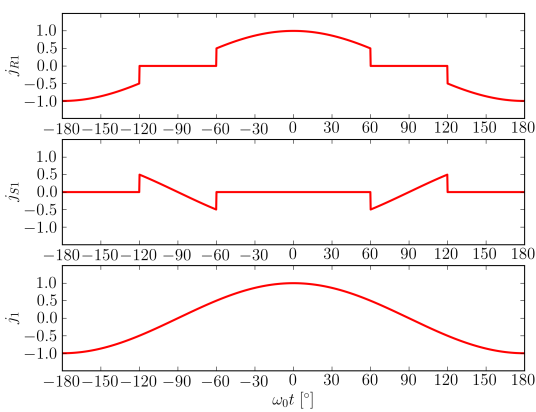
$j_{S1}$ , spectrum, real part



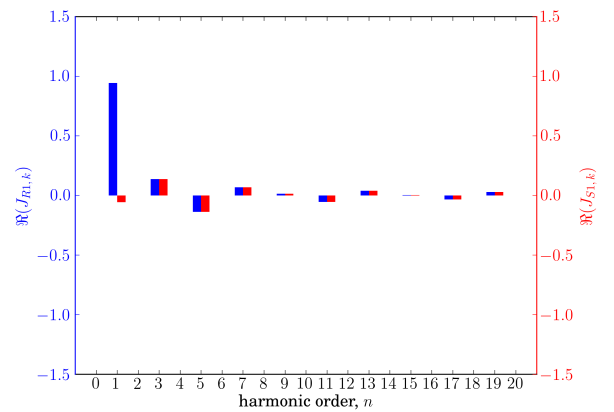
$j_1$ ,  $j_2$ , and  $j_3$



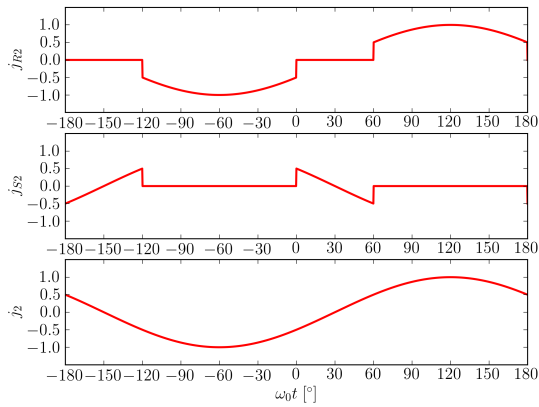
how  $j_1$  is obtained?



how  $j_1$  is obtained: spectral approach



how  $j_2$  is obtained?



$j_{odd}$  and  $j_{even}$

matter of convenience:

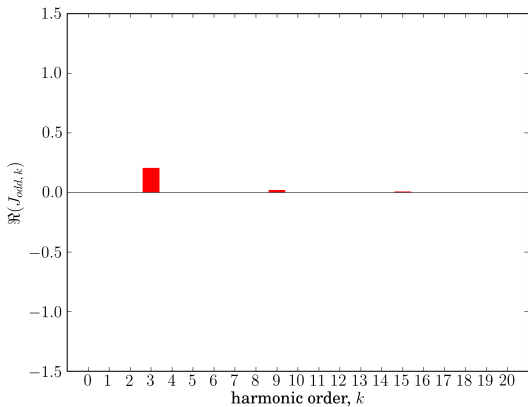
$$j_{IA} = j_{odd} + j_{even}$$

$$j_{IB} = j_{odd} - j_{even}$$

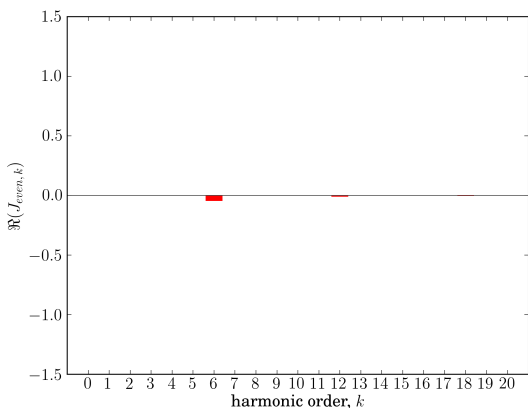
$$j_{odd} = \frac{j_{IA} + j_{IB}}{2}$$

$$j_{even} = \frac{j_{IA} - j_{IB}}{2}$$

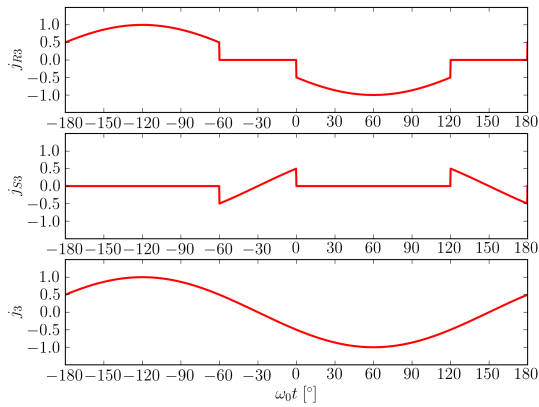
$j_{odd}$ , spectrum, real part



$j_{even}$ , spectrum, real part



how  $j_3$  is obtained?



$j_{odd}$ , spectrum

$$j_{odd} = \sum_{k=1,3,5,\dots}^{\infty} J_{odd,k} \cos(3k\omega_0 t)$$

$$J_{odd,k} = \frac{3\sqrt{3}}{\pi} \frac{1}{9k^2 - 1}$$

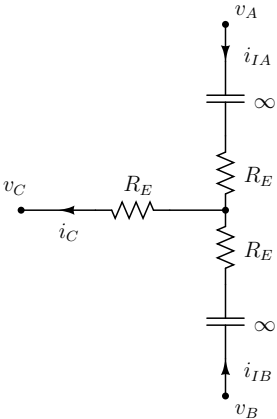
$j_{even}$ , spectrum

$$j_{even} = \sum_{k=2,4,6,\dots}^{\infty} J_{even,k} \cos(3k\omega_0 t)$$

$$J_{even,k} = -\frac{3\sqrt{3}}{\pi} \frac{1}{9k^2 - 1}$$

proportional to  $m_{OUT\ AC}$ ; going to be important

so, we have SCIN #0





wxMaxima ... I won't be able to complete this job manually, it's too boring ...

$$J_{IARMS} = J_{IBRMS} = \frac{\sqrt{4\pi^2 + 3\pi\sqrt{3} - 54}}{2\pi\sqrt{2}}$$

$$P_{vertical R_E} = (J_{IARMS})^2 = \frac{4\pi^2 + 3\pi\sqrt{3} - 54}{8\pi^2} \approx 0.0228$$

vertically placed  $R_E$  resistors are not so hot

$$J_{CRMS} = \sqrt{\frac{2\pi - 3\sqrt{3}}{4\pi}}$$

$$P_{horizontal R_E} = (J_{CRMS})^2 = \frac{2\pi - 3\sqrt{3}}{4\pi} \approx 0.0865$$

horizontally placed  $R_E$  is much hotter, almost 4 times; why  $4\times$ ?

- ▶ transformer not required
- ▶ too many resistors
- ▶ one resistor takes the most of the power

## some relations ...

$$\frac{m_{AV}}{j_{odd}} = 1$$

$$\frac{m_C}{j_{odd}} = -2$$

$$\frac{m_{OUTAC}}{j_{even}} = 2$$

to be used while creating new current injection networks ...

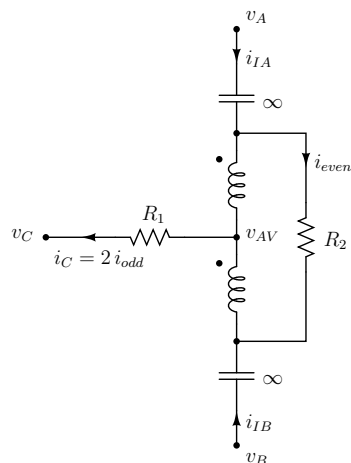
## $R_1$ and $R_2$

$$R_1 = \frac{m_{AV} - m_C}{2j_{odd}} R_E = \frac{3}{2} R_E$$

$$R_2 = \frac{m_{OUTAC}}{j_{even}} = 2 R_E$$

and now you know why “some relations” were needed for

## SCIN #1, separatism



## some power accounting, again ...

$$J_{oddRMS} = \frac{\sqrt{2\pi - 3\sqrt{3}}}{4\sqrt{\pi}}$$

$$P_1 = \frac{3}{2} (J_{oddRMS})^2 = \frac{3}{8} \left( 2 - \frac{3\sqrt{3}}{\pi} \right) \approx 0.1298$$

$$J_{evenRMS} = \frac{\sqrt{3}}{4\pi} \sqrt{2\pi^2 + 3\pi\sqrt{3} - 36}$$

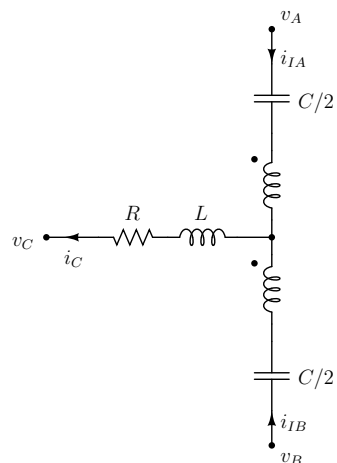
$$P_2 = \frac{3(2\pi^2 + 3\pi\sqrt{3} - 36)}{8\pi^2} \approx 0.0024$$

dissipation is dominant on  $R_1$

## an idea makes a new idea ...

- ▶ power on  $R_2$  is small ...
- ▶ let's get rid of  $R_2$ !
- ▶ result: the same as the 3<sup>rd</sup> harmonic CIN #3 for  $Q = 0$ ,  $THD \approx 4\%$
- ▶ could we get  $Q \neq 0$ ?
- ▶ definitely!
- ▶ do we need it?
- ▶ well, maybe for passive resistance emulation, to be talked about later

## SCIN #2, the 3<sup>rd</sup> harmonic one



$$3\omega_0 = \frac{1}{\sqrt{LC}}$$

$$j_C = \frac{1}{2} \cos(3\omega_0 t)$$

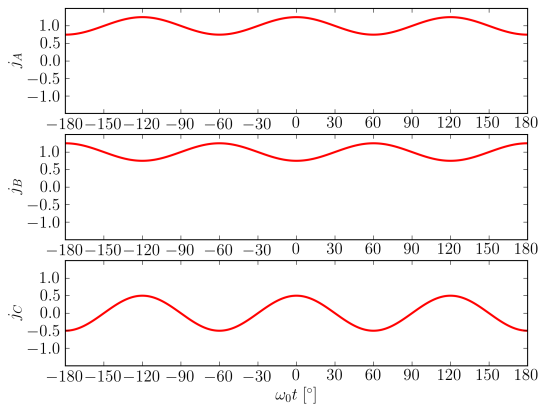
$$M_{AV,1} - M_{C,1} = \frac{9\sqrt{3}}{8\pi}$$

$$R = \frac{9\sqrt{3}}{4\pi} \frac{V_m}{I_{OUT}}$$

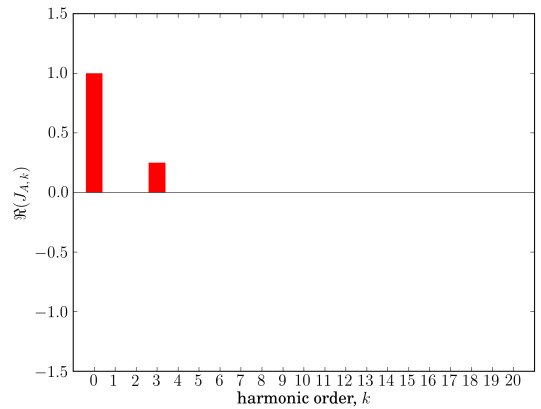
- ▶  $R$   $9 \times \uparrow$
- ▶  $i_R$   $3 \times \downarrow$
- ▶  $v_R$   $3 \times \uparrow$
- ▶  $p_R = v_R i_R$  remains the same, no way to save any power

- ▶ input currents not sinusoidal any more
- ▶ convenient to use  $I_{base} = I_{OUT}$ , instead of  $I_{base} = I_m$
- ▶ mutual relation?
- ▶  $I_{OUT} = \frac{3\sqrt{3}}{2\pi} I_m \approx 0.82699 I_m$ , but only when the input currents are sinusoidal

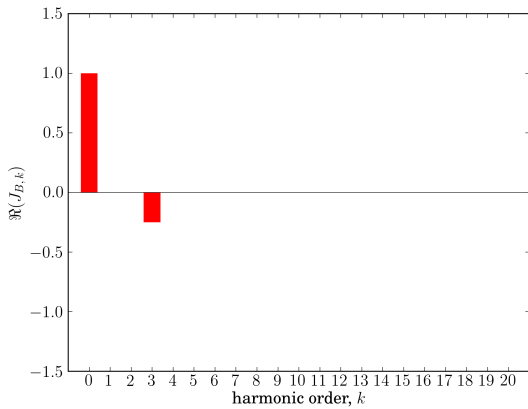
$j_A$ ,  $j_B$ , and  $j_C$ , the 3<sup>rd</sup> harmonic injection



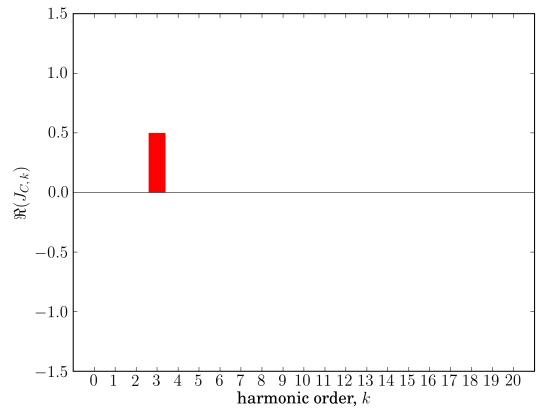
$j_A$ , spectrum, real part, the 3<sup>rd</sup> harmonic injection



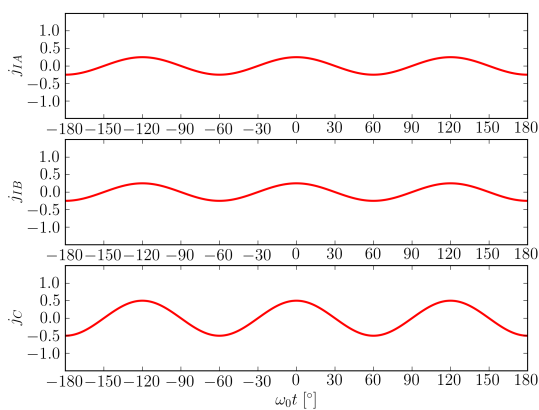
$j_B$ , spectrum, real part, the 3<sup>rd</sup> harmonic injection



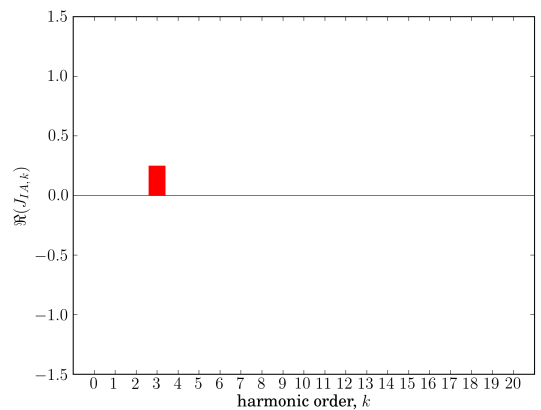
$j_C$ , spectrum, real part, the 3<sup>rd</sup> harmonic injection



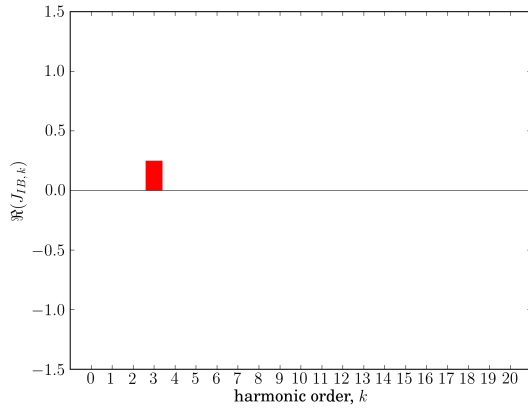
$j_{IA}$ ,  $j_{IB}$ , and  $j_C$ , the 3<sup>rd</sup> harmonic injection



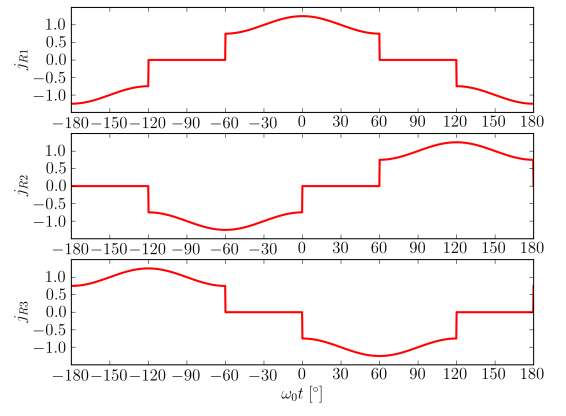
$j_{IA}$ , spectrum, real part, the 3<sup>rd</sup> harmonic injection



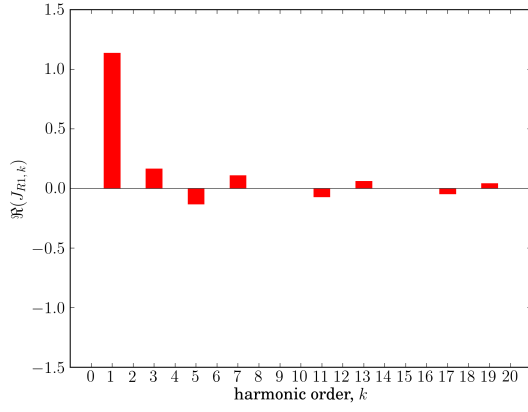
$j_{IB}$ , spectrum, real part, the 3<sup>rd</sup> harmonic injection



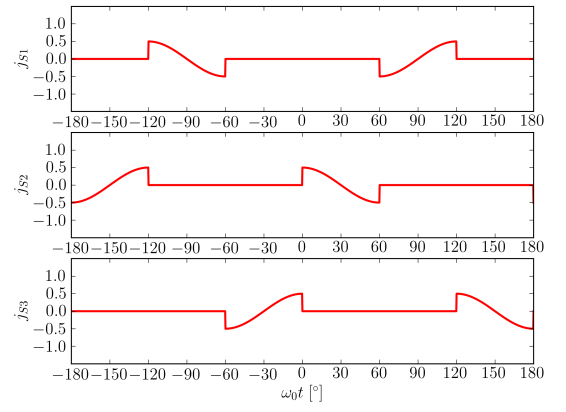
$j_{R1}$ ,  $j_{R2}$ , and  $j_{R3}$ , the 3<sup>rd</sup> harmonic injection



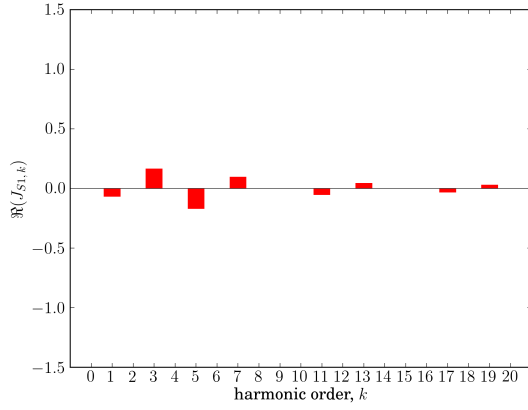
$j_{R1}$ , spectrum, real part, the 3<sup>rd</sup> harmonic injection



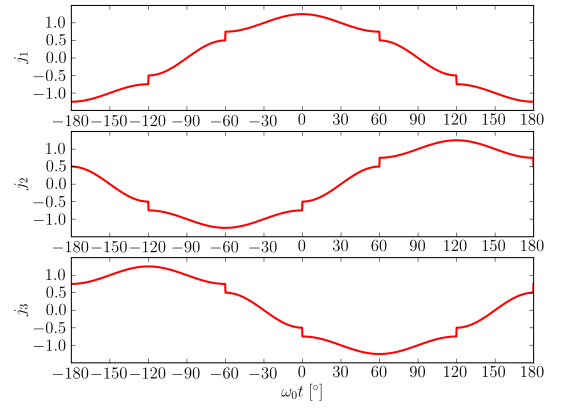
$j_{S1}$ ,  $j_{S2}$ , and  $j_{S3}$ , the 3<sup>rd</sup> harmonic injection



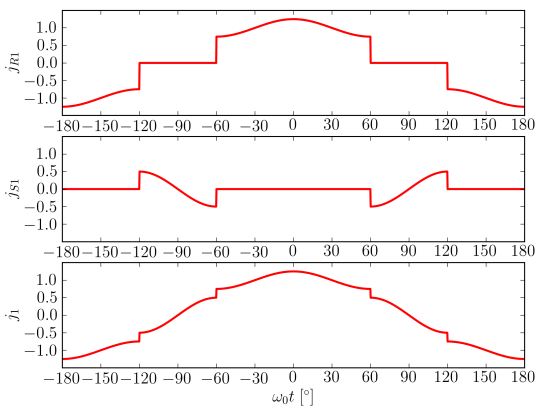
$j_{S1}$ , spectrum, real part, the 3<sup>rd</sup> harmonic injection



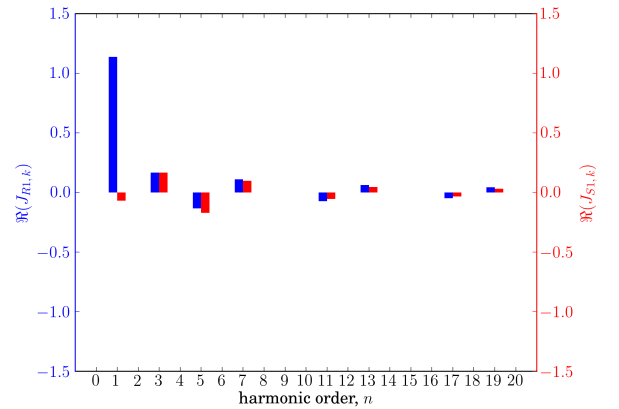
$j_1$ ,  $j_2$ , and  $j_3$ , the 3<sup>rd</sup> harmonic injection



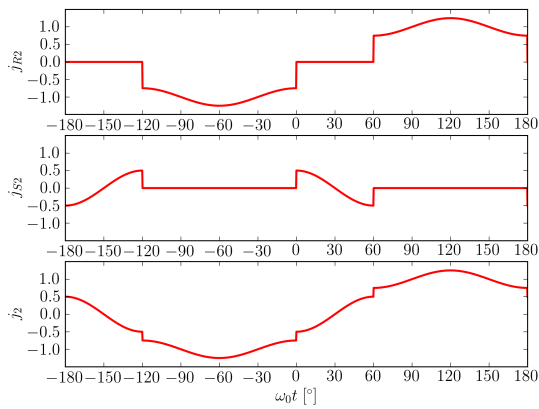
how  $j_1$  is obtained, the 3<sup>rd</sup> harmonic injection



how  $j_1$  is obtained, the 3<sup>rd</sup> harmonic injection



how  $j_2$  is obtained, the 3<sup>rd</sup> harmonic injection



a note about RMSs

$$J_{A,0} = J_{B,0} = 1$$

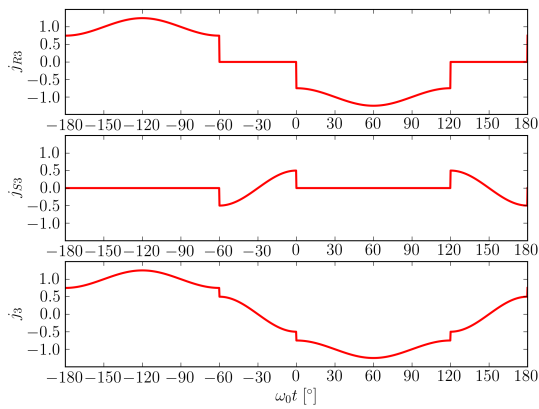
$$J_{ARMS} = J_{BRMS} = \sqrt{\frac{33}{32}} \approx 1.0155$$

much better than for magnetic current injection devices  
(where the increase was about 13%)

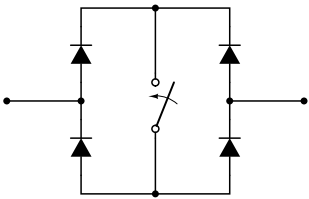
conclusions

- ▶ switching current injection device
- ▶ injection only where needed
- ▶ bidirectional switches required
- ▶ control of the switches, switching at  $2f_0 \dots$
- ▶ three current injection networks proposed, although there are many more
- ▶ the optimal and the third harmonic current injection
- ▶ “dominant” resistor suitable for resistance emulation
- ▶ three times lower currents in comparison to magnetic current injection devices
- ▶ lower RMS of the diode bridge load currents ...

how  $j_3$  is obtained, the 3<sup>rd</sup> harmonic injection



a note about bidirectional switches ...



- ▶ turned out to be easier than expected
- ▶ control problems, interaction with the diode bridge
- ▶ interphase shorts should not be allowed to occur ...
- ▶ primarily in the diode bridge!

“future work”

- ▶ how to restore the power taken by the current injection network?