## Introduction

- three-phase diode bridge rectifier -


## what is this all about?



## input voltages

$$
\begin{gathered}
v_{1}=V_{m} \cos \left(\omega_{0} t\right) \\
v_{2}=V_{m} \cos \left(\omega_{0} t-\frac{2 \pi}{3}\right) \\
v_{3}=V_{m} \cos \left(\omega_{0} t-\frac{4 \pi}{3}\right) \\
v_{k}=V_{m} \cos \left(\omega_{0} t-(k-1) \frac{2 \pi}{3}\right), \quad k \in\{1,2,3\}
\end{gathered}
$$

## input voltages, waveforms





## normalization of voltages

$$
m_{X} \triangleq \frac{v_{X}}{V_{m}}
$$

$$
m_{1}=\cos \left(\omega_{0} t\right)
$$

$$
\begin{aligned}
& m_{2}=\cos \left(\omega_{0} t-\frac{2 \pi}{3}\right) \\
& m_{3}=\cos \left(\omega_{0} t-\frac{4 \pi}{3}\right)
\end{aligned}
$$

## voltages?

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v_{k}=V_{m} \cos \left(\omega_{0} t-(k-1) \frac{2 \pi}{3}\right), \quad k \in\{\mathbf{1}, 2,3\}
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## $v_{1}$, spectrum



## $v_{2}$, spectrum



## $v_{3}$, spectrum



## voltages, quantitative characterization

| $k$ | $V_{k R M S}$ | $T H D\left(v_{k}\right)$ |
| :---: | :---: | :---: |
| 1 | 103.83 V | $3.34 \%$ |
| 2 | 103.70 V | $2.77 \%$ |
| 3 | 105.12 V | $3.06 \%$ |

all graphs and data PyLab processed

## $T H D$

And what is THD?

$$
T H D \triangleq \frac{\sqrt{\sum_{k=2}^{\infty} I_{k R M S}^{2}}}{I_{1 R M S}}
$$

Parseval's identity:

$$
I_{R M S}^{2}=\sum_{k=1}^{\infty} I_{k R M S}^{2} \quad \text { assumed } \quad I_{0}=0
$$

results in

$$
T H D \triangleq \frac{\sqrt{I_{R M S}^{2}-I_{1 R M S}^{2}}}{I_{1 R M S}}
$$

simple, but important computational issues, finite sums ...

## normalization of currents and time

$$
j_{X} \triangleq \frac{i_{X}}{I_{O U T}}
$$

unless otherwise noted

$$
\varphi \triangleq \omega_{0} t
$$

good: physical dimensions lost, reduced number of variables, results are generalized, core of the problem focused
bad: physical dimensions lost, perfect double-check tool is lost
how does it work? part 1: theory


## one of the three: D1, D3, D5




## $v_{A}$, analytical



$$
m_{A}=\max \left(m_{1}, m_{2}, m_{3}\right)
$$

## $v_{A}$, spectrum



$$
m_{A}=\frac{3 \sqrt{3}}{2 \pi}\left(1+2 \sum_{k=1}^{\infty} \frac{(-1)^{k+1}}{9 k^{2}-1} \cos \left(3 k \omega_{0} t\right)\right)
$$

what about $v_{B}$ ?


## one of the three, again: D2, D4, D6




## $v_{B}$, analytical



$$
m_{B}=\min \left(m_{1}, m_{2}, m_{3}\right)
$$

## $v_{B}$, spectrum



$$
m_{B}=\frac{3 \sqrt{3}}{2 \pi}\left(-1+2 \sum_{k=1}^{\infty} \frac{1}{9 k^{2}-1} \cos \left(3 k \omega_{0} t\right)\right)
$$

## the output voltage, $v_{O U T}$


$m_{O U T}=m_{A}-m_{B}=\max \left(m_{1}, m_{2}, m_{3}\right)-\min \left(m_{1}, m_{2}, m_{3}\right)$

## $v_{O U T}$, spectrum



$$
m_{O U T}=\frac{3 \sqrt{3}}{\pi}\left(1-2 \sum_{k=1}^{\infty} \frac{1}{36 k^{2}-1} \cos \left(6 k \omega_{0} t\right)\right)
$$

## currents?

$$
\begin{aligned}
& i_{1}(t)=\left(d_{1}(t)-d_{2}(t)\right) I_{O U T} \\
& i_{2}(t)=\left(d_{3}(t)-d_{4}(t)\right) I_{O U T} \\
& i_{3}(t)=\left(d_{5}(t)-d_{6}(t)\right) I_{O U T}
\end{aligned}
$$

## states of the diodes



## the input currents



## consider $i_{1}$



## spectra of the input currents



## spectra of the input currents, analytical

$$
\begin{gathered}
j_{1}(t)=\sum_{k=1}^{+\infty} J_{1 C, k} \cos \left(k \omega_{0} t\right) \\
J_{1 C, k}=\frac{2 \sqrt{3}}{\pi}\left\{\begin{aligned}
-\frac{1}{k}, & k=6 n-1 \\
\frac{1}{k}, & k=6 n+1 \\
0, & \text { otherwise }
\end{aligned}\right.
\end{gathered}
$$

double-check:

$$
P_{I N}=\frac{3}{2} \times 1 \times \frac{2 \sqrt{3}}{\pi}=\frac{3 \sqrt{3}}{\pi}=P_{O U T}
$$

## numerical verification, Gibbs phenomenon



## THD of the input currents

$$
\begin{gathered}
I_{k R M S}=\sqrt{\frac{2}{3}} I_{O U T} \\
I_{k R M S, 1}=\frac{\sqrt{6}}{\pi} I_{O U T} \\
T H D \triangleq \frac{\sqrt{I_{k R M S}^{2}-I_{k R M S, 1}^{2}}}{I_{k R M S, 1}} \\
T H D=\sqrt{\frac{\pi^{2}}{9}-1} \approx 31.08 \%
\end{gathered}
$$

Parseval's identity based formula turned out to be useful

## voltages and currents



## some more parameters

$$
X_{R M S} \triangleq \sqrt{\frac{1}{2 \pi} \int_{0}^{2 \pi}\left(x\left(\omega_{0} t\right)\right)^{2} d\left(\omega_{0} t\right)}, \quad x \in\{i, v\}
$$

already used for the THD

$$
\begin{gathered}
S \triangleq I_{R M S} V_{R M S} \\
P \triangleq \frac{1}{2 \pi} \int_{0}^{2 \pi} v\left(\omega_{0} t\right) i\left(\omega_{0} t\right) d\left(\omega_{0} t\right) \\
P F \triangleq \frac{P}{S}
\end{gathered}
$$

$$
D P F \triangleq \cos \phi_{1}
$$

## and if the voltages are sinusoidal ...

$$
\begin{gathered}
S=V_{R M S} I_{R M S} \\
P=V_{R M S} I_{1, R M S} \cos \phi_{1} \\
P F=\frac{P}{S}=\frac{I_{1, R M S}}{I_{R M S}} \cos \phi_{1}=\frac{I_{1, R M S}}{I_{R M S}} D P F \\
D P F=\cos \varphi_{1} \\
T H D=\frac{\sqrt{I_{R M S}^{2}-I_{1, R M S}^{2}}}{I_{1, R M S}}=\sqrt{\left(\frac{I_{R M S}}{I_{1, R M S}}\right)^{2}-1}
\end{gathered}
$$

i.e. everything depends on the current waveform and its position

## some more parameters, plain rectifier

$$
\begin{gathered}
I_{k R M S}=\sqrt{\frac{2}{3}} I_{O U T} \quad V_{k R M S}=\frac{1}{\sqrt{2}} V_{m} \\
S=3 \times \sqrt{\frac{2}{3}} I_{O U T} \times \frac{1}{\sqrt{2}} V_{m}=\sqrt{3} V_{m} I_{O U T} \\
P=V_{O U T} I_{O U T}=\frac{3 \sqrt{3}}{\pi} V_{m} I_{O U T} \\
P F=\frac{3}{\pi} \approx 95.5 \% \\
D P F=1
\end{gathered}
$$

actually, not so bad; $T H D$ is the problem
back to the rectifier:
how does it work? part 2: experiment

input, at $I_{O U T}=3 \mathrm{~A}$

input, at $I_{O U T}=3 \mathrm{~A}$

input, at $I_{\text {OUT }}=3 \mathrm{~A}$

input, at $I_{\text {OUT }}=3 \mathrm{~A}$

input, at $I_{O U T}=6 \mathrm{~A}$

input, at $I_{\text {OUT }}=6 \mathrm{~A}$

input, at $I_{\text {OUT }}=6 \mathrm{~A}$

input, at $I_{\text {OUT }}=6 \mathrm{~A}$

input, at $I_{O U T}=9 \mathrm{~A}$

input, at $I_{O U T}=9 \mathrm{~A}$

input, at $I_{\text {OUT }}=9 \mathrm{~A}$

input, at $I_{\text {OUT }}=9 \mathrm{~A}$

output, at $I_{\text {OUT }}=3 \mathrm{~A}$

output, at $I_{\text {OUT }}=3 \mathrm{~A}$


## output, at $I_{\text {OUT }}=6 \mathrm{~A}$


output, at $I_{\text {OUT }}=6 \mathrm{~A}$


## output, at $I_{\text {OUT }}=9 \mathrm{~A}$



## output, at $I_{\text {OUT }}=9 \mathrm{~A}$


in quantitative terms, input, 1st

| $I_{\text {OUT }}$ | $k$ | $I_{k R M S}[\mathrm{~A}]$ | $V_{k R M S}[\mathrm{~V}]$ | $S_{k}[\mathrm{VA}]$ | $P_{k}[\mathrm{~W}]$ |
| ---: | ---: | ---: | ---: | ---: | ---: |
| 0 A | 1 | - | 101.29 | - | - |
|  | 2 | - | 100.63 | - | - |
|  | 3 | - | 102.40 | - | - |
| 3 A | 1 | 2.60 | 98.23 | 255.01 | 245.16 |
|  | 2 | 2.61 | 97.73 | 254.60 | 244.37 |
|  | 3 | 2.63 | 98.82 | 259.71 | 251.00 |
| 6 A | 1 | 5.12 | 94.87 | 485.41 | 466.87 |
|  | 2 | 5.12 | 94.34 | 482.67 | 464.25 |
|  | 3 | 5.13 | 96.80 | 496.95 | 477.08 |
| 9 A | 1 | 7.59 | 92.38 | 701.53 | 673.86 |
|  | 2 | 7.64 | 91.95 | 702.47 | 675.16 |
|  | 3 | 7.66 | 94.04 | 720.00 | 692.30 |

in quantitative terms, input, 2nd

| $I_{\text {OUT }}$ | $k$ | $P F_{k}$ | $T H D\left(i_{k}\right)[\%]$ | $T H D\left(v_{k}\right)[\%]$ |
| ---: | ---: | ---: | ---: | ---: |
| 0 A | 1 | - | - | 4.33 |
|  | 2 | - | - | 3.75 |
|  | 3 | - | - | 4.75 |
| 3 A | 1 | 0.9614 | 30.50 | 4.17 |
|  | 2 | 0.9598 | 29.57 | 3.86 |
|  | 3 | 0.9665 | 29.97 | 5.38 |
| 6 A | 1 | 0.9618 | 29.26 | 3.87 |
|  | 2 | 0.9618 | 28.37 | 3.66 |
|  | 3 | 0.9600 | 28.31 | 3.87 |
| 9 A | 1 | 0.9605 | 28.00 | 4.01 |
|  | 2 | 0.9611 | 27.21 | 3.92 |
|  | 3 | 0.9615 | 27.06 | 4.19 |

## in quantitative terms, output

| $I_{\text {OUT }}[\mathrm{A}]$ | $V_{\text {OUT }}[\mathrm{V}]$ | $P_{\text {OUT }}[\mathrm{W}]$ | $P_{\text {IN }}[\mathrm{W}]$ | $\eta[\%]$ |
| ---: | ---: | ---: | ---: | ---: |
| 0.00 | 239.79 | 1.07 | -0.81 | - |
| 3.21 | 229.51 | 736.72 | 740.53 | 99.49 |
| 6.27 | 221.23 | 1386.56 | 1408.20 | 98.46 |
| 9.41 | 212.91 | 2004.12 | 2041.32 | 98.18 |

## overall impressions

- pretty good rectifier
- simple, robust, cheap
- good symmetry
- excellent DPF
- acceptable $P F$
- poor THD (but not that poor)
- up to this point:
- diode bridge rectifier analyzed
- measurement tools developed
- is there a way to do something with the $T H D$ ?


## fruitless effort \#1: shaping the output current



## fruitless effort \#1: waveforms





## fruitless effort \#1: quantitative

- $T H D=30.79 \%$
- not a big deal of an improvement
- only one degree of freedom, $i_{O U T}$
- shaping $i_{1}, i_{2}$, and $i_{3}$ is the goal
- two degrees of freedom needed, since $i_{1}+i_{2}+i_{3}=0$
fruitless effort \#2: additional deegree of freedom



## fruitless effort \#2: waveforms





## fruitless effort \#2: neutral current



## fruitless effort \#2: quantitative

- $T H D=24.76 \%$
- somewhat better
- all of $i_{1}, i_{2}$, and $i_{3}$ cannot be fixed by programming $i_{A}$ and $i_{B}$ in this circuit
- example: $i_{1}=i_{A}, i_{2}=-i_{B}$, no way to fix $i_{3}$
- gaps in the input currents in both of the "patches"
- the additional degree of freedom is taken by $i_{N}$
- which is a disaster of itself
- we would need another degree of freedom to fix $i_{N}$
- but this is a wrong approach, $i_{N}$ was not an issue before


## conclusions

- three-phase diode bridge rectifier analyzed
- quantitative measures of rectifier performance introduced
- measurement tools developed
- theoretical predictions related to experiments
- gaps in the input currents identified as a problem
- how to fill in the gaps?
- an answer is current injection ...

